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5P191 ${ }^{2}$
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VOL. 20 No. 1 JANUARY 1991


The No 1 Magazine for Electronic \& Computer Projects

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## EVERYDAY <br> = =ccinoccce <br> FEBRUARY ISSUE ON SALE FRIDAY 4TH JANUARY 1991



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## MAINS CONDITIONER D

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601133 Box for transmitter


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## GCSE PROJECTS

Help! Please, teachers, parents, students help us and help yourselves. Almost everyday my postbag carries letters from GCSE students working on crazy projects. Our new series Project Development for GCSE has been written by a GCSE assessor and hopefully the series will go a long way in helping to sort out the basic problems which arise time and time again. I know it would be great to be able to build a radio link and p.a. system to contact the whole school, but it is not a practical project.

## IMPORTANT

Can I say that in my opinion no student should be allowed to even consider anything that involves a radio link of any sort. First of all such a project could not be legally used in the UK and secondly, working at radio frequencies requires specialist knowledge and equipment that is not available at secondary school level.
Sensing objects/people at a distance is equally fraught, it would for instance be necessary to use radar or a laser to measure the distance between two cars travelling at speed. It seems like a good idea to build a "keep your distance" device but it is well outside the capabilities of GCSE students and teachers (and many professional engineers too).
GCSE assessors are not very concerned about the complexity of the project - try something very complex and it is likely to end in dismal failure. Make sure you have plenty of time to consider just how difficult a particular project might be before you commit yourself to it. Discuss it with teachers, parents, someone who works in the electronics industry, anyone who can provide some realistic idea of just what might be involved in realising a working project

## DIFFICULT

I find it very hard to reply to students" letters telling them that they should not be undertaking their chosen project - particularly if they have already spent a lot of time and effort searching for possible solutions. Please read our new series - I believe it will help everyone to get more students achieving Grade $\mathbf{C}$ marks or better.


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All reasonable precautions are taken to ensure that the advice and data given to readers is reliable. We cannot however guarantee it and we cannot accept legal responsibility for it.

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 EQUIPMENTWe would like to advise readers that certain items of radio transmitting and telephone equipment which may be advertised in our pages cannot be legally used in the U.K. Readers should check the law before using any transmitting or telephone equipment as a fine, confiscation of equipment and/or imprisonment can result from illegal use. The laws vary from country to country: overseas readers should check local laws.



#### Abstract

The second constructional project to accompany our Design Your Dwn Circuits series takes the form of a Simple Intercom. This unit provides two-way communication over a simple two-wire circuit and, as with all of the practical constructional projects in this series, a number of modifications are suggested so that the more intrepid constructor cancustomise the unit to his or her own particular requirements.


AsImple intercom can prove to be an asset in almost every home. Such a unit can provide an instant and effective means of two-way communications between nursery and lounge, kitchen and workshop, or wherever else it may be needed. All that is required is a medium gain amplifier and a low-power output stage connected along the lines shown in Fig. 1. A simple oscillator circuit can be added to provide a tone-call facility, as shown in Fig. 2(a). Remote tone calling can be achieved by a.c. coupling the remote
speaker/microphone and switching the d.c. path to the remote station, as in Fig. 2(b).
Our simple intercom has been designed to satisfy the need for a low-cost system which can be readily expanded to permit the connection of extra stations. It also provides a useful practical application of the circuit design techniques described in Part Two of our series.

## Circuit description

The complete circuit of the Simple Intercom is shown in Fig. 3. The circuit is based

on a single operational amplifier stage. IC1, which provides a voltage gain of several hundred (determined by the ratio of R7 to R4). The operational amplifier stage operates in conjunction with a complementary transistor output stage. TR2 and TR3. This stage provides the additional current gain required to drive a 350 ohm impedance loudspeaker, LSI.
A single +9 V supply rail is used and hence, to provide the symmetrical positive and negative supply rails required by' IC 1 . R5 and R6 form a potential divider to set the voltage at the non-inverting input of 1 Cl at 4.5 V (i.e. the "half rail voltage"). This also fixes the quiescent voltage at the output of ICl and at the junction of R8 and R9 at 4.5 V . C2 provides decoupling of the half-supply voltage rail.
Transistors TR4 and TR5 are arranged as an astable oscillator (the frequency of which is determined by R13, R14, C7 and C8. R10 and C6 constitute a simple lowpass filter to shape the square wave oscillator output signal before it is applied to the input of the amplifier circuit. The supply to the oscillator circuit is switched on and off by means of TR1. This transistor operates as a "saturated switch"; TR1 will conduct when either S2 (a normally open push-button switch) is operated or the remote call-button places a short circuit across the intercom cable. When TRI conducts. the d.c. supply to TR4 and TR5 is enabled and oscillation commences.
Switch S3 acts as a "talk/listen" switch. This switch (a double-pole double-throw miniature toggle switch) changes over the internal and remote loudspeakers so that communications can take place in both directions.
The circuit diagram of the remote station is shown in Fig. 4. This circuit requires


Fig. 1. Block diagram of a basic intercom


Fig. 3. Complete circuit diagram of the simple intercom.
little comment other than that SI (a normally open push button switch) places a short-circuit across the line whenever it is operated. Such a short-circuit causes TRI to conduct and thus enables the d.c. supply to the tone oscillator circuit of TR4 and TR5.

## CONSTRUCTION

Construction is very straightforward and the vast majority of the components for the simple intercom are assembled on a singlesided printed circuit board measuring approximately $50 \times 100 \mathrm{~mm}$. The layout of the printed circuit board is shown in Fig. 5.

Components should be assembled on the printed circuit board in the following sequence: p.c.b. headers. d.i.l. socket, resis-

## COMPONENTS

## Resistors <br> R1, R2,

R5, R6 2 k 2 (4 off)
R3 22k
R4, R12.
$\mathrm{R}^{\text {R15 }} 1 \mathrm{k}$ (3 off)

## See

SUOP
TALK
$\begin{array}{ll}\text { R8,R9 } & \left.\begin{array}{l}330 \mathrm{k} \\ \text { Retal film } 0.5 \mathrm{~W} \\ \text { ( } 2 \text { off }\end{array}\right)\end{array}$ R10,R11 680 (2 off)
R13,R14 10k (2 off)
All resistors with the exception of R8 and R9 are 0.25W 5\%

## Capacitors

| Capa | rs |
| :---: | :---: |
| C2, C3 | $100 \mu$ radial elect. 16 V ( 2 off ) |
| C4 | $220 \mu$ radial elect. 35 V |
| C5,C6 | $100 \mathrm{n} \mathrm{100V}$ polyester |
| C7,C8 | 100 n 50 V disc ceramic ( 2 off ) |
| C9 | $10 \mu$ radial elect. 16 V |

## Semiconductors

| TR1,TR3 | BC178 (2 off) |
| :--- | :--- |
| TR2,TR4, |  |
| TR5 | BC108 (3 off) |
| IC1 | 741 |
| D1,D2 | 4 4148 (2 off) |

## Miscellaneous

S1 DPDT miniature toggle switch
S2 Miniature normally-open push-button
S3 DPDT miniature toggle switch
LS1 350hm 4 inch speaker
PL1 5-way straight p.c.b. header ( 0.1 inch pitch)
PL2 10 -way straight p.c.b
header ( 0.1 inch pitch)
SK1 Standard 0.25 inch jack socket
Printed circuit board available from the EE PCB Service, order code EE719: plastic p.c.b. fixing pillars with selftapping No. 6 fixing screws ( 3 off); snap-fit PP3-type battery connector; 8 -pin low-profile di.I. connector; sloping front ABS enclosure approx. $215 \times$ $130 \times 73 \mathrm{~mm}$ (see text).

## REMOTE STATION <br> Capacitor <br> C1 $100 \mu$ axial elect. 25 V

## Miscellaneous

S1 Min. normally-open push-button
LS1 350 hm 4 inch speaker
SK1 Standard open 0.25 inch jack socket; ABS enclosure (see text).


Fig. 5. P.C.B. layout for the simple intercom.

inserted into its socket (taking care to observe the correct orientation) and the board should then be mounted in the base of the plastic enclosure by means of three short snap-fit p.c.b. mounting pillars.

## CASES

Both master and remotestation units are housed in sloping front ABS enclosures (measuring $215 \times 130 \times 43 \mathrm{~mm} \mathrm{~min}$., 73 mm max.). Such enclosures generally have a removable aluminium front panel which can be drilled and cut to accommodate the switches (S1. S2 and S3) and the loudspeaker. LS!

Use of a loudspeaker of relatively large size (i.e. 4 inch diameter) is highly recommended since. in the author's experience. smaller units are rarely satisfactory on the grounds of both sensitivity and sound quality. In addition, medium impedance 35 ohm units should be used (not the more usual 8 ohm units). The remote station is connected by means of a standard 0.25 inch jack connector which is mounted at the lower left-hand side of the enclosure. A battery holder (for a PP3 9 V battery) can consist of a simple L-shaped aluminium bracket secured to the base of the case.

The front panel should be carefully marked out beforę drilling and cutting takes place. As usual, there is nothing particularly critical about the layout of the unit and constructors may wish to experiment with the location of the loudspeaker and front panel controls.

The loudspeaker will require a round hole of suitable diameter $(80 \mathrm{~mm}$ on the prototype). Such an aperture may be cut by means of a tension file and then filed (using a half-round file) to an exact circular profile. Alternatively, an adjustable tankcutter can be used (such a tool was used in the construction of the prototype). The same procedure should be adopted for the remote station.
Connections to the printed circuit board are made using two 0.1 inch pitch printed circuit board headers. A five-way header (PLI) is used for the master station callbutton and the remote station jack connector whilst a 10 -way header (PL2) is used for the talk/listen switch (S3), loudspeaker and supply connections (via $\mathbf{S} 1$ ). The procedure for terminating the female connectors which mate with the headers was described

Fig. 6. Internal wiring of the simple intercom.
tors, capacitors, diodes and transistors. As with all of our projects, it is important to ensure that all of the components are correctly located. Furthermore, in the case of the polarised components (such as electrolytic capacitors. diodes and transistors) it is absolutely essential to ensure that each component is correctly orientated.
When consiruction of the printed circuit board has been completed (and before inserting ICl into its socket) it is well worth carrying out a careful visual check of both the upper and lower sides of the board. The upper (component) side of the printed circuit board should be examined to ensure that the components have been correctly located whilst the lower (copper track) side of the board should be checked to ensure that there are no dry joints or solder bridges between adjacent tracks. This precaution will only take a few minutes to carry out but can be instrumental in preventing much heartache at a later stage!
When assembly of the printed circuit board has been completed. ICI should be

in the first of our constructional projects which appeared last month.
Coloured stranded 0.1 inch pitch ribbon cable is used to make connections to the front and rear panels. The following colour coding is recommended:

## PLI

|  |  | PL1 |
| :---: | :--- | :--- |
| Pin | Colour | Connection to: |
| 1 | Brown | S2 (master call button) <br> 2 |
| Red | S2 (master call button) |  |
| 3 | Orange | Jack connector (outer/ <br> sleeve) |
| 4 | Yellow | Jack connector (inner/tip) <br> 5 |
|  |  | Not connected |

The internal wiring of the simple intercom is shown in Fig. 6.

## TESTING

Before testing the simple intercom, carefully check the wiring of the p.c.b. and front panel mounted components. The remote station should be connected to the master station by means of a length of two-core flex of appropriate length terminated with standard 0.25 inch jacks (it is important to ensure that the outer/sleeve connections are connected together).

Connect a PP3 battery to the battery connector and insert a milliammeter in one lead to the battery in order to measure the supply current to the unit. Now switch the unit on and measure the supply current. This should be in the range 5 mA to 15 mA . If this is not the case. disconnect the supply and carefully check the wiring and p.c.b.


## Fig. 7. Adding an l.e.d. indicator.

If the supply current is correct, switch the unit to the "master talk" position and check that a signal appears at the remote unit. Depress the master call button ( $\mathbf{S} 2$ ) and check that a loud tone is heard at the remote station. Switch to the "master listen" position and check communications in the reverse direction in a similar fashion. If either direction of communication is missing, check the wiring to S3. If the tone-call facility is not operating, check the connection of TRI and the circuit around TR4 and TR5.

## MODIFICATIONS

A number of useful modifications may be made to enhance the performance of the basic intercom circuit. The suggestions made here are provided as "food for thought" and should make a starting point for further development. Constructors are invited to report their own modifications to be incorporated in the Reuders' Feeclhuck which will appear in the final part of our Design series.

## Adding an l.e.d. indicator

An l.e.d. indicator can very easily be added in order to provide the user with a visual warning that the unit is switched on and is "live". Fig. 7 shows the necessary circuit modifications.

## Additional stations

Further remote stations can be added quite easily by means of an additional "station selector" switch and extra jack sockets (one for each additional remote


Fig. 8. Adding further remote stations.
station). Fig. 8 shows the modification required for a single-master/three- remote station system. The station selector switch can be a simple rotary switch (e.g. IP 12W with the rotation stop suitably adjusted).

## Increased output power

In some situations (e.g. in a noisy environment) it may be necessary to provide extra audio output from the intercom. This can easily be achieved by increasing the supply voltage to 18 V by connecting two PP3 batteries in series. The following component changes should be made:

COMPONENT

## Standard version Higher power version

| TR2 | BC108 | BC142 |
| :--- | :--- | :--- |
| TR3 | BC178 | BC143 |
| R16 | 1 kS | 3 k 9 |
| R17 | 1 kS | 3 k 9 |
| R3 | 22 k | 47 k |
| C3 | $100 \mu 16 \mathrm{~V}$ | $100 \mu 35 \mathrm{~V}$ |

## SPECIFICATIONS

$\left.\begin{array}{lc}\text { Output power: } & 300 \mathrm{~mW} \text { r.m.s. max. } \\ \text { Voltage gain: } & \begin{array}{l}250 \text { approx. }\end{array} \\ \text { Number of stations: } \\ \text { I master, I slave (may } \\ \text { be easily expanded) }\end{array}\right\}$

Number of stations: 1 master, I slave (may be easily expanded)
Call tone:
Supply voltage: 9 V (PP3 battery) (tone call)


# ANALOGIC TEST PROBE 

## MARK STUART

# A very versatile test probe that measures voltage and indicates a.c. waveforms 

THE idea for this project came about whilst trying to find a fault in an EE project. David, our new YTS, was using a logic probe to sort out the 16 l.e.d. Chaser Light project (EE Oct 86) and getting some curious results.
The Chaser Light circuit consists of a CMOS two gate oscillator driving a binary up/down counter. the outputs of which drive 16 l.e.d.s via a 4-10-16 line decoder. Normally the outputs and inputs of the logic i.c.s in these circuits are either at full supply or ground and spend little time in between.

The logic probe is fine in these circumstances and often leads directly to the fault. This circuit was different. because there was a voltage on one pin which was hovering around half of the supply. The logic probe read this as a 1 (high) but the i.c. was reading it as a 0 (low)

After some thought a multimeter came to the rescue and the fault was soon sorted out. It was apparent that if a probe with some type of continuous voltage indication had been available the fault would have been found much more easily.
Such a probe would also be useful in most other types of circuit. especially if it could read higher voltages and dis-
criminate between a.c. and d.c. The circuit finally produced (in its original crude stripboard form) has been used on many circuits with success and has now been refined to make a simple useful and interesting project.

## DISPLAY

The type of display required for a probe needs to be compact, shock proof and easily read. A miniature panel meter would be too delicate, an I.c.d digital display would be difficult to read. and both would be too expensive. Bargraph displays in the form of a row of l.e.d.s are rugged and relatively cheap. They have limited resolution (the ability to indicate very small changes) but for a simple probe displaying 10 levels gives sufficient information (much more than the standard two-level logic probe.

## CIRCUIT

As Fig. I shows the circuit for the Analogic Test Probe is very simple, relying on the LM3914N I.e.d. bargraph driver i.c. This is not the cheapest of i.c.s but its simplicity and low external component count make it good value.

Before settling on this approach several other ways of driving a column of l.e.d.s

Fig. 1. Complete circuit diagram for the Analogic Test Probe.

were tried. Some were cheaper. but needed much more p.c.b. space and were more complicated. Using the LM3914N gives several other benefits. It has a stable internal reference voltage, programmable l.e.d. current, and a high input impedance. It is also very tolerant of overloads.

A simplified internal schematic diagram of the i.c. as it is used in this application is shown in Fig. 2. Pin 9 allows the display to be set to either Bar mode (where a column of l.e.d.s light up) or Dot mode (where only one l.e.d. lights at a time). Dot mode is chosen because of its much lower current demand, and is selected by leaving pin 9 open circuit. There is some small overlap between l.e.d.s so that the display appears to move smoothly up and down as the input voltage varies.

The brightness of the display is determined by the value of the resistor from pin 7 to the negative supply. This is set to 4 mA per l.e.d. by a $3 k 3$ resistor, and gives plenty of brightness with good quality standard l.e.d.s, and very good battery life.

The power supply arrangements for the probe are shown in Fig 1. A three-way slide switch SI allows power to be taken from either an internal 9 V battery or via a

## COMPONENTS

| Resistors |  | See |
| :---: | :---: | :---: |
| R1 | 270k | CH |
| R2 | 430 k |  |
| R3 | 30k |  |
| R4 | 100k | Page |
| R5 | 3k3 | n $\pm 5 \%$ |

All $\pm 1 \%$ metal film, except where stated.

## Capacitors

C1 $\quad 100 \mathrm{n} 100 \mathrm{~V}$ ceramic plate C2 $2 \mu 2$ sub.min radial elect. 63 V

## Semiconductors

D1-D10 3 mm red l.e.d.s ( 10 off)
D11 1 N4148 diode
IC1 LM3914N bargraph driver

## Miscellaneous

S1,S2 Single-pole 3-way slide switch (2 off)
Printed circuit board available from the EE PCB Service, order code EE720; red and black insulated crocodile clips and matching extra flexible leads (approx. 50 cm long); PP3 battery clip and battery; M3 fixings for probe; probe case.

## Approx cost <br> guidance only

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crocodile clip lead from the circuit under test. Voltages between 7 and 18 volts can be used as external supplies and up to 25 V can be applied without damage. Current limiting resistor R6 and a polarity protection diode DII prevent damage from incorrect power connection.
The input to pin 5 of IC1 needs to be 1.25 V to give a full scale reading (top l.e.d. on). This is the input of a non-inverting op-amp and has a very low input leakage current and very high impedance. These characteristics make it easy to add a potential divider circuit so that the probe will read the required voltage range. In the prototype the ranges are $0-5 \mathrm{~V}$ and $0-10 \mathrm{~V}$ d.c., and $0-5 \mathrm{~V}$ r.m.s. a.c.

On the a.c. range a capacitor is added in series with the input path. This blocks all d.c. up to the capacitor voltage rating of 100 V , and allows only any a.c. element present to pass. The i.c. displays only positive half cycles and ignores the negative ones, giving a simple form of rectification. The resulting display will indicate the waveform as well as the peak voltage and gives full scale for 5 V r.m.s.

As the a.c. wave rises and falls from its peak value the intermediate le.d.s will respond and their brightness will indicate the amount of time that the signal spends at each voltage level. A triangular wave will light all l.e.d.s evenly as it passes through each value, whilst a square wave will light
the board after the switch has been fitted. To keep the board as small as possible one pin of each switch must be bent under the switch body, this should be done carefully so that the mechanism is not distorted in any way.
The l.e.d.s must be mounted so that they align with the holes in the case. To do this, first fit all of the l.e.d.s into their holes in the p.c.b. leaving them unsoldered and with full length leads.
Fit the board in position. taking care to line up the switches in their slots, and push the l.e.d.s into place in their holes in the case. Ensure that all of the l.e.d.s are the correct way round by checking that the short lead which indicates the cathode ( $k$ ) is


## Fig. 2. Internal circuit details for the LM3914N i.c.

## RANGE <br> SWITCHING

The three position slide switch $\mathbf{S} 2$ selects the range by introducing the necessary divider components. Resistors R1, R3, and R4 on the 5 V range. $\mathrm{R} 1, \mathrm{R} 2$, and R 4 on the 10 V range. and series d.c. blocking capacitor Cl along with RI and R4 on the a.c. range.

Changing the values of R1, R2, and R3 will allow different ranges to be covered. For example making R1 from two 10 M resistors in series will give $0-250 \mathrm{~V}$ d.c. High voltages such as these will not be applied directly to the i.c. because of the potential divider and so are quite acceptable.
The resistor values are selected to divide by four for the 5 V range and eight for the 10 V range as follows:

$$
\begin{array}{rl}
5 V-1.25 V & R 4 /(R 1+R 3+R 4) \\
= & 100 / 270+30+100) \\
= & 100 / 400=1 / 4
\end{array}
$$

$10 \mathrm{~V}-1.25 \mathrm{~V} \quad \mathrm{R} 4 /(\mathrm{R} 1+\mathrm{R} 2+\mathrm{R} 4)$
$=100 / 270+430+100)$
$=100 / 800=1 / 8$
only the peak value as it moves so rapidly between the intermediate values. These types of displays will also be observed when using the d.c. ranges if the voltage is varying. A good illustration of this can be obtained by using the probe to observe the capacitor voltage in an astable 555 i.c. oscillator.

## CONSTRUCTION

A special probe case is used to house the circuit. This is available pre-drilled for the l.e.d.s and slide switches. The circuit is built on a printed circuit board. The foil track pattern and the component layout is shown in Fig 3. The resistors and capacitors should be fitted first. followed by DI and a socket for the i.c. Capacitor C2 and D1 must be fitted the right way round as shown in Fig 3.
Resistor R1 is fitted beneath S 2 and so double check that it is the correct value before fitting the switch over it. Alternatively RI can be fitted to the track side of

The completed Analogic Test Probe showing lettering of the l.e.d.s.



The completed circuit board with the slide switches soldered in position. Resistor R1 is mounted track side - see photo below.
nearer to the i.c.. and solder them into position one by one. ensuring that they are pushed evenly down into the case holes. The board can now be removed and the connecting leads added.
The leads from the battery clip, the probe tip. and the external ground and power leads are connected to rectangular track areas on the rear of the board. Use thin extra-flexible test lead wire for the ground and power leads. Do not fit the clips to the other ends of these leads yet. as first they must be threaded through their hole in the case, after the circuit has been tested.

## TESTING

For testing it is better to apply power via the external power leads than from the battery clip because the series diode and resistor will prevent excess current if there is a fault. Insert the i.c. being careful to get it the right way round, set the switches to "external power" and 10 V d.c., and apply 9 V via the external power leads. There should be no smoke and all l.e.d.s should be out.

Next connect the probe input lead to the positive supply and check that l.e.d.s D8. D9. or D10 light. Switch to the 5V d.c. range and only l.e.d. D10 should be lit. On the a.c. range the l.e.d.s should light briefly as the input is connected and then go out.

If all is well, try to find other voltages and see that the readings correspond with the range selected. If nothing works check for dry joints, reversed power leads and components, and p.c.b. track bridges and cracks. Check the voltages around ICI, in particular. the power supply to pin 3 and the reference voltage on pins 6 and 7 .

A wrong value resistor for R3 could give very low l.e.d. current, and any wrong values at the probe potential divider will give incorrect ranges. There can be few faults with such a simple circuit and hopefully most units will work from scratch.

## FINAL ASSEMBLY

A long M3 screw which is filed to a point and fitted with sleeving over most of its length forms the probe tip. This is fitted to the front panel of the case with one nut and a shakeproof washer. Its connection to the board is via a solder tag and a short length of flexible wire.

The board is held in position by means of the switch bodies which are glued to the case on two sides. Use a flexible glue such as Evo-Stik which will allow the board to be removed in the future if repairs are needed. Apply glue to both surfaces and allow 15 minutes for it to go tacky. The board will then be held firmly immediately it is pressed into position. Alternatively
double-sided sticky pads are suitable but do not give quite as rigid a result.
The case specified has interlocking top and bottom sections that take a certain amount of knack to separate. The plastic material used is very tough however and so should resist considerable force. To retain the battery an internal divider can be made and glued in and the battery retained with foam plastic. Alternatively a sticky pad can be used to hold the battery directly to the case.

## BATTERYLIFE

The circuit consumes approximately 4 mA when the lie.d.s are out. so battery drain should be very small. Take care though to switch off after use as even this low current will flatten a PP3 in a less than a week. Note that either the Off or External Power positions of the switch will disconnect the battery

## USE

There is little that can be said about such a simple and versatile instrument except that it will become indispensable and get regular use. The probe tip is able to withstand overloads in excess of 250 V a.c. on all ranges and the power input is safe up to 25 V . Remember that two connections are required to get sensible voltage readings and so the Ground lead must always be connected to the negutive point in the circuit under test.

The external power lead can be removed if its use is unlikely. When external power is connected always connect the Ground lead first. Accuracy is assured by the use of one per cent potential divider resistors. The i.c. has similar accuracy, and so the readings are as precise as the display allows.

## CUSTOMISING

Individual ranges can be tailored for specific uses by altering the voltage divider resistors as discussed earlier, and extra range switching could be incorporated by adding a small toggle switch and additional resistors. In this way the probe can be set up for unskilled users to make simple checks on equipment in the field or in factory test departments.




THIS is the first part of a six-part series concerned with the GCSE Electronics Practical Assignment. By making clear what is expected of you and in helping you to plan your time and pace the work, you are more likely to obtain the best mark for your ability. This month, I shall give some GCSE background information and discuss how the practical project will be assessed.
Although aimed at mainstream GCSE Electronics, much of this information would be helpful for students making projects for GCSE (mature), GCSE Technology and for certain non-GCSE based examinations.

## Background Information

The examining board offering GCSE Electronics are, MEG (Midland Examining Group), LEAG (London East Anglian Group), SEG (Southern Examining Group), NEA (Northern Examining Association) and the Welsh Joint Education Committee. The system north of the border is different but the Scottish Examining Board operates a system of short courses one of which Electronics Design Project - involves similar processes. Students entering for this would benefit from some of the information given here.
As well as the written examination(s) sat in the summer of the second year of study, GCSE boards require students to construct a practical project or projects. In most cases you have a free choice of topic.
Precise requirements and the percentage of final mark available for the project varies from one board to another. It may also depend on whether or not you are attempting the extension paper. This information is set out in more detail later.

## Extension paper

The extension paper is a more difficult written paper aimed at the most able students. Doing this enables you to obtain a grade A or B and normally you cannot obtain these grades without having sat it. Many schools use the results of a mock
examination to determine whether or not you will be allowed to attempt the extension paper. You will have to trust the judgement of your teacher or lecturer the school or college cannot enter candidates for the extension paper who have no chance of obtaining grade $A, B$ or $C$.
If you are entered for the extension paper and make an unexpectedly poor effort, you will not end up with a grade lower than you would have obtained had you not attempted it. Also, under exceptional circumstances, a candidate not entered for the extension paper may occasionally obtain a grade B with some boards. Note that the syllabus number often depends on whether the extension paper is included or not.

## Detailed requirements

Note: the following requirements apply to candidates entering as full-time students at a school or college. Part-time and privately-entered students should consult the board as should students taking GCSE Electronics as a one-year course. This is because there may be differences in detail.

## Midlands Examining Group (MEG) Syllabus 1751

Candidates must attempt either one long or up to three short practical assignment(s). Although there is a free choice of topic, the board publishes a list of suggestions. The total time spent on the project should be in the region of 20 hours and will be worth up to 40 per cent of your final mark.
You must complete a specification sheet and have it approved by your teacher before work commences. You will be asked to complete all work and hand in your written report sometime in April.

## Southern Examining Group (SEG) Syllabus 1060 or 1061 (with extension paper)

Candidates must undertake one major project. A free choice is allowed but the board requires a project brief to be
approved before starting development work. This must be completed during late February.

The project is worth up to 25 per cent of the final mark and will involve, typically. 24 hours of work. You will be asked to hand in all work, including at least one photograph of the completed artefact, sometime in mid-April. The writfen report should be 1000 to 3000 words in length.

## London East Anglian Group (LEAG) 1505 or 1506 (with extension paper)

In January, schools are sent six outline project briefs in booklet form and each candidate will receive a copy. The school offers at least four and from these you choose one. If you are attempting the extension paper, the project is worth 15 per cent but if not, it is worth 20 per cent. You are expected to spend about 12 hours of laboratory time on your project and will be asked to hand in all work towards the end of April.

## Northern Examining Association (NEA)

One major project is required and this is expected to occupy approximately 30 hours of laboratory time. Although a free choice is allowed, an outline form must be completed and submitted to the board. You will be asked to hand in all your work, including the written report (a maximum of 2000 words is suggested) and a photograph of the completed device. sometime in May. The project is worth up to 35 per cent of the total marks.

## Welsh Joint Education <br> Committee

One major project is required and although a free choice is allowed, approval is needed by the board before work commences. The project is worth up to 30 per cent of the final mark.

## Assessment

When students learn that it is their own teacher who does an initial assessment of their practical work they sometimes think that this degrades the status of the exam. Some mistakenly believe that their teacher will give them higher (or lower!) marks than they deserve. Think though who is better at knowing you as an individual and how much (or how little I) of your own effort has gone into the work you produce?
As regards giving marks which are too high - this is not possible because your teacher follows a standard set of criteria supplied by the board - not his or her own. Also, the work is subject to moderation by the examining board. This means that an assessor, or moderator, appointed by the board, may visit your centre, talk to your teacher and look at candidates' work. The assessor may not visit the centre but instead ask for the work of certain candidates to be sent by post.
The moderator sees the work of several schools and can apply an adjustment where necessary so that the mark you obtain is
fair and reflects the quality of the work not only in your own centre but nationwide. Assessors meet, mark the same pieces of work independently and compare results. The board can then be sure that they are applying uniform standards. You may be assured that assessment is closely controlled and fair.

## Get organised!

When you are told that you may begin preliminary work on your project, there are a few things to get clear in your mind. Firstly, find out which examination board is setting your exam. Remember, this will not necessarily be the same as the Physics or Technology one.
Find out the syllabus number and whether you will be entered for the extension paper or not. Find out what approximate grade you would have achieved based on any mock exam mark you have obtained. Ask your teacher and yourself - what grade you should reasonably aim for and go for it.
Find out the deadlines you will be working to. You teacher will tell you the date on which any outline brief is required as well as that for the finished project and written report. This really is a deadline and schools will not accept material arriving late. Your teacher has quite a lot of administration to do before sending the paperwork off to the board.
'Enter in your diary your own deadline which will take account of little things going wrong. One or two weeks before the true deadline will be sufficient. Find out what proportion of lesson time will be devoted to pursing the project. Some teachers allow only some of the lessons to be used, others allow all available time. Find out whether you will be able to work in lunchtimes.
Generally, you cannot take your project
home to work there because your teacher must certify that it is all your own work and must therefore keep you under supervision. Remember, where a certain number of hours laboratory work are suggested, these are only guidelines. You may spend longer if you wish.

At first the deadline will seem far away and many students only really start to work when the end is in sight and they fee! the pressure. These students are unlikely to obtain a good mark - I have seen this time and time again. Those who start planning straight away do well. Remember best project marks do not necessarily go to candidates who produce the most elaborate device. It is quite usual for some of the best marks to be awarded for very modest - even non-working - projects. You must treat your own situation in the most advantageous way. All this will be discussed in detail in later articles.

## Who pays for the components?

Schools often allow development work to be carried out with their components. This saves you money especially where a few components are destroyed by what may be called "forgivable experimentation". Some schools insist on payment for components which you ruin when you should have known better operating an l.e.d. without a series resistor to limit the current, for example.
You will be asked to pay for any components taken from school stock and used in building the final version of your circuit. This is especially true if you intend keeping it after it has been assessed (although it may not be released by the examining board until September). Different schools have different policies so find out now what costs you will be responsible for.
You may be able to buy components
cheaply through the school as a result of bulk buying. Alternatively, you may be able to form your own purchasing group with a few other students and so save packing and postage costs. Some mail order suppliers give a discount where several of the same item are ordered together.

## Being realistic

I cannot stress enough the importance of having a realistic view of your ability in relation to the complexity of the circuit you are wishing to build. At this time of year, the EE editor's office receives a number of letters from students asking for assistance. Some of these pass through my hands. Frankly, I am dismayed by the complexity of many of the proposed circuits. It is usually clear that the candidate has only the basic knowledge and average ability. Embarking on such a project would certainly end in disaster.
Some candidates wish to use radio links which, apart from being illegal, would be very difficult to construct even by an expert. These complicated circuits are not required for GCSE. I hate to dampen enthusiasm but there would be very little of this left in March or April anyway! Your teacher's advice is, of paramount importance. If you are not sure of the standard required, ask the examination board - they will probably oblige with a list of examples.

## Final point

GCSE examinations are based on positive achievement. This means that you gain marks for what you do well and do not lose them for things which you do badly. In other words, you only stand to win!
That's all for this month. Next time we shall look in detail at the steps you should take in preparing to build your chosen circuit. We shall also look at the development work needed for a fictitious project.

# Hiteinonics SPECIAL PUBLICATIONS 

## TEACH-IN No. 3 EXPLORING ELECTRONICS f2.45 plus pge By Owen Bishop

Designed to explain the workings of electronic components and circuits by involving the reader in experimenting with them. The book does not contain masses of theory or formulae but straightforward explanations and circuits to build and experiment with.

The text is split into 28 easily digestible sections, each with a separate project. The breadboard experiments assume no previous knowledge, start at semiconductor diodes and progress through bistables, timers, amplifiers, binary etc up to f.e.t.s and shift registers.

The projects include radio receivers, various timers and alarms, plus temperature sensors and water detectors etc.
An excellent source book for GCSE courses.

# Feature <br> MILITARY ELECTRONICS 



## PETE CHOWN

MILITARY electronics suddenly became headline news at the start of the Falklands war when HMS Sheffield's Seawolf defence system failed with catastrophic The problem was a classic one of technology being placed in an unfamiliar situation. Because the Exocet was made in France, the system decided that it was friendly (it was designed to defend shipping from an Eastern Block attack) and no attempt was made to shoot it down.
In this article, we can only consider a small part of the vast subject of military electronics; we will be emphasising missiles and guidance systems, battlefield simulations, radar and communications. Finally there is a brief look at what the future of warfare may be like.

## Missiles and Guidance Systems

Despite the huge variety of different missiles, guidance systems follow only a few basic patterns.

Anti-aircraft or anti-missile missiles generally are either heat seeking or radar guided. Heat seeking missiles carry an infra-red camera, which allows them to detect the exhaust plumes from aircraft or missiles. It is much better to detect infra-red than visible light, because there is a large amount of infra-red coming from the exhaust plumes, whereas for visible light it would be necessary to use reflected light. The missiles would then be ineffective at night and more sophisticated electronics would be needed to analyse the pictures coming in.

Radar guided missiles can be used for any target that can easily be distinguished from its background by reference only to its silhouette. They are thus used for anti-ship missiles as well as for anti-missile or aircraft missiles. There are several different approaches here - most of these missiles only have a radar receiver and not a transmitter because this would mean carrying a large power source. A helicopter or ground-based transmitter is therefore used, commonly from the same equipment that fired the missile.

Certain missiles fired from one aircraft to another need the firing aircraft to remain in position until the missile strikes, to provide a radar beam, this tends to make the attacker vulnerable. Consequently most of these missiles are heat-seeking.

## Cruise

There is another type of radar guided missile which is quite different. The cruise missiles (not necessarily nuclear; the term refers to any long range missile that flies fairly close to the ground) are typically guided by a radar system that takes readings from the ground along the path down which it is flying, and compares them with a map of the terrain which it knows it will be flying over. This then tells it which route it is currently on and so how it should change that to stay on course.

Whether an incoming missile is heat or radar guided determines to a large extent what the appropriate defence response should be. Certain methods are obviously effective for both, but the two main responses of aircraft under attack are each only effective against one type. Aircraft fire flares which produce localised intense heat to confuse infra-red guided missiles, and discharge chaff (small aluminium strips) to confuse radar guided missiles.

## Sea Skimming

Sea skimming missiles are becoming increasingly important in battles at sea. They represent a radical departure form the conventional anti-ship missile, in that they fly only a couple of metres above the sea. This makes them much harder to shoot down, because they are reflected in the water, in various different distortions, and so any radar tracking them must be able to distinguish the real missile from its images.

Interestingly, the problem of the sea skimmers seems to have been largely solved; despite many naval conflicts since the Falklands the only ship to have been damaged by one is the USS Stark, which was hit in the Gulf. Even here the damage was only possible because the ship was not in a situation of all-out war, and so the firing aircraft was allowed to get fairly close before any action was taken.

The Europeans and Americans have defended their ships against these missiles in different ways. The Americans designed a rapid-fire gun, which discharges a large cloud of shells, practically guaranteeing a hit. The Europeans have defend their ships mainly with missiles, such as the Seawolf.

Shells are cheaper and likely to be more reliable, being less complex, however the major disadvantage is that a large amount of ammunition needs to be stored, since a large number of shells need to be fired at each target.

## Air-to-Surface

The final kind of missile is the air-to-surface missile, which is used when a very specific target must be destroyed. The large expense associated with this kind of weapon means that bombs tend to be used instead wherever possible. Air-to-surface missiles tend to be laser-sighted - they are aimed by someone on the ground pointing a laser at the target. The missile then homes in on the laser spot.
The reason for adopting this method is that targets on the ground are much more difficult to pick out; a target might be a building, one vehicle among many and so on. Only a human can therefore know what is a valuable target and what is entirely irrelevant.
A variant on these laser-sighted missiles is the laser-sighted bomb. Generally these are not guided in their own right, although a few are, but instead electronics in the aircraft detects the laser spot, and calculates when the pilot must release the bomb in order for it to hit the target.

The reason that guided bombs are sometimes used is that passive bombs are not always reliable, because strong winds can spoil the calculations, and cause the bomb to fall slightly off target. If a bomb only misses a tank slightly, it will of course have no effect, a direct hit is needed to penetrate the armour.

## Battlefield Simulations

In one sense, the military have a unique training problem; if you learn by doing, you are quite likely to get shot.

The answer to this has been to create battlefield simulators, which can teach people basic combat skills without the element of make-believe which has traditionally been present. One of the most important roles of simulators is to teach people to use weapons which are too expensive to fire very often outside war.

One weapon which falls into this category is the MILAN
anti-tank rocket. It is a missile which is guided by moving the launcher from side to side once it has been fired. This causes the missile to move similarly, so that it can be guided onto a moving vehicle, or so that poor aim can be compensated for.

Because one shot cost $£ 20,000$, those who might have to use it in wartime would only have fired it once or twice (this missile would, of course, be very widely used in war, so many people have to be trained to use it). In this case the simulator has been fitted into the original size of the launcher, it is therefore very realistic to use. The small size also allows it to be "fired" while crouching in a ditch, it can thus be used during exercises.

The simulator works by projecting, onto a piece of glass in the sights, a small light which represents the exhaust plume of the missile, this dies away as it gets further into the distance. Eventually the light "flares up" to symbolise the explosion.

The spot of light can be guided onto any target, for example a car on a nearby road, or alternatively the system can provide a symbol as a target, whose speed and direction can be varied by the instructor.

## Tanks and Infantry

Other simulators allow tanks to be simulated - both in being shot at and in firing; or indeed both at the same time in the case of large exercises. Generally these work by looking at the gun's positioning and calculating if it will hit its target. If it will then a smoke canister is set off, otherwise the information is stored and it can be examined later to see what went wrong and why.

Yet another variant is a system which simulates infantry fire. Here individuals' rifles are replaced by simulated weapons which detect whether they have hit their target, if they have a buzzer is set off, this can only be silenced by the victim lying on his back. Meanwhile, the umpires assess what is happening and can also reactivate "dead" people

This system makes the exercises much more realistic than conventional ones in which blanks are used, since it allows a much more realistic representation of a battlefield.

## Radar

Radar has also developed a great deal in the last few years. Gone are the days when someone has to sit watching a screen showing a large amount of clutter (i.e. stationary objects that reflect the radio waves) and estimate the type or attacker from the size of the moving "blip"

Modern radars can to a large extent identify the attacker by computer and so show on a screen the identity of the "blip". They can also scan more quickly, because the single rotating radar dish has been replaced by multi-element radars which can look at a whole range of points at once

Radar has always been, however, rather a mixed blessing. Passive devices have been developed which can detect intruders by their radar transmissions, and not send out a beam of their own which might be intercepted.

Most combat aircraft now incorporate a system which detects when a radar beam is being used to lock a missile onto them and this sets off an alarm to the pilot. Evasive action can then be taken, very often this will cause the missile launcher to lose track of the plane, even before a missile is fired.

## Jamming

Radar jamming is also becoming increasingly effective. It is this technology which it is hoped will make the Stealth bomber invisible to enemy radar - the bombers first approach low down, out of radar view, and then fly up and jam the radar with a high-intensity beam of their own.

One of the problems encountered, however, is that it is also necessary to make the shape of the aeroplane such that it does not produce a large image on the radar; this is not compatible with the other objective of producing a fast plane which can turn reasonably quickly. The speed requirement is further hampered by the fact that the exhausts have to be prevented from showing up too much to avoid heat seeking missiles. Since several of the early bombers crashed, it would seem that they have not been an overwhelming success. Stealth bombers were used in Panama, they were invisible to the radar and did not crash, however they bombed a field behind the targeted barracks. The US military claimed this was deliberate and aimed at stunning the occupants!

## Communications

Communications are very much the weak link in modern war fare; it is not so much the ability to achieve the various communications links as the need to make them secure and safe from jamming. An example illustrates the point:

A few years ago it was decided to develop a system which would fly, what was essentially, a radio controlled model plane over a battlefield and use it to aim a laser at enemy tanks. This would enable allied tanks to fire copperhead guided shells which would home in on the laser spot.

This system could obviously be very beneficial on a modern battlefield, however, the system was never produced because the secure microwave link that was to run between the plane and its ground-based controllers proved to be impossible to develop. Apart from this link the project was comparatively simple, and could have been achieved without much trouble

Communications back to the military headquarters are also very important in any campaign. These links are normally carried by satellite. In this way the signal can be sent up nearly vertically and there is less chance of the enemy being able to jam it.

Enemy interception is not very important because modern codes can normally resist even the most determined assault. The only problem that there might be with interception is if information about the code was passed to the enemy.

## The Future

There are two possible ways for warfare - one is increasing sophistication, the other is simply increasing firepower without really increasing accuracy. Both have been seen very recently. The two major examples of increasing firepower are the A10 fighter and a new rocket launcher being used by the British Army in Germany.

The A10 is more in line with what was accepted as the norm during the second world war in terms of electronics, it has very little in the way of sophisticated navigation, radar or missiles. it does, however, carry laser targeted bombs. The A10's main weapon is a large "gatling gun" which fires around 50 shells per
Sea Eagle missiles on a Buccaneer aircraft. Sea Eagle is in service on Royal Air Force Buccaneers and Royal Navy Sea Harriers. It is a sophisticated long-range fire-and-forget missle designed for use against modern warships up to the largest size and equipped with the latest countermeasures and air defences. British Aerospace (Dynamics) Lid

second. The idea is that the A10 can destroy a large number of tanks using the gun (each one of the shells can destroy a tank!) while being fairly invulnerable itself, due to a heavily armoured area which protects the pilot and important instruments.

The new rocket launcher is also an anti-tank weapon. It has a firepower equivalent to a regiment of artillery, although individual shots are less accurately targeted, and it cannot keep up the rate of fire for long. What it does is to fire a rocket which cruises to the target area and then breaks up into a cloud of tiny bomblets. This cloud can cover around two acres, destroying most if not all of the equipment there.

## Sophistication

The alternative way forward is increasing sophistication, so that although less shots are fired many more of them strike home. The most promising development is also on anti-tank weapon, it will carry out a similar role to the rocket launcher. It comprises a rocket which is fired from some way behind the front line, and given a preprogrammed destination. It then breaks up into a large number of small shots which parachute down. When these shots pass over a tank, they explode, firing a cloud of hot gas and metal vapour downwards with sufficient speed to cut through the armour on the tank.

Another of these sophisticated weapons is a conventionally armed cruise missile. This is given a tactical target such as an airfield in enemy territory, which it flies to and then breaks up into a cloud of smaller bombs which can then completely destroy a relatively small target such as an airfield.

As far as winning a future war goes, it seems that the increased firepower weapons may be the best. The experience of the unproven Seaw olf missiles in the Falklands suggests that sophisticated weapons do not necessarily perform correctly the first time they are sent into a combat situation. Unfortunately, increased firepower could mean increased civilian deaths; the sophisticated weapons hold the promise of being able to destroy military targets and nothing else.


The extensive damage caused to HMS Devonshire following a direct hit by the Sea Eagle missile. In operational circumstances this would have resulted in complete disable. ment for the ship. (British Aerospace (Dynamics) Ltd

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# SPATAL POWER DISPLAY 

## GARY CALLAND

A novel and attractive way of monitoring the output from a stereo system. Provides four separate l.e.d. displays.

UNLIKE its name suggests, the Spatial Power Display, or SPoD for short, is an attractive and novel way in which to monitor the power output from a stereo. Although primarily designed for car stereo systems, the SPoD can be used with other stereos, providing a suitable power supply exists.
The SPoD provides four separate visual displays representing the output from each of the four car stereo loudspeakers. Coloured I.e.d. s are arranged so that their positions represent the probable path of a sound wave: the greater the power, the wider reaching the sound wave is likely to be. In addition, as the power output increases to a maximum, green, yellow and finally red l.e.d's are progressively illuminated to provide additional power output information.

The SPoD can be easily fitted to any stereo system, regardless of power output and loudspeaker impedance: a simple preset control on the circuit board can be adjusted to suit any power indication requirements. Also, since the SPoD is small and compact, it can be discretely mounted virtually anywhere.

Not only can the SPoD be used to help balance loudspeakers, or to give indication of a faulty amplifier output, it provides rather a pleasing visual accompaniment to in-car music, perhaps even helping to ease the frustrations of those all too frequent hold-ups and traffic jams.

## HOWIT WORKS

The basic principle of operation is shown in Fig.!. Firstly, to simplify the explanation, only a single loudspeaker system will be considered. The a.c. voltage waveform developed across the loudspeaker is converted by a rectifier stage into a d.c. voltage, the magnitude of which is proportional to the power output of the loudspeaker.

This d.c. voltage is fed to a set of voltage level comparators. These change their output state from a logic low to a logic high when their input signals exceed a preset reference level.

Each comparator has a higher reference level than the one below it, hence an increase in loudspeaker power increases the number of comparators which have a high
output. The output from the comparators simply switch on or off their associated 1.e.d's.

## MULT/-SPEAKER SYSTEM

To increase the system to accommodate four loudspeaker inputs, the single system described above could simply be repeated. However, this is undesirable due to the cost of components and in the large amount of board space required. Instead, Time Divi-


Fig. 1. Spatial Power Display block diagram and waveforms (below) showing the basic principle of operation



Fig. 2. Block diagram for the "cost saving"time division multiplexing system.
sion Multiplexing, as shown in Fig.2.. is used to decrease costs and space.
Again, as before, the a.c. waveform from the loudspeaker is converted into a d.c. level. However, each of the four d.c. voltages is fed to a single set of tevel comparators, one after another, rather than each being fed to their own set. Likewise. the output from the comparators only activates one of the four l.e.d. display sets. one after another.
Hence in this way, all four inputs are continuously, sequentially sampled to produce a visual output on their associated l.e.d display. This scanning is performed at such a speed that it cannot be delected by the human eye and so the illusion of four permanently activated l.e.d displays is given.
As an idea of the cost and space saving
advantages of this technique, the SPoD only uses six i.c.'s whereas a comparable unit using four identical separate systems would use nearly double that number.

## CIFCUIT DESCRIPTION

The complete circuit diagram for the Spatial Power Display is shown in Fig. 3.
The voltage waveform presented to each set of loudspeaker terminals is fed to a diode rectifier bridge via d.c. blocking capacitors Cl to C8. These are necessary since some car stereos use a bridge amplifier output stage, and the nominal voltage applied to each of the two loudspeaker terminals is about 5-6 volts referenced to ground.

Individual diodes als opposed to a packaged diode bridge were used to reduce space and costs. Each rectified waveform is smoothed by capacitors C9 to C12 and resistors RI to R4 to produce four d.c. voltages.

These are fed into the four inputs of the analogue switches contained in IC3. Only one switch is closed at any one time. depending upon the logic state of the control inputs. Hence, only one of the four d.c. voltages is allowed to pass onto the next stage.
The selected voltage from IC3 is buffered by IC4c and fed into each of the noninverting inputs of the five comparators formed by IC4b and IC5. Zener diode DI8. resistor R6 and preset VRI produce a constant reference voltage in the range 0 to 10 volts. This is buffered by IC4a and potted down into five voltage levels by the resistor network formed by R8 to R12.
These five voltages are fed into the inverting inputs of the comparators to provide the five different reference levels. Varying VRI, and therefore the reference voltages. allows different stereo amplifier outputs to be catered for.
Each comparator output is buffered by IC6 to produce a common l.e.d. drive to each l.e.d. display, via current limiting resistors R17 to R21. Only one l.e.d. display set is activated at any one time since only one of the transistors TRI-TR4 is switched on at any one time.
The transistor control signals are the same signals that close IC 3 's analogue switches. and are obtained from four alternative outputs from the eight available outputs of the octal counter IC? This is clocked by the oscillator formed by ICla, IC1b. capacitor C13 and resistor R5.
All unused logic gate inputs are tied to ground and diode D1 ensures that incor-





ALLLEST

Fig. 4. Printed circuit board component lavout and full size copper foil master patern. The method of mounting the diodes and l.e.d.s is shown above.


Fig. 5. Full size l.e.d. display drilling template for the specified plastic case.
rect connection of the power leads (the wrong way round) does not result in a useless set of smouldering i.c. 's!

## CONSTRUCTION

Construction of the SPoD is straight forward as all the components fit quite tightly on a standard sized p.c.b. The component layout and full size copper foil master pattern is shown in Fig. 4. This board is available from the EE PCB Serrice, code EE714.
It is best to start construction of the p.c.b. with the five wire links and diodes D2 to D17. as these would be difficult to fit if other nearby components were in position. Note the orientation of the diodes.
Next, the resistors and capacitors may be fitted, followed by the diodes DI and D18. Capacitors Cl to C 8 are quite densely packed and so it is best to fit the innermost capacitors first, then work outwards. I.C. sockets are recommended to reduce i.c. damage, and they should be fitted next. together with the four transistors.
Finally, the l.e.d.'s should be mounted, all at the same height above the p.c.b. This is ensured by fitting a 1 cm long piece of plastic sleeving over each l.e.d. lead. Also, the orientation of the l.e.d. is important; the long lead (the anode) of the l.e.d. should always be inserted in the inner most of the two l.e.d. holes.
A careful inspection of the copper track side of the p.c.b. to check for solder bridges is a good idea before the i.c.'s are inserted
into their sockets: the usual CMOS handling precautions should, of course, be observed. Short lengths (about 10 cm ) of power supply and loudspeaker cable can then be soldered into their associated p.c.b. pads. The p.c.b. is then ready for mounting in its case.
The p.c.b. is mounted on the inside of the case lid by four 6BA bolts. The cables are then connected to a 10 -way strip of terminal block mounted on the other side of the case lid. The use of terminal block allows ease of connection to the power supply and amplifier loudspeaker outputs.
Drilling the holes for the I.e.d.'s, in the case body is quite difficult and is best done
1.e.d.s. Appropriate lettering will finish the unit off nicely.

## POWEA <br> INDICATION CALCULATION

The SPoD relies on the formula:

## Power $=V^{2} / \mathrm{R}$

to indicate the peak power output of a loudspeaker where $\mathrm{V}=$ voltage across the speaker terminals, and $R=$ loudspeaker impedance ( 4 ohms for most car speakers).
The following table shows at what approximate power levels each l.e.d. will illuminate for various reference voltage settings.

| Reference Level | 10 V | 9 V | 8 V | 7 V | 6 V | 5 V |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 25 W | 20 W | 16 W | 12 W | 9 W |
| Red | 6 W | 5 W | 4 W | 3 W | 2 W | 6 W |
| YelloW | 1.8 W | 1.5 W | 1.2 W | 0.9 W | 0.6 W | 0.5 W |
| Green | 0.4 W | 0.3 W | 0.25 W | 0.18 W | 0.14 W | 0.09 W |
| Green | 0.09 W | 0.07 W | 0.06 W | 0.04 W | 0.03 W | 0.02 W |
| Green |  |  |  |  |  |  |

using the template, Fig.5. (or a photostat of it). This is stuck onto the case top with sticky tape and a small drill (about 1 mm diameter) is used to produce a hole at each corner of the l.e.d. cutout.
Once this is complete, the template can be removed, and the complete l.e.d. cutout can be made using the small corner holes for guidance. This method will ensure a neat, balanced distribution of l.e.d. holes which lie directly above their associated

## TESTINGAND INSTALLATION

With the unit powered, turning the reference level control preset VR1 to give 0 volts out should result in all the l.e.d.s being illuminated. If this is so, all is well. All that remains is to connect the SPoD to your stereo system, and to adjust the level control to the required power level setting.

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WANTED details on Corvus systems 5MB


#### Abstract

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## ULTRASONIC CAR ALARM



Complete trit including case 44.3678KL ............... \& 30.40

This system is specially designed to protect your car and its contents against potential thiefs. Low current consumption and high noise immunity are just two of its distinguishing features.

In addition the system has a voltage sensing device i.e. the alarm is also triggered if appliances are switched on by an unauthorised person (e.g. the interior lighting when the door is opened).

PC Radio (Eketor Elextronics febury 1890)


## DIGITAL PROFESSIONAL ECHO 1000

(Etektor Electronics June 89)

This low cost echo unit is certain to impress music lovers - amateur and professional - everywhere. Excellent specification and top performance make the EU 1000 a winner and despite meeting professional requirements the unit will not make too big a hole in your pocket.
Working on the delta modulation prin-
ciple on a digital base, delay times up to one second are possible at full bandwidth and large signal to noise patio.

Complete hat
4.255 BKL

C 99.50
Ready assembled module
$44.255 F$.
.................... \& 134.50

## Specification

Input sensitivity:
Input 1: 2 mV Input 2: 200 mV
Dealy Time:
variable from 60 ms to 1 s
Bandwidth:
100 Hz to 12 kHz
Additional features:

- inputs mixable
- single and multiple echo
- adjustable delay level
- switchable vibrator
- switch-controlled noise suppression

This FM radio consists of an insertion card for IBM PC-XTs, ATs and compatibles and is available as a kit or a ready-built and aligned unit. The radio has an on-board AF power amplifier for driving a loudspeaker or a headphone set, and is powered by the computer. A menu-driven program is supplied to control the radio settings.

## Complete kit

44.5448 KL Ready assembled modula
44.544BKL
\& 82.75
4.544F
.................. \& \& 137.30

VM 1000 Video-Modulator

(Elentor Electronics March 90)

Many inexpensive or older TV sets lack a SCART or other composite video input, and can only be connected to a video recorder or other equipment via an RF modulator. The modulator operates at a UHF N channel between 30 and 40 . Use is made of a single-chip RF modulator that couples low cost to excellent sound and picture quality.

Complete kit
44.546BKL

$$
\text { \& } \quad 36.90
$$

Ordering and payment:

- all prices excluding V.A.T. (french customers adod $18.6 \%$ T.V.A.)
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- we deliver wortowide except USA and Canada
- dealer Inquiries weicome


## RFK 7000

## RGB-CVBS Converter

(Elextor Electronics October 89)
Nearty all computers supply as an output signal for colour monitors RG8 signals. With the help of the RFK 7000 it is possible to record this signals with a videorecorder or to give them onto a colour TV (This is only possible, it the

## FRK 7000

## CVBS-RGB Converter

Withs the help of the FRK 7000 e.g. it is possible to use a cheap clour monitor with RGB input on a video recorder. The voltage supply is gained from a $12 \mathrm{~V} / 300 \mathrm{~mA}$-DC voltage mains adaplor.
computer delivers a vertical sync. of 50 Hz and a horizontal sync. of 15.625 Hz ).
The vollage supply is gained from a $12 \mathrm{~V} / 300 \mathrm{~mA}-\mathrm{DC}$ voltage mains adaptor.
Complate kit
44.525BKL ..
$\varepsilon$
66.50

Ready assembled module
$44.525 F$.
I
119.50

Complete kit
44.509BKL

Ready assembled module
4.50ヶf.....................


LPS 8000 / LC 7000 Low Cost Show Laser
(Electronics The Maplin Magarin Dec $88+$ Feb-Mar $\mathbf{C O}$ )

An almost infinite number of circular pattems can be projected onto a wall or ceiling with this super laser show equipment.
The complete project includes a laser tube and accompanying power supply, housed in a metal case, and a laser controller, LC 7000 . The laser controller drives the accompanying deflection unit, fixed onto the laser power supply case, which produces the numerous configurations.
Naturally the laser tube, together with the power supply, can produce beams withou the laser controller and the controller can be used with other, similar lasers.

|  |  |  |
| :---: | :---: | :---: |
| LPS 8000 Laser Power Supply, complete kil Version 240 Volts AC |  |  |
| $44.4288 \mathrm{KLL220}$ | $\varepsilon$ | 86.90 |
| Version 220 Volts AC |  |  |
| $4.4288 \times 1240$. | $\varepsilon$ | 86.90 |

LC 7000 Laser Controller, complete kit
Version 12 Volis DC
$4.4278 \times L$
............... \&
E
60.80

H-N Laser Tube 2 mW
4.428LR

Tubl...........
E
60.80

LPS 8000 Laser Power Supply, ready assembled module Version 240 Volts AC 44.428F240............. f
156.50

Version 220 Volts AC
44.4" 8 F220 $\qquad$ E

Version 12 Volts DC
44.427F.................... \&
$\Sigma$
104.30

Laser Motor-Mirror Set, complete kit
44.506M $\qquad$ .

VIDEO RECORDING
AMPLIFIER
(Elethor Electronics Aprill 89)
Losses can easily occur when copying video tapes resulting in a distinct reduction in quality. By using this video recording amplifier, with no less than four (!) outputs, the modulation range is enlarged and the contrast range of the copy increases.
Two level controllers for edge definition (contour) and amplification (contrast range) allow individual and precise adaplation.


Complets Kit
(including Box, PCB and all parts
44.324BKL ................ \& 14.75

## IBM PC Service Card

This card was developed for assistance in the field of service, development and lest. The card is used as a bus-extension to reach the measurement points very easy. It is also possible to change cards without having a "hanging computer.


Complete kit 44.517BKL \& $\quad 77.95$

Ready as sembled module 44.517F 137.95

## TA 1000 Telephone Answering Unit

(Elettor Electronics January 1890)

This automatical telephone answering unit uses a $256-\mathrm{kbit}$ voice recording circuit to store and replay your spoken message of uo to 15 seconds. Noteworthy features are that it is available as a complete kit, providesd a battery backup facility and does not require alignment. No provision is made, however, to record incoming calls.

Complete kit
44.433BKL

Ready assembled module
44.433F..................... \&
[-_ Jontary (490)
(20) January 1890)
Complete kit
44.433BKL.............. \& $\quad 45.65$
Ready assembled module
44.433F................... \& 87.25


With the ELV IC tester logic function tests can be carried out on nearly all CMOS and TL standard components, accommodated in DIL packages up to 20 pin. The tester is designed as an insertion card for IBM-PC-XT/AT and compatibles. A small ZIF test socket PCB is connected via a flat band cable. Over 500 standard components can be tested using the accompanying comprehensive test software.

IC TESTER for IBM-PC-XT/AT

(Electronics The Maplin Magazine Jun-Jul 88 + Elekor Electronics December 88)

Complete Kit including Textool sokket, connectors, sockets, Flat band cable, PCB, Software
44.4748KL ................ \& 60.85

Ready Assembled Module
4.474F
\& 113.00
Sottware, single
44.474SW
\& 17.85

# FOR YOUR ENTERTIANMENT by Barry Fox 

## Highly Desirable

There is a new breed of shop in Britain which sells electronic gadgetry with a difference. The difference is that everything is high tech, high price and high design, irrespective of manufacturer. You will see a Bang and Olufsen hi fi alongside a Sony personal video player, a Tissot quartz watch carved out of granite, the smallest camcorders, personal stereo players that work under water and radios powered by the sun.

The idea for this kind of shop came from America and over there the format has broadened, to a shop or mail order house, which sells anything that has a high tech feel about it. The leader of the pack is The Sharper Image with a catalogue which costs $\$ 2$ and is read by over twenty million people a month.

The secret of the catalogue is that it's glossy, and makes everything seem eminently desirable. The secret of the shop is that it is laid out to let customers get their hands on the gadgets and try them.
I browsed recently, trying a leather bound chair with motor-driven massaging rollers built in and a footstool that vibrates to sooth tired muscles. A car cassette radio is built into the chair for music while you relax.

An ordinary cordless phone is repackaged in transparent plastics so that all the components are visible. Panasonic has a battery powered vacuum cleaner that looks like a computer mouse, and cleans up desk tops. A Swiss clock runs on air, getting its power from changes in room temperature which cause expansion and contraction of a gas chamber. which in turn winds the main spring.

## Shockproof

There's shockproof baggage, waterproof camcorders, an outdoor loudspeaker system from Bose, and various gaily coloured p.v.c. personal swimming pools with sun-reflecting surfaces, to tan while you soak. A pocket computer helps you quit smoking by telling you when the next one in a cut down programme is due. Of various ultrasonic gadgets, one is for personal massage and another to scare off pests.

Inevitably, there are air ionizers, and a hand-held electronic dictionary that takes phonetic entries and both displays and speaks the correct spelling. A key ring, designed in Germany, has a range of rings which slide round a precision engineered track.

For the car there are radar speed trap detectors and for the health freak there's a hand-held device that takes your blood pressure from a finger ring. A watch with titanium dial creates bizarre colour patterns and a head for a conventional shower produces a pumped spray for
massage. And so it goes on with fancy baggage, sports shoes that float, stylish portable alarms and pocket electronic memos.

The Sharper Image's rival, B N Genius, plays the same mouth-watering game with a video projector that looks like a slide projector (from Sharp), a book with built-in safe, a portable copy machine that sends faxes and a wrist-watch for cyclists that computes speed.

In their own right none of these gadgets is really that special. The trick is either to put them all together in a hands-on showroom or offer them in a glossy catalogue with beautiful colour pictures and text which emphasises the technical tricks which they play.

As so often happens with gadgets, the chances are that most people will buy for the sake of a feature or facility that sounds exciting, but in practice is of no long term value. How many people who have bought video recorders with perfect freeze frame reproduction, actually use it?

## In Reverse

While browsing in The Sharper Image, I came across one gadget which for years I've been hunting. It's called a reverse converter.

Anyone with equipment, like a portable computer, radio, audio recorder or camcorder, which uses rechargeable batteries, will know the value of a charger
which copes with all mains voltages. Too many chargers work on one voltage only, $220 / 240 \mathrm{~V}$ for most of Europe. So when you are in a country like the US, Japan and parts of Southern France with 100 or 110 V mains, you can't charge your batteries - which is no joke if the machine runs only on its own rechargeable cells.

It is easy to buy a plug-in converter that drops $220 / 240 \mathrm{~V}$ mains to $100 / 110 \mathrm{~V}$, so that American and Japanese equipment can be used all over Europe. But until recently there were only two ways of coping with the reverse problem.

One was to buy and carry round a bulky and heavy bench transformer that lifts $100 / 110 \mathrm{~V}$ to $220 / 240 \mathrm{~V}$. The other way is to buy a $100 / 110 \mathrm{~V}$ charger (while abroad) and either carry both chargers or use the foreign one in Europe with a down converter.

- The Sharper Image had the perfect solution. A reverse converter looks exactly like one of the small $220 / 240 \mathrm{~V}-\mathrm{to}-100 / 110 \mathrm{~V}$ droppers but works in the reverse direction, stepping up from $100 / 110 \mathrm{~V}$ to $220 / 240 \mathrm{~V}$.

These reverse converters are made in Hong Kong, but sold in the US by a company called Franzus of Middlebury in Connecticut. The price is around $\$ 15$ or $£ 10$. Doubtless some enterprising firm will soon catch on and import them into the UK.

## Goodbye Disaster

After 35 years of controlling commercial TV and radio the Independent Broadcasting Authority shuts down at the end of the year, to be replaced by the Radio Authority and Independent Television Commission. Although the IBA employs some fine engineers at its laboratory in Winchester, the desk bound brass at its headquarters in Brompton Road have never seemed to understand anything of the technology which ultimately pays their salaries.

For many, the IBA has seemed toothless, even when it did understand what was going on. The Authority let TV-AM win a breakfast TV franchise on the promise of a mission to inform, and then make scuds of money from superficially covering trivia. I am one of several who remembers how the original TV-AM researchers kept wanting help on technical items which never appeared and for which the helpers were never paid.

Brompton Road made a complete hash of the launch of Nicam digital stereo sound for TV, which crept onto the market with almost no publicity.

Most recently the Authority let its satellite operator BSB run amok, to create the

IBA's biggest ever disaster and almost certainly kill the MAC TV system which had been developed at Winchester. With MAC goes the European high definition TV system that Winchester has also been helping to develop.

So when IBA Chairman George Russell gave a farewell speech at the Banqueting House in Whitehall recently the room was packed with people hoping for some horse's mouth comment about the BSB debacle.

But they were disappointed. Although Russell talked at length about the IBA's grand achievements he only mentioned BSB in passing, saying he was sure the audience would understand why he was not going to say more.

What he did do. however, was play a long video tape compilation of excerpts from commercial TV programmes. Although the sound system used in the Hall to amplify Russell's speech worked perfectly, the sound from the IBA's video tape was so appalling that no-one could hear the dialogue or commentary.

All things considered this seemed an appropriate way to say goodbye to the IBA.

# Teach In '91 <br> DESIGN YOUR OWN CIRCUITS 

## Small Signal_Amplifiers

## MIKE TOOLEY BA

PART 2


This ten part series aims to dispell some of the mystique associated with the design of electronic circuits.
It shows how even the relative newcomer to electronics can, with the right approach, design and realise quite complex circuits.
In this secondpart we deal with small-signal amplifiers and introduce a number of useful circuits based on transistors andintegrated circuit operationalamplifiers. Dur design problem is based on a microphone pre-amplifier system and our companion project deals with the construction of a simple intercom.

## Introduction

Amplifiers are found in some form or another in almost every electronic circuit. Since this topic is extremely broad (covering a.c.. d.c. and power amplifiers) we have divided it into two parts. This lirst part deals with amplifiers which operate with signals of relatively small amplitude. These "small-signal" amplifiers are found in the initial stages of a wide variety of electronic apparatus including audio equipment and lest instruments
In general a small-signal amplifies should:
(a) provide a delined value of voltage gain
(b) matintain the desired vallue of voltage gain over a defined range of Treyuencies
(c) exhibit pre-determined constan! values of input and output impedance
(d) operate without producing significant distortion or noise (which would otherwise mar the signal)
(e) operate within prescribed limits over a range of supply volages and ambient temperatures

## Specifications

A typical small-signal amplifier specification may run along the following lines:

| Voltage gain: | 100 |
| :---: | :---: |
| Frequency |  |
| response: | 20 H 21020 kHz att -3dB |
| Hum and noise: | Less than $100 \mu \mathrm{~V}$ r.m.s. |
| Distortion: | Less than $0.01 \%$ t.h.d. |
| Input impedance | 50kilohm |
| Output impedance: | 600hm |
| Supply voltage: | $\begin{aligned} & +9 V \text { to }+12 V \\ & 10 \text { inA (nominal) } \end{aligned}$ |

This specification should be reasonably self-explanatory. The stated volage gain
(i.e. ratio of output voltage to input voltage) is $1(\%)$. Hence, an input signal of 1 mV should produce a corresponding output signal of 100 mV . an input of 2 mV should produce an output of $20(1) \mathrm{mV}$. and so on.
A point will. however, arise when the output voltage no longer increases as the input voltage increases. At this point, the output signal will become "clipped" and severe distortion will result. It is worth noting that. since the maximum amplitude of the output voltage is usually dependent upon the supply voltage. the maximum undistorted output voltage will normally increase if the supply voltage is increased.

The frequency response is a measure of the amplitier's ability to amplify signals over a range of frequencies. Rather than specify the voltage gain at a number of frequencies, we simply state the two frequencies (one lower and one upper) at which "cut-off" is said to occur. The term is. however a little misleading since it would tend to imply that there is no output signal att all beyond the specified cut-off frequencies.

Since the cut-off frequency is usually taken to mean the frequency at which the voltage gain falls by 3 dB from its maximum (mid-band) value. this notion is patently untrue! In fact, a reduction of 3 dB corresponds to a decrease of approximitely 30 per cent (i.e. the voltage gain at the cutoff frequencies will be approximately 70 per cent of its maximum value). Fig. ?. 1 illustrates this point (note how the voltage gain remains substantally constant over the mid-band region but falls off heyond the ewocu-ofl frequencies. $i_{1}$, and $i_{11}$ ).

All amplifiers. however well designed.


Fig. 2.1. Typical frequency response for a small-signal amplifier
produce some noise and some distortion Noise may arise from several sources (including that which is produced within transistors and resistors). In matins operated equipment. "hum" (a signal which is induced from the a.c. supply ma"ns) and "ripple" (a residual a.c. cobrponent present on the d.e. supply rail) are invariably present.
The sample specification lumps all forms of hum and noise logether and simply indicates that the maximum value of unwatnted signal component (i.e. hum and noise) at the output will amount to $100_{\mu} \mathrm{V}$ r.m.s. Since the signal level at this point will normally be several hundred or several thousand times greater. the unwinted component will usually be negligible (this point is further developed later).

Distortion arises from the inherent nonlinearity of transistor characteristics and results in the generation of unwainted hairmonic components. Since a very large number of such components may be present. it is usual to quote a ligure for the "total harmonic distortion" (t.h.d.) present. This is usually specified as a percentage of the maximum output. A value of 0.01 per cent is. to all intents and purposes. negligible.
Input and output impedances are specified in order that the amplifier stage can be "matched" to a particular application. Some input devices (such as microphones and pick-up cartridges) require reasonably accurate termination in order to provide their stated chamacteristies. An amplifier designed for use in such a configuration must. therefore provide a correct value of input impedance.
Matching of output impedances is equally important although an amplifier with a low output impedance will usually drive another amplifier stage having a high input impedance without any special precautions

## Input signal characteristics

Before we start to delve into the circuitry used in a variety of small-signal amplifiers. it is worth considering some of the essential characteristics of the signals which we might have to deal with.
The most important parameters are input levels (usually expressed in r.m.s. or peak-peak voltages) and frequency response (expressed in terms of the maximum and minimum frequency components which must be present in order to ensure reasonably faithful reproduction of the signal).
These parameters will be of considerable importance in the design of the amplifier. We must. for example. ensure that the input stage (and any subsequent stages) are not over-driven by excessive signal amplitudes. Failure to observe this precaution will result in excessive distortion on signal peaks due to "clipping". Furthermore. the frequency response of the circuit must be tailored to the signal.

If the bandwidth of the amplifier is made excessive (in comparison with the signal) we will unnecessarily worsen the signal-tonoise ratio at the output (this topic is also dealt with in greater depth later). If, on the other hand. the bandwidth is made too narrow, the quality of the signal will suffer and the output signal will no longer be a faithful reproduction of the input. This may all sound too ohvious but nevertheless these simple precautions are olten overlooked.
The following information is provided as
a guide to some of the most common signals and sources that will be encountered:

| Signal/source | Typical impedance | Typical amplitude range (r.m.s.) | Typical frequency range |
| :---: | :---: | :---: | :---: |
| High quality microphone used with speech and music | 1500hm | $500 \mu \mathrm{~L}$ (0) 5 miV | 50 Hz .1015 kHz |
| Communications quality microphone (speech only) | 600 hrm | 2 mV 1010 mV | 300 H 710.4 kHz |
| Compact disk (line output) | 50 kilohm | 10 mV to IV | 20 Hz to 20 kHz |
| V.H.F. radio tuner (line output) | 50kilohm | 100 mV to IV | 40 Hz . 1015 kHz |
| Medium wave radio (headphone output) | 1500hın | 200 mV to IV | 200 H 7 to 6 k Hz |
| High quality cassette tape deck (line outpu!) | 50kilohm | 50 mV 101 V | 50 H 7.1012 kH 7 |
| Low quality cassete tape deck (carphone output) | 150 ohm | 200 mV 10 IV | 100 Hz \% 10.6 kHz |
| Record pick-up cartridge (dynamic) | 50 kohm | 2 mV 1010 mV | $20 \mathrm{~Hz}, 1020 \mathrm{kHz}$ |
| Record pick-up (piezoelectric) | 2megohm. | 10 mV 10 1000 mV | 20 Hz 71020 kHz |

## Transistor amplifiers

The prime mover within most practical amplifier circuits is the humble transistor. A single transistor is capable of providing a current gain (in common-emitter mode) of typically between 150 and 300 . This current gain can be easily harnessed to provide voltage amplification.
In order to operate in a linear mode (in which the output signal produced by ant amplifier is a faithful replica of its input) it is necessary to bias the transistor into the linear region of its operating characteristic. In practice. a small current must be applied to the base in order to produce a standing (quiescent) current in the collector. The base bials current is then effectively modulated by the input signal (i.e. the value of base current increases and decreases around the mean value of bias curreit).
The collector current is an amplified version of the base current. This eurrent similarly incereases and decreases around its standing value. Since this action is adequately explained in most electronic text books, we shall develop this theme no further other than 10 state that the value of

| Device | Type | $\underset{\max }{\mathrm{I}_{\mathrm{r}}}$ | $\begin{aligned} & V_{\text {cwi }} \\ & \text { max. } \end{aligned}$ | $V_{\text {stav }}$ max. | $V_{\text {inn }}$ max. | $P_{1}$ max. | $h_{\text {fit }}$ | at 1 . | $f$ typ | Case strile | Application |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 2N2926 | NPN | 100 mA | 18 V | 18 V | 5 V | 300 mW | 200 | 2mA | 200 MHz | T098 | General purpox' |
| 2N370? | PNP | 200 mA | -25V | - 10 V | -5V | 3610 mW | 175 | ? mA | 100 MHz | TO9? | General purpos' |
| 2N3703 | PNP | - 501 mmA | -311mV | . 50 V | . 5 V | $3(1) 10 \mathrm{~mW}$ | 125 | 2 mA | 100 MHz | TO4? | General purposic |
| BC107 | NPN | 100 mA | 4.5 V | 50 V | 5 V | $3(1) \mathrm{mWW}$ | 240 | 2 mA | 250 MHz | T018 | Driver |
| BCIO8 | NPN | $1(\mathrm{MmA}$ | 20 V | 30 V | 5V | $3(1 \mathrm{M}) \mathrm{mW}$ | 240 | 2 mA | 250MHz. | TO18 | Gencral purposic |
| BCIL9) | NPN | 1010 mA | 20 V | 30 V | iV | 3610 mW |  | 2mA | 2510 MHz | TOIX | Low nois |
| BCIH ? | NPN | 800 ma | 610 V | K0V | 5 V | 8(N)IIW | , 3 | 300 mA | $41 \mathrm{MH7}$ | T05 | Driver |
| $\mathrm{BCl}+3$ | PNP | 800 mA | -(b)V | . 60 V | . 5 V | $8(1) 1 \mathrm{~mW}$ | S5 | 2 M InA | (0) MHz | T05 | Drive |
| BC16 | NPN | 50 mA | 30 V | 30 V | 5 V | $30 \% \mathrm{nW}$ | 240 | 2 mA | 251 MHz | TOM? | Gencral purpose |
| BCI69 | NPN | 100 mm | 20 V | 30 V | 5V | 300 mW | 310 | 2 mA | 250 MHz | TO4? | Low noise |
| BC171 | NPN | 100 mA | 45 V | 50 V | iv | 3100 mW | 2 40 | ? mA | 251 MHz | TOY) | Plastic BC107 |
| BC177 | PNP | 100 mA | -4iV | -50V | -5V | 3001 nW | 240 | 2 mA | 2(0) MH 7 | TOIX | BC107 complement |
| BC178 | PNP | - 100 ma | .25V | -30V | -5V | 3010 mW | 240 | 2 mA | ? (1) MHz | TOIX | BC10x complement |
| BC179 | PNP | -101ma | -30V | -35V | -5V | 3(1) 11 mW | 240 | 2mA | ?(1) MHz | TOIX | BC Iny complement |
| BCIXIL | NPN | 20.111 A | 50 V | (i)V | 5 V | 311nW | 225 | 2 mA | 150 MHz | TOM? | Driver |
| BC183L | NPN | ? 30 mA | 30 V | 45 V | 5 V | 3100110 W | 250 | 2 mA | 151) MHz . | TO9? | General purpos. |
| BCIXAL | NPN | ? (MmIA | 30 V | +iv | 5 V | 31010 mW | 250 | 2 mA | I50MHz. | TOY? | Low noisc |
| BC? 12 L | PNP | - $2(1) \mathrm{mA}$ | .50) V | -60V | -iV | 300 nl W | $3(10)$ | 2 mA | 200 M \% | TOY? | BCIXSL complement |
| BC? 13 L | PNP | - $-(1) \mathrm{mA}$ | $-30 \mathrm{~V}$ | -45 | -5V | 310 mW | 3 m | 2 mA | ? (0) MHz | TO9? | BCI 3 S L complement |
| BC? 14 L | PNP | $-201 \mathrm{~mA}$ | -30V | -45V | . 5 V | 310001 W | 225 | 2 mA | ? (1) MH\% | TO9? | BCIXAL complement |
| BC47x | PNP | -50mA | -40V | -40V | -5V | 361 mW | 175 | 2mA | ISOMHz. | TOIX | BCIOs complement |
| BC. $4^{4}$ | NPN | 100 mA | 30 V | 30 V | 5 V | 6251015 | 350 | 2 mA | 300 MHz | TO93, | Plastic BC10\% |
| BC549 | NPN | 10 mm | 30 V | 30 V | 5 V | $6 ? 5 \mathrm{~mW}$ | 250 | 2 mA | 300 MHz | TO4? | Plaslic BCl(1) |
| BC55\% | PNP | $-100 \mathrm{~mA}$ | -45V | $-50 \mathrm{~V}$ | $-5 \mathrm{~V}$ | 615 nW | 250 | 2 mA | 360 MHz | TO92a | Plastic BC 178 |
| BC559 | PNP | -100mA | -30V | -30 | . 5 V | 6.510 mW | 250 | 2 mA | 250 MHz | TO9? | Plastic BCI79 |
| BCY70 | PNP | - 300 mA | -40V | -50V | -5V | 300 moW | 150 | 2 mA | 200 MHz . | T018 | General purpose |
| BCY71 | PNP | - 2000 mA | -4.5V | 45 V | -SV | 3010 mW | 150 | 2 mA | 300 MHz . | TO18 | General purpose |
| ZTX107 | NPN | 100 mA | 50 V | 60 V | 5 V | 350 mW | 340 | 2 mA | 30 MMHz | E-linc | Plastic BC107 |
| ZTX10X | NPN | $1(\mathrm{KmA}$ | 30 V | 45 V | 5 V | 250 mW | 300 | 2 mA | 350 MHz . | E-line | Plastic BC108 |
| ZTX109 | NPN | 100 mA | 30 V | 45 V | 5 V | 250 mW | 400 | 2 mA | 350 MHz | E-line | Plastic BC109 |
| ZTX300 | NPN | 510 mA | 35 V | 25 V | 5 V | 300 mW | 150 | 10 mA | 350 MHz | E-line | General purpose |
| ZTX500 | PNP | - 5010 mm | -25v | .25V | -5V | 3010 mW | 150 | 10 mA | ISOMHz | E-line | ZTX 3(1) complement |

The column headings used in the foregoing table deserve a little more explanation They have the following meanings:
Heading Meaning

Device Manuliaturers coding
Type Transistor type: either npm or pmp
Is max. Maximum value of collector current
$\mathrm{V}_{\text {avo }}$ max. Maximum value of collectoremitter voltage with base left open-circuit
$V_{\text {cho }}$ max. Maximum value of collectorbase voltage with base left open-circuit
$\mathrm{V}_{\mathrm{ch}, \mathrm{m}}$ max. Maximum value of (reverse) emiter-base voltage with base left open circuit
P1max. Maximum total power dissipation
$\mathrm{h}_{\mathrm{fi}} \quad$ Small-signal current gain (in common-emilter configuration)
at $I_{c} \quad V$ alue of collector current at which the sinall-signal current gain is yuoted
fityp. Transition frequency (i.e the frequency all which the common-emitter current gain falls to unity)
Calse style Transistor encapsulation (and pin connections:
Application Recommended use for the device (or generic type if the device is a plastic encapsulated or complementary version of another transistor)

Question 1: The following transistor: are available: BC142. BC107, and BC109. Which of the transistors is most suitable for operating:
(a) in a complementary configuration in conjunction with a BC177
(b) as the first stage of a pre-amplifier in a low-noisc amplifier
(c) in a circuit where the collectoremitter voltage is expected to be in the range 20 V to 40 V with a maximum collector current of 15 mA ?

## Gain groups

Various transistors (such as the BC107. BC 108. BC109. BC 168 . and BC169) are available in one of several "gain groups" The gain group is denoted by means of a sulfix letter ats follows:

> Suffix

Current gain ( $\mathrm{h}_{\mathrm{fc}}$ )

| $A$ | $110-220$ |
| :--- | :--- |
| $B$ | $2(0)-450$ |
| B | $420-800$ |

Thus a BC108B can be expected to exhibit a current gain in the region 200 to 450 (when operated with a nominal ImA colfector current). If high gain is important. it is thus important to obtain at transistor from the highest available gain group.
Readers should note that al least one supplier only supplies devices from the highest gain groups. thereby ensuring that nerformance specifications are more than adequately met!

## Low noise amplifiers

Special precautions must be taken with the design of amplifiers which are to be operated with very small signals. A meaningful indication of the effect of noise within in amplifier is provided by the
"signal-to-noise ratio". Clearly, the effect of noise will be more significant when the relative magnitude of the signal present is reduced (e.g. in the quieter passages of a piece of music).
The signal-(o-noise ratio provides : yardstick which allows the eflect of noise to be yuantilied in relation to the signal present. Since the signal cannot readily he separalted from the noise at the output, the signal-to-noise ratio is usually specified ats:

Signal-to-noise ratio

$$
\begin{aligned}
& =10 \log _{10}\left(\frac{P_{\text {SII } i+N O S I:}}{P_{\text {NOIS: }}}\right) d B \\
& =20 \log _{111}\left(\frac{V_{\text {SII } i+N O S I:}}{V_{\text {NOIS: }}}\right) d B
\end{aligned}
$$

where $P_{\text {ski } i+\text { Nols: }}$ is the r.m.s. output power developed with the signal present (i.e. including the residual noise). $P_{\text {NoISI: }}$ is the r.m.s. outpul power without the signal present (i.e. residual noise only present). $\mathrm{V}_{\text {si(i }+\mathrm{Nolsi}}$ is the r.m.s. oulput voltage developed with the signal present (i.e. including the residual noise), and $\mathrm{V}_{\text {Nolss: }}$ is the r.m.s. output
tively low values of collector current (e.g. $100 \mu \mathrm{~A})$. Maximum current gain, on the other hand. usually corresponds with much larger values of collector current (e.g. $2 \mathrm{~mA})$. Hence, there is a trade-off hetween gain and noise performance.
In the first stage of a multi-stage amplifier we would thus usually sacrifice gain in favour of noise performance and operate the transistor with a low value of collector current. Subsequent stages (where noise is not so critical) could benefit from the barger values of current gain which can be oblained by employing larger values of collector current.

## Transistor amplifier configurations

Three different circuit configurations are employed for transistor amplifiers depending upon which of the three transistor connections is made "common" to both input and output (see Fig. ?.2). The three basic circuit conligurations exhibit the following charateristics (lypical values given in brackets):

| Parameter | Mode of operation |  |  |
| :---: | :---: | :---: | :---: |
|  | Common emitter | Common collector (emitter follower) | Common base |
| Voltage gain | medium/high (40) | unity (1) | $\begin{gathered} \text { high } \\ (200) \end{gathered}$ |
| Current gain | $\begin{aligned} & \text { high } \\ & (200) \end{aligned}$ | high <br> (2())) | unity <br> (I) |
| Power gain | very high (8000) | $\begin{aligned} & \text { high } \\ & (\geq 00) \end{aligned}$ | $\begin{gathered} \text { high } \\ (\underline{2} 00) \end{gathered}$ |
| Input resistance | medium | high (100kilohm) | low <br> (?OWhm) |
| Output resistance | medium high (?0kilohm) | low (10)ohm) | high (lookilohmn) |
| Phase shift | 180 deg . | 0 dcg . | 0 dcg . |
| Fig. reference | 2 (a) | 2(b) | 2 (c) |
| Typical applications | General purpose. a.f. and r.f. amplifiers | Impedance matching, input and output stages | R.F. and V.H.F. amplifiers |

power without the signal present (i.e.
residual noise only present).
The eflect of passing a signal through an amplifier is a degradation of the signal-tonoise ratio. This is sometimes specified as the "noise figure" and is defined as:
Noise ligure

> signal-to-noise ratio at the input
signal-to-noise ratio at the output
Note that, where the signal-to-noise ratios are expressed in decibels ( dB ). the noise figure (expressed in dB ) is found by simply subtracting the signal-to-noise ratio at the output from the signal-to-noise ratio at the input.

Question 2:An amplifier has matched input and output impedances of 600 ohm. The signal (plus noise) at the input has an r.m.s. voltage of 20 mV . In the absence of a signal. the noise at the input has an r.m.s. value of mV . The signal (plus noise) at the output of the amplifier hals an r.m.s. voltage of 800 mV . If. in the absence of the signal the noise at the output has an r.m.s. value of 60 mV . determine:
(a) amplifier voltage gain (expressed as al ratio)
(b) amplifier volage gain (expressed in dB)
(c) signal-to-noise ratio at the input
(d) signal-to-noise ratio al the output
(c) amplifier noise figure

## Noise versus gain

It is importiont to note that most transistors provide optimum low-noise performance when operating with rela-

Fig. 2.2. Transistor amplifier operating configurations



Fig. 2.3. Simple common-emitter amplifier stage


Fig. 2.6. Emitter stabilised commonemitter amplifier stage

## Practical amplifier circuits

The basic circuit of a bipolar transistor operating in common emitter mode is shown in Fig. 2.3. This simple circuit can provide voltage gains of around 100 with a frequency response extending from about 60 Hz to several hundred kilohertz. The input impedance is approximately 2.5 kilohm. The maximum input signal amplitude is approximately 10 mV r.m.s. (approx 30 mV pk-pk).

The circuit ol Fig. 2.3 employs negative feedback stabilised bias (i.e. the base current is derived from the value of collectoremitter voltage). Fig. 2.4 illustrates the feedback process. Both a.c. and d.c. feedback is present. the amount of the former type of feedback (which reduces the overall. stage gain) is very much dependent upon the source impedance. Where the gain reduction is significant, the bials resistor can be split and the al.c. negative feedback component removed by means of the arrangement shown in Fig. 2.5.
An emitter-stabilised common-emitter amplifier stage is shown in Fig. 2.6. This stage provides a voltage gain of around 125 coupled with an input impedance of around 2 kilohm. The bandwidth extends from around 10 Hz to well over 100 kHz .
An improved two-stage amplifier in which TRI operates as a common-emitter stage whilst TR2 is connected as an emitter-follower is shown in Fig. 2.7. The stage employs negative feedback stabilised bias by means of R2. One significant advantage of this circuit over that shown in Fig. 2.6 is that the output impedince is much reduced (typically approximately


Fig. 2.4. Feedback path in the simple common-emitter stage


Fig. 2.7. Two-stage amplifier with lowimpedance output

50 ohm ). The circuit of Fig. 2.7. provides a typical gain of 200 with a bandwidth of 50 Hz to 120 kHz (or more). The input impedance is approximately 3 kilohm.

Unfortunately, the voltage gain offered by the circuit of Fig. 2.7 (and all previously described amplifier stages) is very much dependent upon the individual value of transistor current gain ( $h_{\text {fis }}$ ). Since $h_{\text {ic }}$ can vary (often by as much as 50 per cent) for a given transistor type, the stage gain is not readily predictable unless the transistors are specially selected.
In many applications, it is necessary to determine the actual stage gain by means of the application of a pre-determined amount of negative feedback. Fig. 2.8 shows how this can be achieved within the circuit of Fig. 2.7. The voltage gain of the circuit with this modification in place is given simply by the ratio of $R_{L}$ to $R_{E}$ (provided that this ratio is not too large). Hence, if we require a stage gain of 10 . we could simply make $R_{\mathrm{L}} 4 \mathrm{k} 7$ (as in Fig. 2.7) and add an emitter resistor, $\mathrm{R}_{\mathrm{E}}$, of $470 \Omega$. It is. however. important to note that. by incorporating this type of negative feedback, the input impedance will be raised.

Without $R_{E}$ (i.e. with $R_{E}$ replaced by a short-circuit) the stage offers an input impedance of approximately 3 k . With $\mathrm{R}_{\mathrm{E}}$ oit $470 \Omega$, however, the input impedance is increased to around 25 k . This may, or may not, be desirable!

In order to obtain a high value of input impedance, a lield-effect transistor (connected as a source-follower) may be added at the input of the circuit of Fig. 2-. Fig. 2.9 shows an amplifier based on this arrangement which provides a voltage gain


Fig. 2.5. Reducing the amount of a.c. feedback


Fig. 2.8. Modification to Fig. 2.7 to permit gain setting by negative feedback
of approximately 200 coupled with an input impedance of approximately 4M5. This circuit makes an ideal input stage for an instrumentation amplifier:
The maximum values of voltage gain which can be achieved by the previously described circuits (which essentially all rely on the gain provided by a single commonemitter stage) is limited to about 200 . Clearly this value of gain will be inadequate for a number of applications. Fig. 2. 10 shows a two-stage amplifier in which both stages operate in common-emitter mode. The overall stage gain is given (approximattely) by:

Voltage gain. $A_{v}=\frac{R 3}{R 1} \times \frac{R 4}{R 5}$
Note that maximum gain is achieved with RI short-circuit. in which case the


Fig. 2.9. Adding an f.e.t. input stage to obtain a very high input impedance


Fig. 2.10. Two-stage high gain amplifier


Fig. 2.12. Modification to the general purpose transistor amplifier module in order to permit gain setting


EE21530

Fig. 2.13. Modification to the general purpose transistor amplifier module in order to permit variable gain adjustment


Fig. 2.11. General purpose transistor amplifier module
overall voltage gain will typically be in the region of 350 to 500 with the values shown.
If even more gain is required, then R5 may be bypassed to a.c. by means of an electrolytic capacitor of approptiate value. Gains of up to $5(0) 0$ can be achieved by this means.

## General purpose transistor amplifier module

The circuit diagram of a general purpose transistor amplifier module is shown in Fig. 2.11. This circuit can be configured in a variety of ways and the frequency response and voltage gain can be readily pre-defined. A BC109 (or BC549) should be used for TRI whilst a BC108 (or BC 548 ) should be used for TR2. The circuit requires a 9 V d.c. supply rail and consumes a mere 2 mA , or so.

With C3 connected and values as shown in Fig. 2.11. the voltage gain for the circuit is approximately 5000 , The maximum input voltage in this configuration should not be allowed to exceed ImV pk-pk.
Where reduced values of gain are required. a fixed resistor may be wired in series with C3. as shown in Fig. 2.12. With $R_{E}$ of $\mathbf{2} \mathbf{2} 0 \Omega$, the voltage gain is reduced to approximately 700 . Alternatively. $R_{E}$ may he replaced by a miniature preset potentiometer of 220 , as shown in Fig. 2.13. The voltage gain may then be adjusted


Fig. 2.14. P.C.B. layout for the general purpose transistor amplifier module
to any required value in the range 700 to 5000 approximately.
The low frequency response of the module can be determined by choice of an appropriate value for Cl whilst the high frequency cut-off is determined by the value of C4. Note that, when operating in the high-gain configuration (with C3 present) it may be essential to have a minimum value of C4 present (e.g. 47p) in order to ensure unconditional high-frequency stability.
The copper foil and p.c.b. component layouts for the general purpose transistor amplifier module are shown in Fig. 2.14.

## Low-power output driver

Occasionally, we require a small-signal amplifier to drive a low impedance load. In such canses, we will require a low-power output stage which can source a reasonable amount of current. Fig. 2.15 shows a typical solution to this problem. TR1 and TR2 operate as complementary emitterfollowers. Resistors R1, R2 and the forward biased diodes. DI and D2, provide the necessary bias for TRI and TR2. R3 and R4 provide a degree of lemperature stabilisation for the output transistors which may operate at an appreciable collector current when driving a lowimpedance load with a signal ol appreciable amplitude. The stage provides unity voltage gain.

## Operational amplifiers

In recent years. the advent of the high-performance operational amplifier has provided the electronic circuit designer with another means of faithfully amplify-
ing low-level signals. For the benelit of the newcomer. operational amplifiers are simply analogue integrated circuits which olfer near-ideal characteristics (virtually infinite voltage gain and input resistance coupled with low output resistance and wide bandwidth). As such. they can be used effectively as a "block of gain" which can be inserted readily into a circuit without the need for complex bias arrangements.
Operational amplifiers are available packaged singly, in pairs (dual types). or in fours (quad types). The TL081. for example. is a single general purpose b.i.f.c.t. operational amplifier which is also available in dual (TL082) and quad (TL084) forms.


## EE28556

Fig. 2.15. Low-power complementary output dirver for use with low impedance loads.

Where several operational amplifiers are required in a circuit, it usually makes good sense to use dual or quad devices rather than single devices in order to minimise the number of components and reduce the size of the p.c.b. required.

## Selecting operational amplifiers

When selecting an operational amplifier for use in a small-signal amplifier, the open-loop voltage gain and input bias current ratings are invariably less important parameters than supply voltage range and slew rate. This last named parameter provides a measure of the response of an operational amplifier to a step function input; the larger the slew rate, the wider the frequency reponse.

The following table relates to the most popular types of operational amplifier fot use in small-signal amplifiers (integrated circuit power amplifiers will be dealt with separately in Part 3). Note that the TL071 and TL081 series of f.e.t. input operational amplifiers offer excellent all-round performance in terms of operating voltage range. open-loop voltage gain and slew rate.
whilst that for Fig. 2.16(b) is given by:

$$
A_{V}=\frac{V_{\text {OUT }}}{V_{\text {IN }}}=\frac{R_{\mathrm{K}}}{R_{\text {IN }}}
$$

For optimum performance. $R^{\prime}$ is equal to the parallel combination of $R_{1 N}$ and $R_{F}$ in order to retain symmetry: Hence (for both circuits):

$$
R^{\prime}=\frac{R_{1 N} \times R_{1}}{R_{1 N}+R_{F}}
$$

Practical inverting and non-inverting amplifier stages are shown in Fiys. 2.17 and 2.18. With the values shown, these stages provide a nominal voltage gain of 10 (note that the capacitors should be non-electrolytic types). Fig. 2.19 shows an audio frequency amplifier which has cut-off frequencies and bandwidth determined by the component values employed. The following equations relate to this circuif: Mid-band voltage gain: $A_{V}=R_{2} / R_{1}$ Mid-band input impedance: $Z_{1 N}=R_{1}$ ohm Lower cut-off frequency:

$$
r_{L}=\frac{1}{2 \pi C_{1} R_{1}}=\frac{0.159}{C_{1} R_{1}} \mathrm{~Hz}
$$

Upper cut-off frequency:
$f_{H}=\frac{1}{2 \pi C_{2} R_{2}}=\frac{0.159}{C_{2} R_{2}} \mathrm{~Hz}$

| Device | Type | Package | Supply <br> voltage <br> range <br> (V) | Open loop voltage gain (dB) | Input bias curren (A) | Slew rate ( $\mathrm{V} / \mathrm{us}$ ) | Output voltage (V) (note I) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| LF347 | FET | Quad | $\pm 510 \pm 18$ | 110 | 50p | 13 | $\pm 13.5$ |
| LF351 | FET | Single | $\pm 510 \pm 18$ | 110 | 50p | 13 | $\pm 13.5$ |
| LF 353 | FET | Dual | $\pm 510 \pm 18$ | 110 | 50p | 13 | $\pm 13.5$ |
| LF 355 | FET | Single | $\pm 410 \pm 18$ | 106 | 30 p | 5 | $\pm 1.3$ |
| LM 324 | Bipolar | Quad | 31032 (or $\pm 15$ | 5) 100 | $45 n$ | 0.6 | $28($ or $\pm 14)($ note 2) |
| LM 348 | Bipolar | Quad | $\pm 10$ to $\pm 18$ | 96 | 30 n | 0.6 | 28 (note 2) |
| NE531 | Bipolar | Single | $\pm 510 \pm 18$ | 96 | 400n | 35 | $\pm 15$ |
| TL071 | FET | Single | $\pm 310 \pm 22$ | 106 | 30p | 13 | $\pm 13.5$ |
| TL07? | FET | Dual | $\pm 310 \pm 18$ | 106 | 30p | 13 | $\pm 13.5$ |
| TL074 | FET | Quad | $\pm 310 \pm 18$ | 106 | 30 p | 13 | $\pm 13.5$ |
| TL081 | FET | Single | $\pm 310 \pm 18$ | 106 | 30 p | 13 | $\pm 13.5$ |
| TL082 | FET | Dual | $\pm 310 \pm 18$ | 106 | 30p | 13 | $\pm 13.5$ |
| TL084 | FET | Quad | $\pm 310 \pm 18$ | 106 | 30p | 13 | $\pm 13.5$ |
| 741 | Bipolar | Single | $\pm 510 \pm 18$ | 106 | 80 n | 0.5 | $\pm 13$ |
| 7415 | Bipolar | Single | $\pm 510 \pm 18$ | 100 | 200 n | 20 | $\pm 13$ |
| 747 | Bipolar | Dual | $\pm 710 \pm 18$ | 106 | 80 n | 0.5 | $\pm 13$ |
| 748 | Bipolar | Single | $\pm 710 \pm 18$ | 106 | $80 n$ | 0.8 | $\pm 13$ |
| 759 | Bipolar | Single | $\begin{aligned} & 710.36 \text { (or } \pm 3 \text {. } \\ & 10 \pm 18) \end{aligned}$ | . 5106 | $50 n$ | 0.5 | $\pm 12.5$ |

Notes

1. Measured with supplies of 15 V (or $\pm 15 \mathrm{~V}$ ).

## Operational amplifier <br> configurations

The three basic configurations for an operational amplifier are shown in Fig. 2.16. The circuit of Fig. 2.16(a) shows an inverting amplifier which provides a phase shift of 180 degrees between the input and output signals. The circuit of Fig. 2.16(b) shows a non-inverting amplifier which provides a phase shift of 0 degrees (i.c. input and output signals are in-phase) whilst that of Fig. 2.16(c) shows a differential amplifier (the output voltage is proportional to the difference between the two input voltages).
The voltage gain ( $A_{v}$ ) of the circuit in Fig. 2.16(a) is given by

$$
A_{V}=\frac{V_{\text {OUT }}}{V_{\text {IN }}}=-\frac{R_{E}}{R_{\text {IN }}}
$$

(the minus sign indicates inversion not atttenuation)

## Bandwidth: $f_{W}=f_{H}-f_{L} H z$

## General purpose operational amplifier module

The circuit diagram of a general purpose operational amplifier module is shown in Fig. 2.20. This circuit can be configured in a variety of ways and the frequency response and voltage gain can be readily pre-defined. The operational amplifier. ICI, should be a TL072 or TL08? dual device.
With the values shown. each half of the circuit provides a nominal mid-band voltage gain of 22 and the frequency response extends from 160 Hz to 7.2 kHz approximately. The two stages may be used independently (e.g. as "left" and "right" channel amplifiers in a stereo system) or may be conneeted in tandem to provide a high gain

Fig. 2.16. Operational amplifier circuit configurations

(a) inverting amplifier

(b) non-inverting amplifier

(c) differential amplifier


Fig. 2.17. Practical inverting amplifier


Fig. 2.18. Practical non-inverting amplifier


Fig. 2.19. Bandwidth limited inverting operational amplifier


Fig. 2.21. P.C.B. layout for the general purpose operational amplifier module
aimplifier stage. The circuit requires a dual rail power supply of $\pm 5 \mathrm{~V}$ to $\pm 15 \mathrm{~V}$. The copper foil and p.c.b. component layouts for the general purpose operational amplifier module are shown in Fig. 2.?1
Question 3: An audio amplilier stage is to operate according to the following specifications:

Mid-band voltage gain $=100$
Mid-band input impedance $=6(\%) \mathrm{ohm}$
Lowercut-oll frequency $=300 \mathrm{~Hz}$
Upper cut-oll irequency $=\mathbf{3 . 4} \mathbf{k H z}$.
Devise a suitable circuit arrangement based on an operational amplifier. Specify the value of atl components used.

## Design Problem

This month's design problem (as with all ol the design problems presented in this series) is designed for readers who would welcome the opportunity of tackling a little "homework". The exercise may be tackled purely "on paper" or may be used as the basis of a complete constructional project.
This month's problem arises from the need for at means of amplifying the low-level audio signal from a dynamic microphone:
A puhlic cuddress umplifier provides full


Fig. 2.20. Circuit diagram of the general purpose operational amplifier module

## COMPONEVTS

General Purpose Transistor Amplifier
Resistors

| R1 | 4 k 7 |
| :--- | :--- |
| R2 | 330 k |
| R3 | 1 k |
| R4 | 10 k |
| R5 | 2 k 2 |
| R6 | 470 |

All resistors are $0.25 \mathrm{~W} \pm 5 \%$ carbon

## Capacitors

| C1 | $1 \mu$ axial elect. 63 V |
| :--- | :--- |
| C2 | $220 \mu$ radial elect. 35 V |
| C | $100 \mu$ radial elect. 35 V |
| C4 | 47 p miniature plate ceramic |
| C5 | $10 \mu$ radial elect. 16 V |

See
TALK
Page

## Semiconductors

| TR1 | BC109 |
| :--- | ---: |
| TR2 | BC108 |

## Miscellaneous

PL1 to PL3
2-way straight p.c.b. headers ( 0.1 inch pitch), 3 off
SK1 to SK3 2-way socket housing ( 0.1 inch pitch), 3 off
Printed circuit board, available from the EE PCB Service, order code EE717

## General Purpose Operational Amplifier

## Resistors

$$
\begin{array}{ll}
\text { R1, R3, R5, R6 } & \text { 10k (4 off } \\
\text { R2. R4 } & 220 \mathrm{k}(2 \text { off }) \\
\text { R7, R8 } & 100(2 \text { off }
\end{array}
$$

All resistors are $0.25 \mathrm{~W} \pm 5 \%$

## Capacitors

| C1 | 100n 100 V miniature dipped polyester |
| :--- | :--- |
| C2 | $100 \mathrm{p} \mathrm{100V}$ miniature ceramic plate |
| C3 | 100 n 100 V miniature dipped polyester |
| C4 | $100 \mathrm{p} \mathrm{100V}$ miniature ceramic plate |
| C5 | 220 1 radial elect. 35V |
| C6 | $220 \mu$ radial elect. 35 V |

## Semiconductors

IC1 TL072 or TL082

## Miscellaneous

PL1, PL2 3-way straight p.c.b. headers ( 0.1 inch pitch), 2 off
SK1, SK2 3-way socket housing ( 0.1 inch pitch). 2 off
Printed circuit board, available from the EE PCB Service, order code EE718; terminal pins 1 mm ( 3 off ).


Fig. 2.22. Voltage calibrator circuit (solution to last month's Design Problem)
rated output from a nominal line input of IV r.m.s. The amplifier is to be fitted with a microphone input facility. which will matcha a sok impedence dinemvic microphome. If the microphone produces a nominal oututut of 2 mV r.m..... design a simple pre-cunplifier hased on a clual operational amplifier:

## Answer tolast month's design problem:

An oscillescope is to be filted with an accurate d.c. pollage calihrator. This circuit is to provide fixed d.c. levels of loomv. IV and loV at "fomt-panel test point. The roltage le bels should be accurate to within $\pm 2$ per cenn and the voluge level should be maintained within this range when the lest posint is "loaded" h! a standerd oscillosscope input (equivalent to 1 M ). Dessign a suituble circuit arrangement based on a standard roltage regulator operating firmn at internal +18 V rail. Specifi: all component values required.

One solution to last month's design problem is shown in Fig. 2.22. A lowpower 12 V regulator (78L012) is used to provide an accurate (to within $\pm 2$ per cent) supply which is then applied to a voltage divider chain. To keep things fairly simple, we have made the total resistance of the divider chain $1 \mathbf{k} 2$ and, since the output of the regulator is 12 V , a current of 10 mA will flow in the divider chain. The required voltages are simply "tapped off" from resistance of $10 \Omega(100 \mathrm{mV}), 100 \Omega$ ( IV ), and $1 \mathrm{k}(10 \mathrm{~V})$. The rather awkward values of

Cumulative index to modules

| Title | Part | Function/specification |
| :--- | :--- | :--- |
| Dual output power <br> supply module | 1 | Dual $\pm 5 \mathrm{~V}$. $\pm 12 \mathrm{~V}$ or $\pm 15 \mathrm{~V}$ regulated power supply <br> rated at 1 A max. output. |
| 723 variable power <br> supply module | 1 | Signal variable output of +2 V to +37 V at up to 5 A max <br> Output voltage and current limit are set by means of <br> preset controls. |
| L200 variable power <br> supply module | 1 | Single variable output of +2.7 V to +35 V at up 102 A <br> max. Output voltage and current limit are set by means of <br> variable controls. |
| General purpose <br> transistor <br> amplifier module | 2 | Pre-defined voltage gain and frequency response. <br> Low/medium input impedance, low output impedance. <br> Requires a single 9 V d.c. supply at 2 mA nominal. |
| General purpose <br> operational <br> amplifier module | 2 | Pre-defined voltage gain and frequency response. Two <br> stages may be used independently (e.g. for stereo <br> operation) or connected in tandem. Requires a dual <br> supply of between $\pm 5 \mathrm{~V}$ and $\pm 15 \mathrm{~V}$ at 10 mA nominal. |

resistance needed in the chain can easily be made up from preferred resistors as follows:

R1 (200 2 ): $2 \times 100 \Omega$ in series
R2 ( $900 \Omega$ ): $2 \times 1 \mathrm{k} 8$ in parallel
R3 ( $90 \Omega$ ): $2 \times 180 \Omega$ in parallel
All the resistors should be two per cent (or better) close tolerance types.

## Answers to questions in Part Two

Question 1

| (a) | BC 107 |
| :--- | :--- |
| (b) | BC 109 |
| (c) | BC 142 |

Question 2

| (a) | 40 |
| :--- | :--- |
| (b) | 32 dB |
| (c) | 26 dB |
| (d) | 27.5 dB |
| (e) | 3.5 dB |

## Question 3

Since the mid-band input impedance is to be $600 \Omega$. R1 must be $600 \Omega$ (in practice we could use two 1 k ? resistors connected in parallel). In order to provide a mid-band voltage gain of 100, R2 must be 100 times $600 \Omega$ (or 60 k ). This value can be achieved by wiring two 120 k resistor in parallel.
If. however, we can tolerate a small error in the input impedance fless than 10 per
cent) we could make use of resistors of the nearest preferred values. We could thus make RI and R2 $560 \Omega$ and 56 k . respectively.
The values of Cl and C 2 can be calculated from:

$$
\mathbf{C l}^{1}=\frac{0.159}{R \mid f_{\mathrm{L}}}
$$

hence

$$
\mathrm{C}^{\mathrm{C}}=\frac{0.159}{560 \times 300}=9.46 \times 10^{.7} \mathrm{~F}
$$

thus
$\mathbf{C l}=0.946$ (nearest preferred value $1 \mu$ )
Similarly.

$$
C 2=\frac{0.159}{R 2 f_{H}}
$$

hence

$$
\mathbf{C} 2=\frac{0.159}{56 \times 10^{9} \times 3.4 \times 10^{0}}=0.835 \times 10^{-9} \mathrm{~F}
$$

thus
$C 2=835$ p (nearest preferred value 820 p)
Hence the final circuit arrangement will be similar to that shown in Fig. 2.19 with:
$R 1=560 \Omega, \quad \mathrm{R} 2=56 \mathrm{k} . \quad \mathrm{Cl}=1 \mu$, and C2 $=820 \mathrm{p}$

Next month: In next month's instalment we deal with power amplifiers and consumer audio frequency integrated circuits. Our design problem involves a guitar amplifier and our constructional project features a bench amplifier/signal tracer.

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## Constructional Project

# AMSTRAD PCW STEREO SOUND GENERATOR 

 JASON SHARPE
#### Abstract

Uninspired by your PCW's 'bleeping'? Well now you can have three channel stereo sound effects.


THIS project gives the PCW three channel (left, left and right (centre), right) stereo sound, the frequency and volume of each channel is software programmable, "noise" can also be mixed into any or all channels, there is also an envelope generator.
The circuit was originally designed and built around the AY-3-8913, however this chip is no longer available and the slightly
more expensive AY-3-89/2 has had to be substituted. The later type is an identical device but has an additional I/O port which is not required in this design.

## HOWIT WORKS

The AY-3-8912 programmable sound generator (more about this chip later). is connected to the PCW using a Z80 PIO (ICl in Fig.1). The PIO has two ports (port
$A$ and port $B$ ), port $B$ is used to transfer data and addresses to IC4. three of the eight lines on port A are used for control signals and resetting IC4, the other five lines are unused.
Integrated circuit IC2 is a three to eight line decoder, which is used as an address decoder to select the PIO. The PIO is enabled between ports 160 and 163 (inclusive). IC 3 divides the $\mathrm{Z} 80^{\prime}$ 's 4 MHz clock by two, to supply IC 4 with a 2 MHz clock.
The three audio outputs (A,B,C) of the AY-3-8912 are available on pins 5, 4 and I respectively, these are fed into a simple mixer circuit comprising of resistors RI. R2, R3 and R4. The result is two sound channels, one consisting of channel $A$ and half (volume) of channel B. the other channel $\mathbf{C}$ and half of channel $\mathbf{B}$. this gives the

effect of channel B being "between" the two speakers.
The output of the mixer is fed into a stereo amplifier (IC5) through volume control (VRI). IC5 is a TDA2822 low power stereo amplifier, this amplifier i.c. is used because of its low voltage requirement and also because it needs very few external components. The output of this i.c. is sufficient to drive a pair of personal stereo headphones or small speakers, alternatively the output can be amplified again by using an external amplifier.

## CONSTRUCTION

The sound generator can be quite easily constructed on the p.c.b. shown in Fig. 2. when soldering in the components be very careful not to short out any tracks with solder as some of them are very close together.
When assembling fit the resistors and link first, then the i.c. sockets. followed by capacitors. The 50 way cable can now be connected to the board, the prototype used a ready made cable terminated with a 50 way edge connector at one end and a 50 way transition header at the other. Next insert the i.c.s and then the stereo output socket and volume control, use screened cable for these to prevent to much digital noise getting into the signal. also if the case of the volume control is metal connect it to 0 V .

## TESTING

Before plugging the unit into the computer check the board for short circuits caused by solder etc.

Make sure you have the edge connector the right way up, with pin 1 as shown in Fig. 2. (It is best to fit a polarising key between pins 22 and 24 (see Fig. 3.), and plug the unit into the expansion slot on the back of the PCW. Plug the headphones or speakers in. switch on and boot up as nor-

\section*{COMPONFVITS <br> Resistors <br> R1, R2 470 k ( 2 off ) <br> R3, R4 820k (2 off) <br> R5, R6 $4 \Omega 7$ (2 off) <br> All $1 / 2 \mathrm{~W} 5 \%$ carbon. <br> Potentiometers <br> VR1 47 k stereo log. <br> Capacitors <br> C1 to C3 $100 \mu$ elect. 12 V ( 3 off) C4, C5 $570 \mu$ elect. 12 V ( 2 off ) C6 to C9 100n (4 off) <br> Semiconductors <br> | IC1 | Z8OPIO |
| :--- | :--- |
| IC2 | 74 LS 138 |
| IC3 | 74LS93 |
| IC4 | AY-3-8912 |
| IC5 | TDA2822 | <br> Miscellaneous <br> SK1 50 way transition header with ribbon cable and Amstrad connector. <br> SK2 $\quad 3.5 \mathrm{~m}$ stereo jack socket <br> P.C.B., available from the EE PCB Service, order code EE715, screened lead for audio connections; suitable case, minimum size, $120 \times 95 \times 45 \mathrm{~mm}$; control knob.}



Fig. 2. P.C.B. for the sound generator.


Table 1. Register Functions

| Register | Function | Range |
| :---: | :--- | :--- |
| 0 | Channel A frequency, fine tune | $0 \rightarrow 255$ |
| 1 | Channel A frequency, course tune | $0 \rightarrow 15$ |
| 2 | Channel B frequency, fine tune | $0 \rightarrow 255$ |
| 3 | Channel B frequency, course tune | $0 \rightarrow 15$ |
| 4 | Channel C frequency, fine tune | $0 \rightarrow 255$ |
| 5 | Channel C frequency, course tune | $0 \rightarrow 15$ |
| 6 | Noise frequency control | $0 \rightarrow 31$ |
| 7 | Not mixer enable | $0 \rightarrow 255$ |
| 8 | Channel A amplitude (15 loud. I6 envelope) | $0 \rightarrow 16$ |
| 9 | Channel B amplitude (IS loud. 16 envelope) | $0 \rightarrow 16$ |
| 10 | Channel Camplitude (15 loud. 16 envelope) | $0 \rightarrow 16$ |
| 11 | Envelope period, fine tune | $0 \rightarrow 255$ |
| 12 | Envelope period, course fune | $0 \rightarrow 255$ |
| 13 | Envelope shape cycle | $0 \rightarrow 15$ |

Table 2

| reg. value | envelope cycle shape |
| :---: | :---: |
| 0 | $I$ |
| 4 | $1$ |
| 8 |  |
| 10 |  |
| 11 | $\sqrt{ }$ |
| 12 |  |
| 13 |  |
| 14 |  |

mal. If the computer will not boot up. or does anything abnormal. turn the computer off. unplug the unit and check all wiring. joints etc. For test program see PROGRAMMING.

## REGISTER <br> FUNCTIONS

The AY-3-8912 is register based. this means that you need to set the values of the registers (like memory locations) in the 8912 to certain values. depending on what you want it to do, the register functions are shown in Table I, and explained below.
FREQUENCY CONTROL:
The frequency of each channel is determined by setting registers 0 to 5 . the frequency of the output is equal to: $125000 \div$ (FTV + $256 \times$ CTV)
where FTV is the FINE TUNE VALUE and CTV is the COURSE TUNE

## VALUE

NOISE CONTROL:
Noise can be mixed into any or all of the channels for special effects (drums. guns. waves, etc.)., the base frequency of the noise is given by:
$125000 \div$ (VALUE OF NOISE

## REGISTER)

## NOT MIXER ENABLE:

The value of this register controls which frequencies are "switched on" and whether the channels have noise mixed into them.


Fig. 3. Polarising key positioning in the Amstrad connector.

## NOT NOISE ENABLE

## NOT FREQUENCY ENABLE

| CHANNEL | C | B | A | C | B | A |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| VALUE | 32 | 16 | 8 | 4 | $?$ | 1 |

The value you need to set this port to enable the required things can be worked out by $255-\left(\mathrm{V}_{1}+\mathrm{V}_{2}+\ldots+\mathrm{V}_{\mathrm{n}}\right)$ Where $V_{1}$ to $V_{n}$ are the values of the things you wish to enable. e.g. 255 $(16+4+2+1)$. This will enable all of the frequencies and enable noise to be mixed with channel $B$.

## AMPLITUDE CONTROL:

The value of these registers control the amplitude of the channels. 0 is QUIET. 15 is LOUD, setting the register to 16 lets the amplitude of that signal be controlled by the envelope generator.

## ENVELOPES:

As mentioned earlier the AY-3-8912 has envelope capability, but what is an envelope? An envelope is a waveform that varies the amplitued of a signal (in this case the channels that have the amplitued control set to 16 ), another way of explaining this is that it is similar to turning the volume control on a radio up to full and then down again then up to full etc. The speed at which you turn the volume down (or up) is the envelope period, this can be varied from approx. 7812 Hz to about 0.12 Hz ( $\cong 8.4$ second period).

## ENVELOPE PERIOD.

These registers determine the frequency of the envelope cycle which is given by: $7812 \div(256 \times$ ECT + EFT $)$
Where ECT is the ENVELOPE COURSE TUNE value and EFT is the ENVELOPE FINE TUNE value.

## ENVELOPESHAPE:

This controls the shape of the envelope cycle (see Table 2).

## PROGRAMMING

Program I selects random registers and then sends random data to the selected register. the effect is a jumble of frequencies. noise, etc. from the headphones, so this is a good program to test the sound generator.
If you wish to write your own programs you must include lines 100 to 140 as these set up the PIO and also reset the sound i.c. Line 220 is a subroutine to select a register. the value of which is in A. Line 240 is similar to 220 but sends data (in A) to the previously selected register.
Program 2 is almost the same as Program 1, but instead of choosing random registers and data they are input by the user.


## TEST PROGRAMS

## PROGRAM ONE

PROGRAM TWO
100 PAD=160: $\mathrm{PAC}=161: \mathrm{PBD}=162: \mathrm{PBC}=162: \mathrm{REN}$ PAD=PORT A DATA 120 OUT PAC, 15:OUT PBC, 15: REM SET PIO TO OUTPUT MODE

140 OUT PAD, 0 : REM ${ }^{\circ}=$ RESET SOUMD CHIP ${ }^{\circ}+{ }^{*}$
$160 \mathrm{~A}=\mathrm{INT}($ RND 15 ) : PRINT "Register:";A|:GOSUB 220
$180 \mathrm{~A}=$ INT (RND*255): PRINT "Value:";A :GOSUB 240
200 GOTO 160

220 OUT PAD, 4: OUT PBD, A:OUT PAD, 7: RETURN
230 REK *WRITE DATA (A) TO LAST REGISTBR SELECTED**
240 OUT PAD. 4 : OUT PBD, A: OUT PAD, 5: RETURI

```
100 PAD=160: PAC=161:PBD=162:PBC=102 :REK PAD=PORT A DATA
120 OUT PAC, 15:OUT PBC,15: REK SET PIO TO OUTPUT YODE
140 OUT PAD.0 :REM *?*RESET SOUND CHIP****
160 INPUT "Register:". A:GOSUB 220
180 INPUT " Value:",A :GOSUB 240
200 GOTO 160
```



```
220 OUT PAD, 4:OUT PBD,A:OUT PAD, 7: RETURN
230 REK **VRITE DATA (A) TO LAST REGISTER SELECTED**
240 OUT PAD, 4;OUT PBD, A&OUT PAD, 5: RETURN
```

Program 3 shows how noise can be used to create a drum effect, channel $B$ is set-up so that the tone is disabled and the noise is enabled, the amplitude register is set to 16 (amplitude controlled by envelope generator). The effect of this is that this channel has plain "noise" on it, this is modulated by envelope cycle shape 8 (amplitude starts at 15 and fades to 0 . this repeats until another envelope is defined or the i.c. is reset). This creates a (snare) drum beat.

If the tone on channel B is also enabled. and set to a high frequency. the drum will sound more metallic.
The next part of the program reads the data for notes from lines 280 onwards, until it finds a data statement with 999 in it, at which point the data is read from line 280. and repeats.
The data is in the form:- NOTE $A$. VOLUME A. NOTE C. VOLUME C. PAUSE
Where NOTEA.C is the frequency of the note (see Table 3).
VOLUMEA.C is the volume of the channel (from 0 to 15).
PAUSE is the duration of this note.
The sound i.c. cannot use the frequency information directly, so it is converted at line 200 (and 210).
Frequencies can be converted into time periods for the sound i.c. by:

FTV $=(125000 /$ frequency $)$ AND 255
$C T V=\operatorname{INT}(125000 /(256 \times$ frequency $))$
where FTV is the fine tune value. CTV is the course tune value. Frequency is in Hertz.
The minimum frequency possible is 31 Hz and the maximum is 125 kHz .
Program 4 is a simple piano program. use keys I to 7 to play notes. To make the piano sound slightly more realistic the note is modulated by an envelope similar to the drum in Program 3, although this only happens once (not repeatedly like the drum). This sound takes about a second to die away, unless another key is pressed first, in which case the new note will start immediately. Envelope cycle shape 0 is used for this, and must be initalised directly before the note is programmed into the i.c. (line 200).

Program 5 is a sound effect (that sounds quite like a plane diving and then exploding!). This shows the effect of mixing noise with a tone, if the noise is turned off you would just hear a "scale". but with the noise the sound is a lot more realistic.
The explosion at the end is created by turning off the tone, and then fading out the noise by using envelope cycle shape 0 .


## PROGRAM THREE


110 PAD=160: $P A C=161: P B D=162: P B C=163:$ OUT PBC, 15: OUT PAC. 15
120 OUT PAD, O: REM **RESET SOUND CHIPEs
130 REM **SET CHANNEL B AS DRUM BEAT(NOISE \& ENVELOPE ORLY)***
140 RESTORE: FOR $\mathrm{V}=1$ TO 5: READ R, A: GOSUB 1010: NEXT
$150 \mathrm{R}=7$ : $\mathrm{A}=234$ : GOSUB 1010
160 RESTORE 280

180 READ NA, VA, NB, VB, P:IF NA $=999$ THEN 160
190 REM * CHANGE FREQUENCY INTO PERIOD, AND SEND TO SOUND CHIP**
$200 \mathrm{R}=0$ : $\mathrm{A}=(125000!/ \mathrm{NA})$ AND 255; COSUB 1010: $\mathrm{R}=1: \mathrm{A}=\mathrm{INT}(1250001 /(256$ / NA ) ) : GOSUB 1010 $210 \mathrm{R}=4, \mathrm{~A}=(1250001 / \mathrm{NB})$ AND 255;GOSUB 1010:R=5:A=INT(125000:/(256/NB)):GOSUB1010 220 PEK teme
$230 \mathrm{R}=8: \mathrm{A}=\mathrm{VA}:$ GOSUB 1010: $\mathrm{R}=10: \mathrm{A}=\mathrm{VB}$ : GOSUB 1010
240 FOR $N=1$ TO P: NEXT:GOTO 180

260 DATA $9,16,6,5,11,30,12,13,13,8:$ REM VOL $=16$, NOISE $=5$, ENY $=13256+30$, ENVSHAPE

280 DATA $156,15,156,15,40,175,15,175,15,30,165,15,165,15,40,156,15,156,15,50$ 290 DATA $175,15,195,15,60,185,15,175,15,60,175,15,185,15,60,185,15,175,15,60$ 300 DATA $175,15,195,15,40,195,15,175,15,40,195,15,233,15,40,220,15,220,15,40$ 310 DATA $999,0,0,0,0$
320

1010 OUT PAD, 4:OUT PBD, R:OUT PAD, 7
1020 REM *EE**WRITE DATA (IN A) TO LAST REGISTER SELECTEDFE\#\#* 1030 OUT PAD, 4, OUT PBD, A: OUT PAD, 5: RETURN

## PROGRAM FOUR

 20 REM *BASIC NOTE FREQUENCY IS FADED OUT SLOVLY BY ENVELOPE** 30

$110 \mathrm{PAD}=160: \mathrm{PAC}=161: \mathrm{PBD}=162: \mathrm{PBC}=163: \mathrm{OUT} \mathrm{PBC}, 15:$ OUT PAC, 15
110 PAD $=160: \mathrm{PAC}=161: \mathrm{PBD}=162: \mathrm{PBC}=163:$ OUI $\mathrm{PBC}, 15$
120 OUT PAD, O: REM
$\begin{array}{ll}120 & \text { OUT PAD, 0: REM } \\ 130 & R=9 \\ \text { : } A=16: G O S U B & \text { 280: REM SET VOLUNE OF B TO BE CONTROLED BY ENVELOPE }\end{array}$
$130 \quad R=9: A=16:$ GOSUB 280: REM SET VOLUNE OF B TO BE CONTROLED
$140 \quad R=11: A=0$ :GOSUB 280: REM SET ENVELOPB FINE TUNE PERIOD
$140 \mathrm{R}=11: \mathrm{A}=0$ : GOSUB 280: REM SET ENVELOPE FINE TUNE PERIOD
$150 \mathrm{R}=12: \mathrm{A}=28:$ GOSUB 280: REM SET ENVELOPE COURSE TUNE PERIOD
$150 \mathrm{R}=12: \mathrm{A}=28:$ GOSUB 280: REM SET ENVELOPE COURSE TU
160 $\mathrm{R}=7: \mathrm{A}=253$ : GOSUB 280; REM ENABLE CHANNEL B ONLY
$160 \mathrm{R}=7: \mathrm{A}=253$ :GOSUB 280: REM ENABLE CHANNEL B ONLY
$170 \mathrm{DIM} \mathrm{F}(7): F O R \mathrm{~N}=1$ TO 7:READ $\mathrm{F}(\mathrm{N}):$ : REXT: REM EREAD NOTE FREQUENCIES*
170 DIM $F(7): F O R$ N=1 TO 7: READ F(N): NEXT: REM ©RE
180 AS=INKEYS:IF A\$="" THEN 180 ELSE NA=VAL (AS)
180 A $\$=I N K E Y S: I F A \$="$ THEN 180 ELSE NA=VAL
190 IF NAくI OR NA>7 THEN 180 ELSE NA=F(NA)
190 IF NA<I OR NA>7 THEN 180 ELSE NA=F (NA)
$200 \mathrm{R}=13$ : $\mathrm{A}=0$ : GOSUB 280: REM © 2 SET ENVELOPE CYCLE TO SHAPE O
210 REM *CHANGE FREQUENCY INTO PERIOD, AND SEND TO SOUND CHIP \#
$220 \mathrm{R}=2: \mathrm{A}=(125000$ :/NA) AND $255: \operatorname{GOSUB} 280: \mathrm{R}=3: \mathrm{A}=\mathrm{INT}(125000$ :/(256*NA)): GOSUB 280 230 GOTO 180

250 DATA $220,247,262,294,330,349,392$
260
270 REM **E*********SELECT REGISTER (IN R)****************
280 OUT PAD, 4: OUT PBD, R: OUT PAD, 7
290 REM ****URITE DATA (IN A) TO LAST REGISTER SELECTED***** 300 OUT PAD, 4 : OUT PBD, A: OUT PAD, 5: RETURN

Table $\mathbf{3}$ Notes and Frequencies

## NOTE FREQUENCY

| A | 110 | .220 | .440 | .880 |
| :--- | :--- | :--- | :--- | :--- |
| B | 123 | .247 | .494 | .988 |
| C (MIDDLE) | .523 (HIGH) | .1047 |  |  |
| D | 131 (LOW) | .262 (MID | .587 | .1175 |
| E | 147 | .294 | .659 | .1319 |
| G | 165 | .330 | .698 | .1397 |

This table gives the approximate frequencies of the notes, though this is not the entire range of frequencies a vailable.

## PROGRAM FIVE

10 REM *******PLANE FALLING OUT OF SKY AND EXPLODING ***
 30

$110 \mathrm{P} \wedge \mathrm{D}=160: \mathrm{P} \wedge \mathrm{C}=161: \mathrm{PBD}=162: \mathrm{PBC}=163$ : OUT PBC, 15: OUT PAC, 15
120 OUT PAD, O: REM RESET SOUND CHIP
$130 \mathrm{R}=9$ : $\mathrm{A}=9$ : GOSUB 260: REM SET VOLUME OF B TO 9
$140 \quad R=11: A=0$ : GOSUB 260: PEM SET ENVELOPE FINE TUNE PBRIOD
$150 \mathrm{R}=12$ : $\mathrm{A}=180$ : GOSUB 250:REM SET ENVELOPE COURSE TUNE PERIOD
$160 \mathrm{R}=7$ : $\mathrm{A}=237$; GOSUB 260: REM ENABLE CHANNEL B TONE AND NOISE
170 FOR NA=100 TO 450:REM FREQUENCY OF NOTE IMCREASES FROM 100 Hz TO $450 \mathrm{H} z$
180 REM *CHANGE FREQUENCY INTO PERIOD, AND SEND TO SOUND CHIP **
$190 \mathrm{R}=2: \mathrm{A}=(1250001 / \mathrm{NA}) \mathrm{AND} 255: \operatorname{GOSUB} 260: \mathrm{R}=3: \mathrm{A}=\mathrm{INT}(125000!/(256 \mathrm{NA}))$ : $\operatorname{COSUB} 260$ $200 \mathrm{FOR} \mathbb{T}=1$ TO 40: NEXT N, NA

$220 \mathrm{R}=9$ : $A=16$ : GOSUB $260: R=13$ : $A=0$ : GOSUB $260: R=7$ : $A=239$ : GOSUB 260
230 END
240

250 OUT PAD, 4 : OUT PBD, R: OUT PAD, 7
270 REM *****VRITE DATA (IN A) TO LAST REGISTER SELECTED*****
280 OUT PAD, 4 : OUT PBD, A: OUT PAD, 5: RETURN

## SOUNDS

The sounds made by musical instruments have quite complex waveforms, which this unit is not capable of generating, but with the use of envelopes the sounds can be
made slightly less digital (as with the piano program), and simple tunes can be played. Sound effects for cars. explosions. drums. etc. are a lot more realistic if they have noise mixed in with them rather than just the basic frequency.

When writing your own program follow this list..

1) Make sure the first thing you do is to set-up the PIO and reset the sound chip (lines 110 and 120 ).
2) Include lines 1010 to 1030 on your program, when you want to change the contents of a register set $R=$ REGISTER $N \cdot A=$ NEW VALUE OF REGISTER. GOSUB 1010. On return register $R$ will be set to value $A$.
3) Set the volumes of the channels of the channels you wish to use (and any envelope delays, noise, or frequencies that you need).
4) Enable the required channel (register 7).
5) Remember that if you want to use frequencies (the sound i.c. uses tone periods). to use the conversion program (at line 200). The first R is FINE TUNE VALUE Reg. $(0=A, 2=B$. $4=C$ ). The second $R$ is COURSE TUNE Reg. $(1=A, 3=B, 5=C)$, and $N A=$ frequency of required note (in Hz ).

NOTE: All line numbers above refer to PROGRAM 3.

with David Barrington

## Kit News

The improvement in the quality, range and professional approach adopted by kit/module advertisers over the years is highlighted by a new audio company this month.

After an extensive search of the overseas manufacturers, Platinum Audio (0273 685904) has been set up to import a range of audio products in kit form. Ranging from simple preamplifiers to a 300 W slave amplifier, all kits come complete with any necessary heatsinks/hardware.

Talking of amplifiers, we see that the latest 1990/91 Velleman catalogue from High-a Electronics (0707 263 562) contains details of a "valve" amplifier that is claimed to give up to 200 W music power output. The catalogue contains approximately 30 additions to their extensive range of kits.

## CONSTRUCTIONAL PROJECTS

## Simple Intercom

On checking out the prototype model for the Simple Intercom, this month's back-up project for the Teach-In '91 series on "Design Your Own Circuits," we have found that the loudspeaker used is a $31 / 2 \mathrm{in}$. diameter type. This speaker has a 35 ohm speech coil and was purchased from RS Components through Electromail 0536 204555), their mail order operation.

Although the speaker is stamped with the reference 9L957, looking through the catalogue it can only be from their general purpose range and be the one coded 248-274 (£4.78). Provided the speaker has a coil rated at 35 ohm , and it
will fit inside the chosen case, practically any reasonable size speaker will be suitable for this simple circuit.

The printed circuit board for the intercom is available from the $E E P C B$ Service, code 719. For those wishing to build up the two demonstration amplifiers contained in this month's Design Your Own Circuits teach-in, the General Purpose Transistor Amplifier (EE717) and the Operational Amplifier (EE718) boards are also available from the EE PCB Service, see page 68.

## Analogic Test Probe

The interlocking plastic case for the Analogic Test Probe is only available from Magenta Electronics. The rest of the components seem to be standard lines and should not cause any local sourcing problems.

A full kit of parts including the printed circuit board, drilled case, but excluding battery, is available from Magenta Electronics. Dept EE, 135 Hunter Street, Burton on Trent, Staffs, DE14 2ST 0283 65435) for the sum of $£ 12.95$. Add a further $£ 2$ for posting and packing.

The small printed circuit board for the probe is available through the EE PCB Service, code EE720 (see page 68).

## Amstrad PCW Sound Generator

There should be no purchasing problems when shopping for the components required to build the Amstrad PCW Sound Generator. However, the programmable sound generator chip AY-3-8912 may prove to be in short supply in some local areas. If you do have difficulty locating this device, it is
currently listed by Maplin code RA90X (AY-3-8912). They also stock the 1 W stereo amp i.c.

Most components suppliers stock a fairly good supply of computer parts and should be able to offer the 50 -way header, with ribbon cable and Amstrad connector.

The single-sided printed circuit board is available from the EE PCB Service, code EE715, See page 68 for details.

## Sound Operated Switch

We do not expect any component buying problems to be encountered when building the Sound Operated Switch project. Some component stockists may not carry the 4.7 megohm preset potentiometer and the designer suggests that a 2.2 megohm preset is more readily available and can be used in this circuit.

The crystal microphone inserts are generally available and come in two types of casing, metal or plastic. Because of the high sensitivity of the circuit, and to avoid "stray" pick up, it is best to use a metal case type and "earth" the casing as outlined in the article.

## Spatial Power Display

Once again this month, there should be no nightmares about obtaining components for the Spatial Power Display. they are all standard "of-the-shelf" stock items.

When pricing up this project we found the bulk of the cost was taken up by the l.e.d.s. As we priced them on the high side at $26 p$ each, quite a considerable saving could be made if you purchased one of the "l.e.d. component packs" on offer from Greenweld 0703 236363). In fact, it might be a good idea to approach your supplier about a "quantity" discount price.

The printed circuit board for the l.e.d. display project is available from the $E E$ PCB Service, code EE714 (see page 68).

## Help

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## Robert Penfold



THIS MONTH we will continue with the topic of interfacing to the IBM PCs and compatible microcomputers. Last month we briefly considered the 8 -bit expansion bus. This month the input/output map and address/control bus decoding will be discussed. Note that the 8088 series of microprocessors used in the PCs have separate input/output and memory maps.

As user add-ons will not normally have memory circuits to place in the memory map, they will only fit into the input/output map. Consequently, here we will only be concerned with this method of interfacing, and will not consider units that fit into the memory map.

## Map Reading

As pointed out last month, the 8088 series of microprocessors use the sixteen least significant address lines for mapping input/output devices, but in the PCs only the ten least significant address lines are actually utilized. This gives 1024 usable addresses (from 0 to 1023), rather than the 65536 that the microprocessors are capable of using.

Addresses from 1024 to 65535 are simply echoes of the 1024 base addresses. Obviously this reduced number still leaves plenty of address space for standard peripherals such as printer ports and display cards, with sufficient left over for more specialised add-ons.

Table. 1 shows the input/output map for the PCs, and it will be apparent from this that the entire memory map has been allocated to specific functions. This does not mean that there is nowhere in the map that can accommodate user add-ons.

The lower half of the map is reserved for the motherboard, and addresses from 000 to IFF are definitely out of bounds (and are not included in Table. 1). The addresses in Table. 1 are all in hexadecimal incidentally, and from here onwards, all addresses will be given in hexadecimal form.

If you look at the map you will note that addresses from 300 to 31 F are for prototype cards. Possibly this address space is really intended for testing prototype cards which will actually be located elsewhere in the map in their final form.

However, it represents an obvious and pretty safe area of the map for your own circuits. This section of the map provides some thirty-two input and output addresses, which obviously permits a substantial amount of expansion. Half-a-dozen 8 -bit input/output ports plus a couple of analogue ports would use up only one quarter of the available addresses!

If further addresses are required, there are other parts in the map which could
probably be exploited. In fact any block of addresses which is reserved for hardware which your computer does not actually have fitted is a legitimate section of the map for your own circuits.

Bear in mind though, that this is a slightly unsafe method if you are designing circuits which will be used in computers other than your own PCs. The fact that your computers do not have fixed disks and matching controller cards installed, does not mean that other PCs will lack this equipment.

If you decide to use a section of the input/output map other than the one set aside for prototype cards, then it is obviously better to select one that is reserved for an unusual piece of hardware that will be installed in very few PCs. Something like the address block reserved for SDLC communications represents a safer choice than a section of the map such as the one reserved for the second serial port.

There would appear to be small sections of the map which are not allocated to any purpose, or reserved for future expansion, and these might offer suitable niches for your own add-ons. Where possible, it is probably best to only use the addresses set aside for prototype cards though. Here we will only consider adc'ress decoding for this particular section of t:ie input output map, but the general principles described in this article apply to interfacing using any section of the input/output map.
Table. 1: Input/Output Map Address

| Address | Range |
| :--- | :--- |
| Runction |  |
| -200-20F | Game Port |
| 210-217 | Expansion Unit |
| 220-24F | Reserved |
| 278-27F | Printer Port 2 |
| 2F0-2F7 | Reserved |
| 2F8-2FF | Serial Port 2 |
| 300-31F | Prototype Card |
| 320-32F | Hard Disc Controller |
| 378-37F | Printer Port 1 |
| 380-38C | SDLC Communications, or |
| 380-389 | Synchronous Port 2 |
| 3A0-3A9 | Synchronous Port 1 |
| 3B0-3BF | Mono display/Printer Port |
| 3C0-3CF | Reserved |
| 3D0-3DFF | Colour Graphics Adaptor |
| 3E0-3F7 | Reserved |
| 3F0-3F7 | Floppy Disk Controller |
| 3F8-3FF | Serial Port 1 |

## Echoes

It is perhaps worth mentioning a technique that effectively enables the 300 to 31 F address range to be used several times over, permitting more expansion that anyone is ever likely to need. Suppose that several circuits are all mapped to address 300 , but some or all of address lines A10 to

A15 are decoded so that each device is only activated if these lines are at a certain set of logic levels, with a different set being used for each device.

Normally a device at address 300 would be activated if the base address was used, or if one of the echoes at 700, B00, F00, etc. was accessed. By decoding A10 to A15 it is possible to map devices to the base address or specific echoes of the base address.
Although the computer's normal hardware does not make use of these address lines, they can still be utilized by your own add-ons. It is only echoes of addresses in the range 300 to 31 F that are made available, and most of the addresses in the upper 63 K of the map remain unusable. However, with each of thirty-two base addresses having up to sixty three available echo addresses, these inaccessible parts of the input/output map are not really needed.

## PC Languages

If you are going to experiment with user add-ons, then you will need a programming language that provides access to the registers of the hardware in the input/output map. This may seem an obvious point, but less obviously, not all PC programming languages actually give direct access to the hardware. There is obviously no difficulty with assembly language or machine code, where the microprocessor's full range of input and output instructions are available.
Most BASICs provide access to the hardware via the INP function and the OUT instruction, or an equivalent to these. Most PCs are supplied complete with Microsoft's GW BASIC, which has suitable commands available. Many other BASICs for the PC also provide easy access to the hardware.
There are exceptions though. In particular, the BASIC 2 that was supplied with my Amstrad PC1512 seems to offer no easy means of directly accessing the hardware. I could be overlooking something, but it would seem that Amstrad PC users will need to invest in an alternative programming language if they wish to dabble with home constructed add-ons.
The situation with non-BASIC high level languages for the PC seems to vary from one language to another, or even from one dialect to another. Often access to the hardware registers is possible, but you might need to delve deep into the manuals in order to discover precisely how it is achieved. In some cases it might only be possible via specially written machine code routines which the main programming language can call up and run, as and when necessary.

Although GW BASIC is not exactly a state of the art PC programming language, it is quite a good one when experimenting with home constructed add-ons. It provides very easy access to the add-ons, and for most purposes it is fast enough. A compatible compiled BASIC is a good alternative where higher operating speed is required.

## Address Decoding

On the face of it there are a substantial number of lines that must be decoded by any user add-on. For a start, there are the ten least significant address lines (AO to A9). There are also numerous control lines on the expansion port.

In practice, matters are not quite as difficult as it might at first appear. It is only necessary to decode all ten of these address lines if a device is to be placed at a single address. Often this will not be the case, and some simplification will be possible.
In an extreme example you might only wish to add one input or output port, and this could then occupy the base address plus echoes at the other addresses up to 3IF. This would avoid the need to decode the lower five address lines ( $A O$ to $A 4$ ).
The five upper address lines (A5 to A9) must always be fully decoded though. Table. 2 shows the states of these lines when a device in the address range 300 to 31 F is activated.

| Table. 2: Address Line |  |
| :---: | :---: |
| Condition |  |
| Address Line | Logic State |
| A5 | LOW |
| A6 | LOW |
| A7 | LOW |
| A8 | HIGH |
| A9 | HIGH |

Although there are quite a number of control bus lines available on the expansion port, for most purposes the vast majority of these can simply be ignored. There are some which must be decoded though. The input/out read and write lines (IOW and IOR) certainly fall in this category.

Input devices must only be activated if IOR is low, and output circuits must only be activated if IOW is low. Therefore, at
least one of these lines must always be decoded, and for an input/output device they must both be decoded

In most cases the only other control line that must be decoded is AEN, which goes low when a processor bus cycle (rather than a DMA type) is in progress. Any user add-on, whether it is an input or an output type, must therefore only be activated when $A E N$ is low.

The most fundamental decoding that will provide usable results therefore consists of having the add-on activated when A8 and A9 are high, and AEN plus A5 to A7 are low. Additionally, input devices should only be activated when IOR is low, and output devices should only be activated when IOW is low.

This means that a practical add-on might need to decode as few as seven or eight lines. Of course, if the add-on occupies several addresses, then further address lines will need to be decoded, possibly taking the number of lines to be decoded as high as twelve.

## Address Decoder

Although the PCs operate at relatively high clock frequencies, ordinary 74LS or 74 HCT logic gates etc. are quite fast enough to provide address decoding, latch output data, and provide the other usual interfacing tasks. Using gates, inverters, 3 to 8 line decoders, etc., there are numerous ways of providing any desired address/control bus decoding. Fig. 1 shows a simple method using a 8 -input NAND gate and a hex inverter.
The output of IC2 will go low if all eight of its inputs are taken high. Obviously in this case we require a decoded output that assumes the active state when certain inputs are high and others are low, not when they are all high.
However, simply adding an inverter ahead of an input effectively makes an active low type instead of an active high type. In this case only A8 and A9 will be high when the add- on is accessed, and so all the other lines drive IC2 via inverters.

IC2 provides a negative output pulse if any address in the range 300 to 30 F is read. Slightly less than the minimum of address decoding is used here, since A4 is decoded.

If the inverter on this line is omitted, then the circuit will be activated when any address from 310 to 31 F is read.

In practise it is a good idea to at most only use half the address space for your first PC add-on, so that you have some address space left for any subsequent DIY cards you may wish to add. Note that by decoding IOW instead of IOR the circuit will be activated by write operations instead of read types.

## Separate Outputs

Separate decoded read and write outputs will often be needed. Using entirely separate circuits to provide these outputs is definitely doing things the hard way. Also, ideally any add-on should not load the bus by more than one LS TTL input per line.

A better way of providing separate read and write outputs is to first decode AEN and the appropriate address lines, and to then process the decoded output with the IOR and IOW signals using a couple of OR gates. Fig. 2 shows how the two OR gates are used.

The output of an OR gate is only low if both inputs are low. The output of IC1a therefore goes low if there is a read operation to the decoded address range, while IC2b's output goes low if there is a write operation to this address range.

The circuit of Fig. 1 can still be used to provide the main decoding, but pin 12 of IC2 is tied to the +5 volt rail. Alternatively, it could be used to decode A3, and this would permit up to four cards occupying eight addresses each to be used.
If a standard $80^{* *}$ bus interface chip is used (such as the 8255), this will have inputs for IOR and IOW, making it unnecessary for the address decoder to process them. Most of these chips have several registers, with the desired one being selected via register select inputs. These would normally be driven from the least significant address lines of the processor.

NEXT MONTH: A Simple Input/Output Port will be described, and we will look at the potentially awkward mechanical side of PC interfacing.


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# ROBOTROUNDUP 

 Nigel Clark
## SUCCESSFUL CHAOS

Assemble a group of robot enthusiasts from all over the world, asking them to bring along their latest developments, and you have a major ingredient for chaos. Add a large amount of flexibility to allow for the expected breakdowns and the fact that a similar event has never been held before and any hopes that you will achieve any sort of smoothly running organisation can be thrown out. In a nut shell you have the first Robot Olympics held at the Turing Institute, Glasgow in September.

However with 60 robots from 12 countries and a large amount of interest from spectators and the Press it could be considered a success.
"It has taken me three weeks to recover," said Dr Peter Mowforth, who organised the event. "It was much bigger than I had expected and considering that there could not be a lot of advance planning it went very well. We were carried along by the enthusiasm and the motivation of the people involved."

He added that on one of the days 1,500 school children visited the Olympics and showed a great deal of interest.
"If just a few of them decided to become electronics engineers rather than accountants or lawyers then it will have all been worth while."

The organisers are now setting out rules and regulations as a basis for future Olympics, the next of which will be held in two year's tine and every two years after that. It is intended to hold the event in Glasgow every four years. Bids from organisations interested in holding the next competition are being sought and there has already been an enquiry from Nagoya in Japan.

## CONTESTS

Mowforth said that the first event had had to be flexible because they did not know what or who to expect. However with the experience gained they would be better prepared in the future. The flexibility had continued throughout the Olympics. The different contests were not set up until the organisers knew which robots were available. On registration entrants filled in the capabilities of their machines and wherever there were capabilities in common a contest was created.

This resulted in eleven sections in all, wall climbing, two-legged races, races for more than two-legged, javelin, talking, obstacle avoidance, phototrophic, pole balancing, manipulation, behaviour and wall following. There were also a number of special awards, see below.

Mowforth said that another reason for this approach had been that he thought existing international robotic competitions did not help the general advancement of robotics as they had too narrow a specification which kept robot builders in a very narrow niche. He mentioned as specific examples Dr John Billingsley's Micro-Mouse and Robot Ping Pong competitions.

However Dave Buckley of the Shadow Group thought that the approach had resulted in a series of competitions which were not challenging enough to encourage the development of robot technology. He cited the success of a number of simple devices or ones which had been around for a few years. The most obvious example was the 19th century Japanese archer, brought along by the Museum of Automata in York, which won the javelin event.


Obstacle Avoidance

1. ASTERIX Toronto University, Canada. Anthony Green, Pavel Rozalski
2. OSCAR Artificial Intelligence Department, Edinburgh University. Scotland
3. YAMABICO Tsukuba University, Japan. Shouji Suzuki

Pole Balancing

1. SALFORD PENDULUM Salford University, England. F Nagy, G A Medrano-Derda
2. LANKY Lancaster University, England

Phototrophic

1. ALPHA PHOTON Kent University, England. David Bisset
2. ICARUS The Shadow Group, London, England. D Buckley

Manipulators

1. BELGRADE/USC HAND University of Belgrade, Yugoslavia
2. BCI St Patricks High School, Coatbridge, Scotland

## Biped Race

1. CARDIFF BIPED University of Wales, Cardiff. Paul Channon, Simon Hopkins, Prof Pham
2. ROBBIE Paisley College of Technology. Scotland. Ken MacFarlane, Gordon Allan

## Javelin

1. YORK ARCHER Museum of Automata, York, England
2. WILBERFORCE East London Polytechnic, England. Martin Smith
3. ELEPHANTS TRUNK Heriot-Watt University, Edinburgh, Scotland. J BC Davies, J Morrison

## Multi Legged Race

1. PENELOPE Edinburgh University, Scotland, D J Todd
2. GENGHIS Massachusetts Institute of Technology, AI Lab, USA. Olaf Beck, Prof Rodney Brookes, Colin Angle

## Wall following

1. YAMABICO Tsukuba University, Japan. Shouji Suzuki
2. SAM Kent University, England. David Bisset, Jason Garforth, Jeremy Laycock

## Talking

1. RICHARD 1st The Turing Institute, Glasgow, Scotland. Ketil Undbekken, Peter Mowforth
2. SHADOW WALKER The Shadow Group. London, England. D Buckley

## Wall Climbing

1. ZIG ZAG Portsmouth Polytechnic, England. A A Collie, Prof J Billingsley, R. P Smith
2. RVP // Soviet Academy of Sciences, Moscow. USSR. Prof Chernousko, Prof Gradestsky
3. RVP / Soviet Academy of Sciences, Moscow, USSR. Prof Chernousko. Prof Gradetsky

## Behaviour

1. GHENGHIS Massachusetts Institute of Technology, AI Lab, USA. Olaf Beck, Prof Rodney Brookes, Colin Angle
2. SHEEP AND SHEEP DOG Computer Science Department, Strathcylde University, Scotland. Peter Barrie, Gerry Haran
3. SIAS City Montessori School, India. Mr Ashish Panwar

## IEEE ROBOTICS \&

AUTOMATION SOCIETY YOUNG ROBOTICIST AWARD
BC1 Brian Carr (School pupil). St Patricks High School, Coatbridge, Scotland.
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NATWEST BANK PRIZE FOR
TECHNOLOGY TRANSFER
ATTILA Massachusetts Institute of Technology, Al Lab, USA. Olaf Beck, Prof Rodney Brookes, Colin Aingle.
Awarded with a Caithness Crystal
Bowl and 1200 from NatWest
BEST SCHOOL AWARD
XYBOT Inverkeithing School, Class 7S.
Presented with a Cup from the Turing
Institute and 1100 special prize from NatWest

## OLYMPIC CHAMPION

YAMABICO Tsukuba University. Japan. Shouji Suzuki.
Awarded with a Caithness Glass
Trophy

While accepting that not all the entries were at the forefront of technology Mowforth pointed out there was much to be admired particularly in Yamabico. from Tsukuba University in Japan, which was named Olympic Champion. The basis for making this choice was the quality of the electronic and mechanical building, the sophistication of the resulting behaviour and its novelty.

Earlier in the contests the Japanese machine had won the wall following and came third in the obstacle avoidance. Problems with pronunciation meant it was not placed in the talking contest but that did not prevent it making a small speech in Japanese thanking the humans for inviting it to Scotland.

## ZIG ZAG DISQUALIFIED

Mowforth was also impressed by Zig Zag from Portsmouth Polytechnic. The machine is the latest development of the work of Arthur Collie which began with his cockroach-inspired six-legged walking robot Robug. Zig Zag won the wall climbing contest easily against less sophisticated machines from the Soviet Academy of Sciences in Moscow. The contest might have been closer if Collie's more complex Robug II had not been disqualified for stepping on one of the Russian robots. However there was no international incident the transgressor being dealt with quickly and firmly. making it the first machine to be disciplined in a Robot Olympics.

Collie was not too unhappy because this unexpected event allowed him to show the robustness of the technology


Zig-Zag from Portsmouth Polytechnic won the wall climbing event.
used in his series of mobiles. In disentangling Robug II the pneumatic suckers were switched off and, due to a misunderstanding, the person holding the safety rope also let go, resulting in Robug falling three metres to the floor. However it was able to get up and work without any ill affects.
"The construction is so compliant that its operation does not depend on it being mechanically perfect," said Collie. "It could work even if one of its limbs was broken or bent.

He could also claim to be the only person at the event who had a commercial product and had sold his first robot. Called NERO, nuclear environment robot operative - anything for a good acronym - it will be used for working on walls inside nuclear installations. Collie is so confident of the technology that he is sure that a wall-climbing robot could be adapted to arij customer's requirements as it was po:isible to build something which could sense objects in its path or an uneven surface and check that each of its suckers had got a proper hold.

## OVERDOSE

All the expected names were at the Robot Olympics, MIT and various university and polytechnic departments as well as schools and enthusiastic amateurs TAG of Northumberland took some machines but they were not working and the Shadow Group took its walker but it overdosed on voltage and lost the use of its chips. Replacements were sought but the correct ones could not be found.

It is good news that the concept is to be continued. While accepting that all the technology was not the most up-to-date state-of-the-art, there was plenty that was. It is also worthwhile being reminded that new does not always mean better.

The only concern must be that there will be fall-off in interest by the media and thus a shortage of sponsors. Newspapers and television quickly lose interest if they do not see their expectations of a robot servant being realised


## OWEN BISHOP

## A versatile sound switch with multiple uses. It will also update the Phoney Phone (Aug ' 9 II to a phone-bell repeater or baby alarm.

This circuit was developed for use with the Phoney Phone, published in our August 1990 issue. Although the Phoney Phone was originally intended as a device for amusement, it has several serious applications. Among these are its use as a phone-bell repeater and as a baby alarm or alarm for an invalid or an elderly person.
A phone-bell repeater is located in a room such as a workshop, or perhaps in the summer house, where the normal telephone bell cannot be heard. The repeater sounds whenever the phone-bell rings.
To set up such a system we need the Sound Trigger, which is placed close to the telephone and connected to the Phoney Phone which is placed in the workshop or other remote area. When the normal telephone bell rings, its sound activates the Sound Trigger which in turn sets the Phoney Phone ringing.
As a baby alarm, the Sound Trigger is in the bedroom, beside baby's cot, and the Phoney Phone is located in the living room. The trigger is extremely sensitive, so even the faintest whimper will activate it. In turn, the trigger sets the Phoney Phone ringing. Its use as an alert for an invalid or elderly person is similar.
The sensitivity of this device can be set so high that it could also be used as an intruder detector - provided that the ambient noise level is normally low. The device also has other applications as a sound-operated trigger with logic output.

## HOWIT WORKS

The system block diagram Fig. 1 shows that the design is quite straightforward. A cheap crystal microphone insert has its output amplified in two stages by two operational amplifiers. The second of these has provision for adjusting the sensitivity.

Once the bistable has been triggered (set) it remains in that condition until it is reset by pressing the Reset button. This means that the Phoney Phone continues to ring until you go to the real phone and press the button on the Sound Trigger. Of course, you could just switch off the Phoney Phone, but that would be cheating!
Also, when used as a baby/invalid/ intruder alarm, a single sound of even very short duration (such as a cough or a click of a door-latch) makes the Phoney Phone ring until the trigger is reset. This is a


Fig. 1. Block diagram for the Sound Trigger.

In the absence of sound, the output of the second amplifier is only about IV, which counts as logical "low" to the logic circuit that follows it. When a sound is detected, the amplifier output rises to logical "high"
The logic circuit is a bistable. It requires a normally low input from the amplifier, and is triggered when the input goes high. The output of the bistable is normally high. but goes low when the bistable is triggered. This mode of operation is compatible with the Phoney Phone circuit.

Fig. 2. Complete circuit diagram for the Sound Trigger.

that needs to be considered when installing the device and adjusting sensitivity, as will be explained later.

## DESCRIPTION

The complete circuit diagram for the Sound Trigger is shown in Fig. 2. The circuit is intended to be driven from the power supply of the Phoney Phone, which operates at 12 V . At this voltage the Sound Trigger requires only 2 mA when quiescent. However, it can also be operated at $5 \mathrm{~V}, 6 \mathrm{~V}$ and 9 V .
Its output is compatible with CMOS at any of these voltages. It can also drive one LS gate if used with TTL at 5 V . The split supply for the operational amplifiers is obtained by the potential-dividing resistors R1 and R2.
The TL064C, ICI, contains four operational amplifiers, of which we use two connected in the non-inverting mode. The gain for the first amplifier is about 1000 . The signal is further amplified by the second amplifier which becomes saturated when even a small sound is detected
Its output swings rapidly between the voltages of the positive and negative supply rails. When it swings to the positive rail voltage, this gives the equivalent of a logical high.
The bistable is constructed of two NOR gates, IC2a and IC2b. Both inputs to the bistable are normally low. The output from
the second amplifier is normally about IV. so satisfies this requirement. The other input of the bistable is connected to the 0 V rail through resistor R9.

Under these conditions the output of the bistable is high. The output is fed to a third NOR gate IC2c, with the inputs connected so that it acts as an INVERT gate. Its output is low and the l.e.d. DI is off.
When sound is detected, the output of amplifier ICIb goes high, setting the bistable. The bistable output goes low and the l.e.d. comes on. The bistable is reset by pressing switch S .
The difficulty with resetting is that the sound made by the button as it is pressed and released is enough to trigger the circuit. This problem is overcome by the capacitor C3.
When S 1 is pressed, the voltage at IC2b pin 13 rises to high, resetting the bistable and charging the capacitor. When the button is released, the capacitor takes a second or two to discharge through resistor R9. Until it is half discharged, the input to pin 13 remains high and the bistable cannot be permanently triggered. This gives time for the button to be fully released and the vibrations produced to die away.

## CONSTRUCTION

The circuit is assembled on a piece of 0.1 in matrix stripboard, size 16 holes $\times 24$ strips. Details of the component layout and breaks required in the underside copper tracks is shown in Fig. 3.
The prototype was designed to fit into a small "microphone-sized" plastic box. As a result, the circuit layout is slightly cramped. If you are unused to working neatly in a confined area, use a larger box and expand the layout sideways by a few holes on either side of the i.c. sockets.
There should be no problem with assembling the circuit and getting it to work first time. The only point to consider is the possibility of electrical interference from the mains or other sources being picked up by the microphone leads.

If your microphone insert MIC. 1 is cased in metal (usually aluminium), you will probably find that one of its two terminals


Fig. 3. Stripboard component layout and details of breaks required in the

is connected to the case. If so, this is the terminal to be wired to the solder pin at A10. The "earthed" case shields the microphone element from interference.
In the prototype the leads between the board and microphone are only 2 cm long and interference is negligible. But when using temporary leads 16 cm long during testing, interference (mains hum) made the amplifier output oscillate at 50 Hz . If you are planning to connect the microphone with leads more than a few centimetres long, use screened cable with the screen connected to the solder pin at $A 10$.


## Potentiometer

VR1 4M7 skeleton preset, vert.

## Capacitors

$\begin{array}{ll}\text { Capacitors } \\ \text { C1, C2 } & 10 \mu \text { axial elec. } 16 \mathrm{~V} \\ \text { C3 } 3 & 47 \mu \text { axial elec. } 16 \mathrm{~V}\end{array}$
Semiconductors
S1 TIL209 or similar I.e.d.
IC1 TL064C quad low-power op.amp
IC2 CD4001 BE quad 2-input NOR gate

## Miscellaneous

MIC. 1 Crystal mic. insert
S1 Push-to-make push-button Push-to-
switch
Stripboard, 0.1 in matrix, size 24 strips $\times 16$ holes; plastic case (ABS), approx. $82 \mathrm{~mm} \times 53 \mathrm{~mm} \times 30 \mathrm{~mm}$; 14 -pin i.c. socket ( 2 off ); 1 mm solder pins (9 off); Blutack; connecting wire, solder etc.


Fig. 4. Guide to the layout of components inside the small plastic case. The microphone insert sits above IC1 and is held in place with a piece of Blutack.

## TESTING

Commence testing the unit by turning the wiper of VR! to a mid-way position. Switch on the power. The l.e.d. DI usually comes on. Press switch S1 and the I.e.d. goes out - but not if you make a noise at the same time.
The circuit responds to any noise - taps, bangs, whilstles, and speech - made within a radius of a few metres. As soon as the noise is detected the l.e.d. comes on and stays on until $S$ 1 is again pressed.
You may decide that the device is 100 sensitive for your purposes. Other noises made in the vicinity of the Sound Trigger may result in false alarms. The solution is to reduce the sensitivity of the circuit and place it as near as possible to the source of sound (telephone bell, baby, etc).
To reduce sensitivity, turn the wiper of VR1 towards the end of its track nearest resistor R7. If it is still too sensitive, try fastening adhesive tape over the apertures in the microphone case. Alternatively substitute a resistor of lower value for R5

## ASSEMBLY

When the circuit board has been tested successfully, prepare the case by drilling holes in the lid of the box, as shown in Fig. 4. The circuit board is a snug fit in the bottom of the plastic box, and is held there with a piece of Blutack.
The microphone MIC:1 is located immediately on top of ICI. If the microphone case is made of metal, stick insulating tape on its rear surface to prevent possible contact with the circuit. A piece of Blutack on IC1 helps to hold the microphone in place.
You may find that the microphone comes into contact with the lid when this is screwed in place. Three small blobs of Blutack spaced around the circumference of the microphone help to keep it acoustically insulated from the lid of the box and thus less subject to the noise from the push-button.

## LINKING TO THE PHONEY PHONE

The Sound Trigger is connected to the Phoney Phone (Aug. ${ }^{90}$ ) by a 3 -core lead. The positive and negative supply leads are
connected to the 0 V and 12 V lines of the Phoney Phone circuit. The third line connects the output pin of the trigger to the control input point (CP) of the Phoney Phone.


Fig. 5. Fitting a mode select switch to the Phoney Phone (Aug '90).

The control input point of the Phoney Phone is shown in the circuit diagram Fig. 4 (Aug. '90), and corresponds to the pin at $J 5$ in the stripboard layout (Fig. 9).
If you are using the Phoney Phone solely as a repeater or alarm, you may omit the monostable multivibrator section of the circuit (1C1, C1-C3, R3-R7, VR1-VR4, S2 and S3). The output from the Sound Trigger then goes to the terminal pin at $J 5$.

An alternative is to have a mode select switch which allows you to use the Phoney Phone as a timer/joke-phone or as a repeater/alarm. Cut the long wire that runs from ICI Pin 3 to IC4 pin 11 and connect the cut ends to a SPDT switch, as in Fig. 5. The output from the Sound Trigger goes to the other terminal of the switch.

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# ACTUALLY <br>  <br> by Robert Penfold 

Semiconductors must rank as one of the most confusing aspects of modern electronics. There seems to be endless different types of semiconductor, with countless examples of each of these types. The electronic component catalogues only list a fraction of the semiconductors that are actually produced, but in some cases they still list what I would guesstimate to be well in excess of a thousand different semiconductor devices. Some aspects of these components can give a few headaches even to those who have many years of experience of electronics, but most of what you need to know is not difficult to grasp.

In this month's Actually Doing It we will consider some fundamentals of this important topic. We will not consider transistors and integrated circuits on this occasion, as these are worthy of their own articles.

## ONE WAY

The mosi simple of the semiconductor components is the diode, which is a sort of electronic valve. It will allow an electric current to flow in one direction, but it will block a current flow in the opposite direction. Obviously diodes must be connected into circuit the right way round. Otherwise they block the required current flow but permit one in the wrong direction. Fig. 1 shows the circuit symbol for a diode, together with some typical physical representations.

The two terminals of a diode are called the "anode" and "cathode", and these are often shown in the abreviated forms " $a$ " and " $k$ ". respectively. In the circuit diagrams that appear in Everyday Electronics the " $a$ " and " $k$ " markings are included on diodes, but in other publications you may well find that they are omitted, with a " + " sign perhaps being included in place of the " $k$ " marking. I suppose that either type of legend is unnecessary since the diode symbol itself clearly indicates the polarity of the component. However, these markings are
useful reminders for those who are not too familiar with circuit diagrams and the circuit symbols.

## MARKINGS

Physically, ordinary diodes are virtually all in.the form of small glass bodied components having rounded ends. Rectifiers, which are merely diodes designed to handle high currents, are slightly different in that they usually have black plastic cases and are not rounded at the ends. In both cases the polarity is usually indicated by bands marked around the body of the component towards the cathode (" + " or " $k$ ") end of the component.

The three main exceptions to this method of polarity marking are diodes which have multiple bands, some high power rectifiers, and bridge rectifiers. The latter are actually four rectifiers ready connected in the bridge configuration, and contained in a single plastic encapsulation. They have four terminals: two which take the a.c. input signal and two which supply the d.c. output signal. The a.c. input can be coupled to the input terminals either way round, and they are normally marked with an a.c. waveform ("~"). The positive and negative terminals are usually just marked with " + " and " - " signs.

Some rectifiers have no band around the body, but instead have one end of the component tapered slightly. This thinner end is the cathode one (i.e. the one that would otherwise carry the white band)

## MULTI BANDS

Diodes having multiple bands are not particularly common, but do seem to crop up from time to time. You are only likely to encounter these in the form of the popular 1 N4148 silicon diode which sometimes has four coloured bands. Apparently this is a form of colour coding which has similarities to the standard resistor colour code system. It is more simple though, in that the four colours simply represent the four digits of the


Fig. 1. Common methods of polarity indication for diodes and rectifiers.


Fig. 2. Polarity indication for l.e.d.s.
type number that follow the " 1 N " prefix In the case of a 1 N 4148 , the colours are therefore yellow, brown, yellow, and grey.

So how is the polarity of the component indicated? This is achieved by merely having the bands offset towards one end of the component. They are offset towards the cathode (" + " or " " $k$ ") end of the device. In my experience the offset is quite small, which is to some extent dictated by the small body size of these components.
The band nearest the cathode leadout should be a bit wider than the others, but the difference might be quite marginal. You therefore need to look carefully at these components before deciding on their correct orientation and connecting them into circuit.

## HEAT OF THE MOMENT

It is perhaps worth pointing out that although germanium transistors are now largely obsolete, germanium diodes are still used to a significant extent. They have a lower forward voltage drop than silicon types - a factor that enables them to operate in some applications where silicon types give mediocre results or even fail completely.

A point to keep in mind when dealing with germanium diodes such as the popular OA90 and OA91, or germanium transistors, is that they are more vulnerable to heat damage than the more familiar silicon devices.

The safest way to deal with these, or any other heat sensitive components, is to use a heat-shunt when soldering them into circuit. Most heat-shunts look rather like a pair of tweezers, but they do not require you to manually hold them in place. Instead, they have some form of spring mechanism that enables them to be clipped in place, leaving both of your hands free to solder the component in place.

In use the heat-shunt is simply clipped in place over the leadout wire that is about to be soldered, somewhere between the body of the component and the soldered connection. As the heat from the iron travels up the leadout wire towards the body of the component, most of the heat is diverted into the heat-shunt. This keeps the component substantially cooler, and avoids damage unless the iron is kept in place on the joint for several seconds (which is totally unnecessary and likely to produce a poor joint).

You can improvise a heat-shunt from something like a pair of tweezers plus a small rubber band, or a crocodile clip can be used. These methods are unlikely to be as efficient as a proper heat-shunt though, and could lull you into a false sense of security.

To be honest, I do not use a heat-shunt at all, and find that no damage results to germanium components provided the soldered joints are completed fairly swiftly. However, if you are not very experienced at soldering, and are at all hesitant about this task, then it would probably be best to play safe and invest in a proper heat-shunt. These cost a matter of pence rather than pounds.

## LED ASTRAY

There are several specialised types of diode, such as variable capacitance (varicap) and Zener types. Many of these
(a)
(b)


Fig. 3. Simple l.e.d. polarity test circuit. If the l.e.d. lights it has the polarity shown in (a), if not he polarity is as shown in (b).
use one of the methods mentioned above to identify the anode and cathode leadouts. However, some types, such as many infra-red photo-diodes and multiple varicap types, have unusual encapsulations. In order to sort out the correct method of connection with these components it may well be necessary to consult the manufacturers or retailers data.

The humble light emitting diode (I.e.d.) is a specialised form of diode that is worthy of special mention. This is a component that often leads to difficulties. These are true diodes, and unlike ordinary filament bulbs, they will only light up if they are connected the right way round. There is no standard method of indicating the polarities of these components, but Fig. 2 shows two common methods. Unfortunately, there are a few l.e.d.s where the shorter lead is the anode, but these are few and far between these days. I have come across many l.e.d.s that seem to be symmetrical, with no obvious means of differentiating between the two leadout wires.

If in any doubt about the polarity of a l.e.d., remember that it is perfectly all right to use trial and error to find the right method of connection. If a l.e.d. is connected with the wrong polarity, it will not light up, but it will not be harmed either. This method is not very practical if you are fitting dozens of l.e.d.s onto a printed circuit board to produce a complex display.

The polarity of a l.e.d. can be checked using a multimeter, and the instructions supplied with the meter should give details of this simple process. If you do not have a suitable meter, then a 9 volt battery and a 1 k resistor used in the manner shown in Fig. 3 will do the job just as well.

Some of the retailers catalogues give connection details for l.e.d.s. If you are in doubt about the polarity of one of these components, consulting the appropriate catalogue might solve the problem. However, in my experience the components supplied do not always physically match the ones illustrated in catalogues (especially with I.e.d.s). This is presumably due to changes in sources of supply in order to obtain a better buy-in price, and to changes in components introduced by manufacturers. Anyway, as I have pointed out several times before, a few component catalogues should be regarded as absolutely essential for anyone pursuing electronics as a hobby, and they are a valuable source of useful data.

Having experienced difficulties with l.e.d.s on numerous occasions since they first became available, I now always check their polarity with a meter before connecting them. This saves a lot of desoldering and rewiring.

## S.C.R.s

Three types of silicon controlled rectifier (s.c.r.) are commonly used in electronic projects. The thyristor is the most basic type, and is used for switching d.c. loads. This is a three lead device, and its terminals are called the anode (a), cathod ( k ), and gate ( g ). A triac can switch a.c. loads, and is another three terminal device. Its terminals are called mains terminal 1 ( mt 1 ), mains terminal 2 ( mt 2 ), and the gate ( g ).

Both of these components are normally contained in transistor type encapsulations. A few of these devices are intended for low power applications and have small signal transistor type encapsulations. Most though, are power devices, and physically look much like power transistors.

Like transistors, each encapsulation can have several leadout configurations. It is not safe to assume that the device you are using, although of a different type number to the specified device, has the same leadout configuration as the correct device because it also has the same type of encapsulation. If you use a substitute type, check the leadout diagrams for the two components to ensure that they are the same. Triacs are bidirectional, but note that they will still only operate properly if the mt 1 and mt 2 terminals are connected around the right way.

Although, on the face of it, if you use a thyristor or a triac having suitably high current and voltage ratings it should work correctly, matters are not always as simple as this. Some s.c.r.s need a trigger current of only about 200 microamps, while others need a trigger current around one hundred times higher than this. Some of the other parameters are prone to very large variations. Unless the book or article concerned states that any thyristor or triac having certain minimum current and voltage ratings will suffice, it is best to use the specified type.

## DIACS

Diacs are two terminal components, and are mainly used as trigger devices for thyristors and triacs. They are most commonly used in a.c. power control circuits of the lamp dimmer type. Although some of these devices have encapsulations or markings which suggest that they are polarised, they are actually bidirectional. In other words, it does not matter which way round you connect them.

Some triacs have built-in diacs. It is important to use a type that has an integral diac where appropriate, and to use an ordinary type where the design does not call for a built-in diac. Using the wrong type will at best give erratic operation, and at worst will result in the circuit failing to work at all. With triacs it is definitely advisable to use the specified type, rather than dabbling with substitutes which may well turn out to be totally unsuitable.

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back with-Can you wreck a nice beach?! Here are some of the mistakes it made when read it a list of television programs. What should it have printed?

## 1) Colliders Cope

2) Casual Tea
should have been
Cora Nation's Treat should have been
3) Stale Ucky should have been
4) Start Wreck should have been
5) Woe Gun should have been

Right now we're working on a nudie vice for hoe moaners - sorry. I must remember to turn off the speech circuit. A new device for home
owners is what I meant to say, of course. Anyway, now you've filled in your answers you can choose any one of the gifts below!

1) Special blade for cutting grooves and notches
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# YOUNG AMATEUR OF THE YEAR 

Seventeen year old David Martin from Glasgow has received the Young Amateur of the Year Award 1990, sponsored by the DTI's Radiocommunications Agency. Chosen from a short-list of six. David, whose callsign is GMONVE, is a co-founder of YAGIS (Young Amateur's Group in Scotland), has converted PMR (private mobile radio) equipment for use on the amateur bands, is active in radio clubs, has run a special event station in the Scouts Jamboree-on-the-Air, and is a member of RAYNET the Radio Amateur Emergency Network. He received a $£ 250$ prize and Certificate, plus a number of other prizes, including a $2 m$ transceiver from Navico, a packet radio controller from Siskin Electronics, and a digital multimeter from Cirkit. Together with the runner-up he will also be invited to a guided tour of the RA's Monitoring Centre at Baldock in Hertfordshire.

Runner-up was Simon Glanville, G7DCY, a student committee member of the Coventry Amateur Society, who regularly assists a local elderly radio amateur who has problems using his hands. He founded his school's Remote Imaging Group which obtains data and weather pictures from satellites, encourages an interest in amateur radio by demonstrating the hobby to pupils, parents and teachers, and illustrates operating procedures to students at his local Radio Amateurs' Examination class.

The Young Amateur of the Year Award is organised by the Radio Society of Great Britain as part of its project YEAR (Youth into Electronics via Amateur Radio) initiative which aims to interest more young people in the hobby.

## REPEATER ABUSEIN U.S.A.

British amateurs are not alone in suffering repeater abuse as illustrated by a recent story in the W5YI Report concerning Richard Burton, ex-WB6JAC, who lost his licence in 1981 for malicious interference, failure to use his callsign, use of profane and obscene language, jamming a repeater, and illegal broadcasting. He refused to go off the air and in 1982 was sentenced by a Los Angeles court to eight years imprisonment, suspended for all but six months, followed by five years probation, including 1500 hours community work. It was the first time an American radio amateur had been imprisoned for illegal activity on the amateur bands.

Early in 1990 there were complaints that Burton was back on the 2 metre band, even though he had not been re-licenced, behaving in much the same manner as before. He was re-arrested on May 16 on three charges of unlicensed radio operation, having allegedly made on-the-air threats against the judge and the local FCC engineer-in-charge.

He apparently felt that the government had no prerogative to restrain his First Amendment right to say what he thought.

He refused to enter a plea and the court appointed a public defender on his behalf. Bail was set at $\$ 10,000$ and in the absence of surety he was jailed pending a trial but later, after psychiatric evaluation, was released from custody. On July 20, a jury found him guilty of transmitting without an amateur licence on one only of the three charges, and on October 1 st he was sentenced to one year's probation, a $\$ 2,000$ fine and continuing psychiatric care.

This interesting insight into the behaviour of one "repeater vandal" raises several debatable points. Firstly, as many of those who interfere with UK repeaters behave in much the same way as Burton, should they be dealt with much more severely, as offenders evidently are in the USA?

Or has he got a point. Should he be free to behave as he wishes on the air? Apparently a Los Angeles amateur "Free Radio" community felt that this was a Freedom of Speech issue, and early on some of his supporters appealed for funds, on the 2 metre band, in an attempt to raise his bail.

Another issue concerns the current trend to make entry into amateur radio much easier, to attract more recruits. Will this open the door to more subversive elements, bent on interfering with the activities of the responsible majority? Some have already got in even though they have had to pass the existing examinations to do so. How many more will follow, the easier entry gets, and what will their impact on the hobby be?

## AMATEUR RADIO IN THE THIRD WORLD

The Pacific nation of Vanuatu, formerly the New Hebrides, celebrated its 10th anniversary last July. According to a report in Amateur Radio, journal of the Wireless Institute of Australia, amateur radio is slowly establishing itself in the infant nation. There are about 20 licensees, all expatriates except for 24 year old Touasi Taiwia, YJ8NTT, a YL (young lady) who recently qualified for her novice licence.

Vanuatu is, however, better known in the amateur world than its present status would suggest. Up to 30 amateurs visit the country each year to operate from this unspoiled Dx paradise, with some setting up high scoring stations during wellknown international events such as the CQ World-Wide contest.

On behalf of the authorities the Vanuatu Amateur Radio Society (VARS) appoints an examiner for Morse examinations and invigilates the main examinations which are based on Australian licensing standards. It is not easy,
however, for the indigenous population to master the requirements of the examination. Without electricity in their homes, for example, study after sunset at 6.p.m. is difficult.

Language presents another problem when trying to grasp such concepts as inductance, capacitance, etc, from student material written in English. One possible means of overcoming this difficulty would be videotapes of basic electronic concepts obtained from Australia or elsewhere which VARS could use for instructional purposes, but there are obviously many difficulties to be overcome before amateur radio becomes a "people's hobby"

## COMPONENTS NOT AVAILABLE

There is, for instance, a great need for suitable and inexpensive equipment, but even simple components, things that we take very much for granted, are not readily available. Some years ago I corresponded with an amateur in Sri Lanka after I wrote an article about QRP (low power) operating in the American $C Q$ magazine. I recommended making one's own equipment since a handful of components could equate to a surprising degree of achievement and satisfaction. He was converted by my enthusiasm but, where, he asked, could he get the components?

In his country it was just not possible to get resistors, capacitors, transistors and so on. His immediate problem was solved by a simple ORP transceiver loaned to him through an American aid programme intended to help third world countries beginning to develop amateur radio as a national resource. Eventually he had to pass the set on to another aspiring amateur, but by that time he had become "hooked" and somehow managed to get a second-hand commercially made set from Germany.

One thing strikes me about all this. There are moves afoot in the more developed nations to make amateur radio more hi-tech, to "bring it into the space-age." There was, for instance, a proposal (which was in fact rejected) at a recent IARU conference that the amateur Morse test should be replaced by a test of computer operating skills because an increasing number of operators today use their computers in conjunction with amateur radio for data transmissions and other purposes.

If such a proposal ever comes to fruition where would this leave the amateur radio communities slowly building up in Vanuatu and elsewhere? Most beginners will of necessity be using simple equipment for Morse or speech transmission. Will they find that the goal posts have been moved, and that international contacts will not be possible unless they also have computers and master the intricacies of data communications?



BACK in the 18th Century, Benjamin Franklin flew a kite in a thunderstorm. The idea was to find out whether lightening was electrical. By observing detectors attached to the kite string he concluded that it was. Shortly afterwards someone tried to repeat the experiment. He was electrocuted.

Electric shock remains by far the greatest hazard to the experimenter, which is why magazines like this emphasize the need for safety when working with high voltages. In mains powered equipment the danger is obvious. But even equipment powered by dry batteries can deliver nasty shocks. A common case is the electronic photographic flash unit, where a large capacitance is charged to perhaps several hundred volts and may hold its charge for some time after switch-off.

## HIGH <br> FRECUENCIES

It is not sufficiently known that the body's ability to detect current flowing through it diminishes as the frequency is raised. At mains frequency the average human can feel 1 mA , but at 100 kHz even 100 mA may be imperceptible. This suggests that there is a special need for care in working with high-power, highfrequency equipment such as radio transmitters.

As the frequency is raised, there is a risk even when there is no contact with the equipment. This is because the body acts as a lossy aerial and absorbs energy when immersed in a radiofrequency field. According to the Na tional Radiological Protection Board (N.R.P.B.) which keeps an eye on these matters, the human aerial is most efficient in the range $40-100 \mathrm{MHz}$ ( $3-7$ metres wavelength).

If you walk barefoot in a radio frequency electric field the current in your body can reach $\left(0.11 \times h^{2} \times f \times E\right) / 1000$ milliamps. (Here $h$ is your height in metres, $f$ is the frequency in kilohertz and $E$ is the field strength in volts per metre). The permitted powers of amateur radio transmitters are low enough to make the risk small, and when high-power broad-
cast transmitters are built their boundary fences are normally sited so that passersby are in no danger.

Apart from shock, damage can be done by the heating effect of currents in the body. In the "standing near a radio mast" situation the ankles get the most concentrated current.

However, at microwave frequencies the eyes are vulnerable. The very short wavelengths can cause local heating. In most parts of the body blood-flow acts as a coolant but the eyes are poorly equipped with blood vessels and can suffer.

## RADAR

This is one reason why microwave cookers are carefully screened and are automatically shut off when their doors are opened. However, these sort of frequencies are also used in radar, which may be very high-powered. Radar technicians must be protected and of course be aware of hazards such as looking into a waveguide carrying microwave power.

The question arises, does long-term exposure to small doses carry a risk? Attempts have been made to show that there is a higher than normal incidence of cataracts among radar technicians, but with no very clear-cut results. If there is a risk it is a small one, provided proper precautions are taken.

## Moscow MYSTERY

Back in the 1950s, the Americans discovered that their Moscow embassy was being irradiated from outside by microwaves. The energy level was fairly low (up to 0.15 watts per square metre, corresponding to an electric field strength of 7.5 volts per metre) Nevertheless the embassy staff complained of all sorts of health problems.

Very careful investigation failed to demonstrate any connection between staff health and the microwave radiation.

## PLACEBO EFFECT

The embassy affair demonstrates how difficult it can be to establish a real connection between cause and apparent effect. If the staff knew they were being irradiated by an unfriendly power and then had headaches they were very likely to attribute the headaches to the irradiation. That's only human.

The reverse situation can also occur. If you start with a human problem and then treat it in any way the patient is likely to feel better. This is the "placebo effect", so called because doctors who suspect a patient's illness may be imaginary may prescribe a placebo, that is, a pill which is merely sugar or starch but looks genuine. Quite often, it works - for a while

After World War Two there was an interest in something called "radio-aesthesia". People were offered electronic "black boxes" said to create healthful emanations.

Convinced that this was mere quackery, somebody brought a court case against the black box providers, and lost. The judge said that, in a spirit of impartial enquiry, he had had a black box installed in his house, and felt benefit from it.

## OTHER HAZARDS

Recently there have been several attempts to show that people involved in electrical and electronic work, or merely
living near overhead power lines, are exposed to special health risks. In particular, certain rare forms of cancer seem to be less rare among these people.

In the power-line case, the initial idea, that long-term exposure to electrical and magnetic fields might be responsible, suffered a setback when it was established that fields from domestic appliances are much stronger than those from the power lines. However, the anomaly is real and requires explanation.

In the case of radio and TV service engineers one suggestion is that their work involves exposure not only to fields but also to chemicals such as the fumes from solvents and solder. Because it is suspected from animal experiments that solder fumes might cause cancer it is now normal in British factories to fit hand soldering irons with a suction device to remove them. Once again, the risk, if real, is small, and is presumably smatler still in the case of the electronics hobbyist who is not exposed for long periods.

One must never discount small risks. Nor must one be unduly worried by them. Life is full of risks. You are certainly more likely to be killed in a road accident than by solder fumes, but I don't think this need keep you off the streets.

My feeling is that until somebody shows otherwise, one should worry about electric shock and not worry unduly about other theoretical hazards. After all, there are millions of semiconductor devices in use, containing arsenic compounds. Have you ever heard of anyone being poisoned by them?

## USEFUL <br> REFERENCES

1) Advice on Protection of Workers and Members of the Public from the Possible Hazards of Electric and Magnetic Fields with Frequencies Below 300 GHz (1986).
2) Health Issues in the Siting of a Low Frequency Transmission Mast (1988).
Both published by N.R.P.B., Chilton, Didcot, Oxon. OX110RQ.


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UTLX Ultra-miniature telephone transmitter. Smallest telephone transmitter kit available. Incredible size of $10 \mathrm{~mm} \times 20 \mathrm{~mm}$. Connects to line (anywhere) and switches on and off with phone use. All conversations transmitted Powered from line, 500 m range
214.95

TLX700 Micro-miniature telephone transmitter. Best selling telephone transmitter. Being $20 \mathrm{~mm} \times 20 \mathrm{~mm}$ it is easier to assemble than UTLX. Connects to line (anywhere) and switches on and off with phone use. All conversations transmitted. Powered from line, 1000 m range...
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15.95

TKX900 Signalling/tracking transmitter. Transmits a continuous stream of audio pulses with variable tone and rate. Ideal for signalling or tracking purposes. High power output gives range up to 3000 m . Size $25 \mathrm{~mm} \times 63 \mathrm{~mm}$, 9 V operation...
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Boards for some older projects - not listed here-can often be obtained from Magenta Electronics, 135 Hunter St., Burton-on-Trent, Staffs DE14 2ST. Tel: 028365435 or Lake Electronics, 7 Middleton Close, Nuthall, Nottingham NG16 1BX. Tel: 0602382509.
NOTE: While $90 \%$ of our boards are now held in stock and are dispatched within seven days of receipt of order, please allow a maximum of 28 days for delivery -overseas readers allow extra if ordered by surface mail. Please check price and availability in the latest issue before ordering. We can only supply boards listed in the latest issue. Boards can only be supplied on a payment with order basis.

| PROJECT TITLE | Order Code | Cost |
| :---: | :---: | :---: |
| Noise Gate SEP'87 | 577 | £4.41 |
| BBC Sideways RAM/ROM NOV'87 | 585 | £4.10 |
| Pipe \& Cable Locator APR'88 | 598 | £3.00 |
| Multi-Channel Remote Light Dim <br> Receiver <br> Door Séntinel | $\begin{aligned} & 599 \\ & 600 \\ & 605 \end{aligned}$ | $\begin{aligned} & £ 3.00 \\ & £ 3.07 \\ & £ 3.00 \end{aligned}$ |
| Multi-Chan Remote Light Dim Relay/Decoder Dimmer Board Power Supply | $\begin{aligned} & 601 \\ & 602 \\ & 603 \\ & \hline \end{aligned}$ | $\begin{aligned} & £ 4.86 \\ & £ 3.07 \\ & £ 3.00 \end{aligned}$ |
| Video Wiper JUL'88 | 612 | £6.75 |
| Tea Tune <br> Time Switch Suntan Timer Car Alarm | $\begin{aligned} & 609 \\ & 614 \\ & 610 \\ & 615 \end{aligned}$ | $\begin{aligned} & £ 3.00 \\ & £ 4.84 \\ & £ 3.07 \\ & £ 3.12 \end{aligned}$ |
| Breaking Glass Alarm SEP'88 | 617 | ¢4 27 |
| EPROM Eraser OCT'88 | 620 | ¢4.07 |
|  | $\begin{aligned} & 616 \\ & 622 \\ & 623 \\ & 624 \\ & 625 \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { £3.56 } \\ & £ 4.61 \\ & £ 3.23 \\ & £ 3.05 \\ & £ 4.84 \end{aligned}$ |
| Downbeat Metronome EPROM Programmer (On Spec) DEC'88 Phasor | $\begin{aligned} & 629 \\ & 630 \\ & 631 \\ & \hline \end{aligned}$ | $\begin{array}{r} \mathrm{£} 4.84 \\ £ 8.29 \\ £ 5.64 \end{array}$ |
| Monkey/Hunter Game JAN'89 | 634 | £3.36 |
| Continuity Tester FEB'89 | 619 | £2.67 |
| Sound-to-Light Interface <br> Midi Pedal <br> Midi Merge <br> Audio Lead Tester | $\begin{aligned} & 637 \\ & 639 \\ & 640 \\ & 641 \end{aligned}$ | $\begin{array}{r} £ 6.24 \\ £ 7.00 \\ £ 3.00 \\ £ 5.77 \end{array}$ |
| Light Sentinel <br> Main Control Board <br> Remote Interface (4 boards) <br> 4-Channel Auto-Fade Interface | $\begin{aligned} & 632 \\ & 633 \\ & 642 \end{aligned}$ | $\begin{array}{r} £ 9.20 \\ £ 4.59 \\ £ 6.80 \end{array}$ |
| Pet Scarer <br> Electron A/D Interface MAY'89 | $\begin{aligned} & 644 \\ & 645 \\ & \hline \end{aligned}$ | $\begin{aligned} & £ 3.00 \\ & £ 4.84 \end{aligned}$ |
| Spectrum EPROM Programmer JUN'89 Bat Detector | $\begin{aligned} & 628 \\ & 647 \end{aligned}$ | $\begin{aligned} & \hline 7.87 \\ & £ 4.95 \\ & \hline \end{aligned}$ |
| Programmable Pocket Timer JUL'89 | 648 | £3.82 |
| Electronic Spirit Level AUG 89 <br> Distance Recorder  <br> Treasure Hunter  <br> Xenon  | $\begin{aligned} & 649 \\ & 651 \\ & 652 \end{aligned}$ | $\begin{aligned} & £ 3.85 \\ & £ 5.23 \\ & £ 3.73 \end{aligned}$ |
|  | $\begin{aligned} & 650 \\ & 653 \\ & 654 \\ & 655 \end{aligned}$ | $£ 4.13$ $£ 4.12$ $£ 4.08$ $£ 4.48$ |
| Music on Hold <br> Power Supplies - 25V 700 mA <br> 30 V 1 A <br> EE Seismograph - Control <br> Detector <br> Lego/Logo \& Spectrum | 6 656 657 658 659 660 | $\begin{aligned} & £ 3.85 \\ & £ 4.35 \\ & £ 4.55 \\ & £ 4.08 \\ & £ 4.22 \\ & £ 6.49 \end{aligned}$ |
| Wash Pro <br> Biofeedback Monitor - Front End <br> Processor <br> Power Supplies -1.5 V - 25V 2A <br> Logo/Lego \& Spectrum Interiace | $\begin{aligned} & 643 \\ & 661 \\ & 662 \\ & 663 \\ & 664 \end{aligned}$ | $\begin{aligned} & £ 3.83 \\ & £ 4.52 \\ & £ 4.56 \\ & £ 4.78 \\ & £ 5.60 \end{aligned}$ |


| PROJECT TITLE | Order Code | Cost |
| :---: | :---: | :---: |
| EEG Electrode Impedance Meter [DEC'89] | 665 | $£ 3.98$ |
| Biofeedback Signal Generator JAN 90 <br> Four-Channel Light Chaser  | $\begin{aligned} & 666 \\ & 667 \end{aligned}$ | $\begin{aligned} & £ 4.08 \\ & £ 6.70 \end{aligned}$ |
| Quick Cap Tester <br> Weather Station <br> Anemometer - Freq. /Volt Board Optional Display <br> Wind Direction <br> System Power Supply <br> Prophet In-Car Ioniser | $\begin{aligned} & \hline 668 \\ & \\ & 670 \\ & 669 \\ & 673 / 674 \\ & 675 \\ & 676 \end{aligned}$ | $\begin{aligned} & £ 3.92 \\ & \\ & £ 3.94 \\ & £ 3.73 \\ & £ 4.22 \\ & £ 3.59 \\ & £ 3.18 \\ & \hline \end{aligned}$ |
| EE Weather Station <br> Display Driver <br> Display and Sensor <br> Fermostat Mk2 <br> Superhet Broadcast Receiver-Tuner/Amf | $\begin{aligned} & 672 \& 678 \\ & 671 \\ & 677 \\ & 679 / 680 \end{aligned}$ | $\begin{aligned} & £ 4.22 \\ & £ 4.47 \\ & £ 4.28 \\ & £ 4.22 \end{aligned}$ |
| Stereo Noise Generator APR'90 Digital Experimenter's Unit - Pulse Generator Enlarger Timer EE Weather Station Supply Rainfall/Sunlight Display Rainfall Sen and Sunlight Sen | 681 <br> 682 <br> 683 <br> 684 <br> 685 <br> 686/687 | £4.24 £4.46 £3.66 £4.28 <br> £4.27 £4.16 |
| Amstrad Speech Synthesiser MAY'g0 Quizmaster | $\begin{aligned} & 689 \\ & 690 \end{aligned}$ | $\begin{array}{r} £ 4.68 \\ £ 4.74 \end{array}$ |
| 80 Metre Direct Conversion Radio JUN'90 Mains Appliance Remote Control Infra-Red Transmitter | $\begin{aligned} & 691 \\ & 692 / 693 \end{aligned}$ | $£ 4.95$ <br> $£ 4.75$ |
| Mains Appliance Remote Control Encoder Board A Encoder Board B The Tester | $\begin{aligned} & 694 \\ & 695 \\ & 696 \end{aligned}$ | $\begin{aligned} & £ 6.61 \\ & £ 4.78 \\ & £ 4.15 \end{aligned}$ |
| Mains Appliance Remote Control <br> Mains ON/OFF Decoder <br> ( 5 or more $697^{\circ}$ s ordered together $£ 3.25$ each) Simple Metronome | $\begin{array}{r}697 \\ 698 \\ \hline\end{array}$ | $\begin{array}{r}£ 4.55 \\ £ 3.94 \\ \hline\end{array}$ |
| Hand Tally Main Board (double-sided) Display Alarm Bell Time-Out Mains Appliance Remote Control Temperature Controler (p.c.b. only) | $\begin{aligned} & 699,700 \\ & 701 \\ & 702 \\ & \hline \end{aligned}$ | $\begin{aligned} & £ 10.95 \\ & £ 4.10 \\ & £ 5.20 \\ & \hline \end{aligned}$ |
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| Colour Changing Christmas Lights <br> Microcontroller Light Sequencer <br> Versatile Bench Power Supply Unit <br> Teach-In '91, Part 1 - Design Your Own Circuits L200 Module <br> Dual Output Module <br> LM723 Module | $\begin{aligned} & \hline 707 \\ & 708 / 709 \\ & 710 \\ & 711 \\ & 712 \\ & 713 \\ & \hline \end{aligned}$ | $\begin{aligned} & £ 4.39 \\ & £ 10.90 \\ & £ 4.24 \\ & £ 3.93 \\ & £ 4.13 \\ & £ 4.21 \end{aligned}$ |
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| 1W amplifier Mullard ref 1172.9 V . | 1 | £1.00 | BD114 |
| Mini mono amp 3W into 4 ohms 12v operation. | 1 | $£ 1.00$ | BD495 |
| Stereo ex personal cass amplifier high gain for mag input. | 1 | $£ 1.00$ | BD680 |
| 150w stereo power amp 12v 20-20KHZ. Cased. | 1 | £57.00 | 57P1 |
| Personal stereo with FM radios (Customer returns). | 1 | $£ 6.00$ | 6P34 |
| 7 channel graphic equalizer plus 60 w power amp. 12v. BASES | 1 | £25.00 | 25P14 |
| B7G valve bases. chassis mount. | 5 | £1.00 | BD94 |
| B9A valve bases chassis mount. | 5 | £1.00 | BD95 |
| 11 pin moulded bases for valves or relays. chassis mnt. | 4 | £1.00 | BD93 |
| 14 pin continental relay bases. | 4 | £1.00 | BD524 |
| BATTERY CHARGERS |  |  |  |
| Nicad battery charger with 6 v output at .9 VA . | 1 | $£ 2.00$ | 2P154 |
| Constant current charger for button cells. 240 v op. | 2 | £1.00 | BD30 |
| Transformer type Nicad charger, mains op. | 2 | £1.00 | BD385 |
| Nicad charger 5.2 v output at .7 VA . BATTERIES AND BATTERY HOLDERS | 1 | £2.00 | 2P153 |


| DESCRIPTION | $\left\lvert\, \begin{array}{c\|} \text { PACK } \\ \text { OUANTITY } \end{array}\right.$ | PACK | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| Nicad battery PCB mount $25 \times 25 \times 15 \mathrm{~mm} \mathrm{3.6v} 100 \mathrm{maH}$. | 1 | £2.00 | 2P340 |
| 4aH D size nicad cell. | 6 | £10.00 | 10 P 47 |
| Battery holder for 6 D cells. | 1 | £1.00 | BD286 |
| Battery holder for 2 D cells. | 2 | £1.00 | BD287 |
| PP3 battery connectors with leads. | 5 | £1.00 | BD759 |
| Lithium battery 3 v . $24 \mathrm{~mm} \times 2 \mathrm{~mm}$ PCB mount. | 2 | £1.00 | BD558 |
| AA nicad rechargeable battery. | 4 | £4.00 | 4P44 |
| C size nicad battery. | 2 | £4.00 | 4P73 |
| D size nicad battery. | 4 | $\varepsilon 9.00$ | 9 P 12 |
| PP3 size 9v nicad battery. | 1 | ¢6.00 | 6P35 |
| Universal charger to take all above batteries. | 1 | £6.00 | 6P36 |
| YUASHA 6 volt 10 AH sealed lead acid. Rechargeable. | 1 | £12.00 | 12P24 |
| Lithium battery 9 volt $33 \times 13 \mathrm{~mm}$. BOOKS | 1 | £2.00 | 2P290 |
| 25 simple amateur band aerials. | 1 | £1.95 | BP125 |
| 25 indoor and window aerials. | 1 | £1.75 | BP136 |
| 30 solderless breadboard projects book 1. | 1 | £2.95 | BP107 |
| 30 solderless breadboard projects book 2. | 1 | £2.25 | BP113 |
| 50 FET projects. | 1 | $£ 2.95$ | BP39 |
| 50 circuits using 7400 series ic's. | 1 | $£ 2.50$ | BP228 |
| 50 circuits using germanium \& silicon diodes. | 1 | £1.50 | BP278 |
| 50 projects using relays scr's and triacs. | 1 | £2.95 | BP37 |
| 50 simple LED projects book 2. | 1 | £1.95 | BP87 |
| A concise introduction to MS-DOS. | 1 | $£ 2.95$ | BP264 |
| A concise introduction to D-BASE. | 1 | £3.95 | BP263 |
| A concise introduction to GEM. | 1 | £2.95 | BP230 |
| A concise introduction to MS-DOS. | 1 | £2.95 | BP232 |
| A concise introduction to OS-2. | 1 | £2.95 | BP260 |
| A concise introduction to SYMPHONY. | 1 | £3.95 | BP270 |
| A concise introduction to BBC BASIC. | 1 | £1.95 | BP149 |
| A concise introduction to WORD PERFECT. | 1 | £2.95 | BP262 |
| A concise introduction to SUPERCALC. | 1 | £3.95 | BP274 |
| A concise introduction to LOTUS 1-2-3. | 1 | £2.95 | BP261 |
| Word processing on the 8256 \& 8512. | 1 | $£ 5.95$ | BP187 |
| TV DXers handbook. | 1 | $£ 5.95$ | BP176 |
| A Z80 workshop manual. | 1 | £3.95 | BP112 |
| Advanced shortwave receiver construction. | 1 | £3.95 | BP276 |
| Alternating current theory. | 1 | £3.50 | BP63 |
| An introduction to BASIC on a PC. | 1 | $£ 5.95$ | BP199 |
| An introduction to amateur radio. | 1 | £3.50 | BP257 |
| An introduction to computer graphics. | 1 | £4.95 | BP268 |
| An introduction to desktop publishing. | 1 | £5.95 | BP269 |
| An introduction to loudspeaker enclosure design. | 1 | £2.95 | BP256 |
| An introduction to satellite television. | 1 | £5.95 | BP195 |


| DESCRIPTION | $\left\lvert\, \begin{gathered} \text { PACK } \\ \text { OUNNTITY } \end{gathered}\right.$ | PACK PRICE | ORDER <br> NUMBEA |
| :---: | :---: | :---: | :---: |
| An Introduction to Amstrad PC's. | 1 | £5.95 | BP197 |
| An introduction to $\mathbf{Z 8 0}$ machine code. | 1 | £2.75 | BP152 |
| Audio amplifier construction. | 1 | $£ 2.95$ | BP122 |
| Audio projects. | 1 | £2.50 | BP90 |
| Coil design and construction manual. | 1 | £2.50 | BP160 |
| Computer hobbyists hand book. | 1 | £5.95 | BP251 |
| Computer music projects. | 1 | $\varepsilon 2.95$ | BP173 |
| Digital audio projects. | 1 | £2.95 | BP245 |
| Digital IC equivalents and pin connections. | 1 | $£ 5.95$ | BP140 |
| Electronic circuits for model railways. | 1 | £2.95 | BP180 |
| Electronic circuits for robots. | 1 | £2.95 | BP179 |
| Electronic hobbyists handbook. | 1 | £4.95 | BP233 |
| Electronic projects for cars and boats. | 1 | £1.95 | BP94 |
| Electronic science projects. | 1 | £2.95 | BP104 |
| Electronic security projects. | 1 | £2.50 | BP56 |
| Crystal set construction. | 1 | £1.75 | BP92 |
| Electronic calculations and formulae. | 1 | £4.95 | BP144 |
| HI FI loud speaker enclosures. | 1 | £2.95 | BP205 |
| High power amplifier construction. | 1 | £3.95 | BP277 |
| How to design and make your own PCB's. | 1 | £2.50 | BP121 |
| How to design electronic projects. | 1 | £2. 25 | BP127 |
| How to get your electronic projects working. | 1 | £2.50 | BP110 |
| How to identify unmarked IC's. | 1 | £0.95 | BP101 |
| 555 timer project book. | 1 | $\underline{\mathrm{E}} .95$ | BP44 |
| Interfacing PC's and compatables. | 1 | £0.95 | BP272 |
| International diode equivalents book. | 1 | £2.95 | BP108 |
| Intemational radio stations guide. | 1 | £4.95 | BP255 |
| Intemational transistor equivalents. | 1 | £3.50 | BP85 |
| An introduction to UNIX. | 1. | £2.95 | BP259 |
| Linear IC's equivalents and pin connections. | 1 | $£ 5.95$ | BP141 |
| Midi projects. | 1 | $£ 2.95$ | BP184 |
| Model railway projects. | 1 | $\varepsilon 2.95$ | BP95 |
| Modern op-amp projects. | 1 | £1.95 | BP106 |
| Modem opto device projects. | 1 | £2.95 | BP194 |
| Advanced electronic security projects. | 1 | £2.95 | BP190 |
| Advanced midi projects. | 1 | £2.95 | BP247 |
| Advanced power supply projects. | 1 | £2.95 | BP192 |
| Popular electronic circuits book1. | 1 | £2.95 | BP80 |
| Popular electronic circuits book2. | 1 | £2.95 | BP98 |
| Power transistor selctor guide. | 1 | £4.95 | BP235 |
| Power supply projects. | 1 | $\varepsilon 2.50$ | BP76 |
| Projects in opto electronics. | 1 | £1.95 | BP45 |
| Upgrading and repairing PC's and compatables. | 1 | £3.95 | BP272 |
| Remote control handbook. | 1 | £3.95 | BP240 |


| DESCRIPTION | PACK QUANTITY | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| Simple shortwave recelver construction. | 1 | £3.95 | BP275 |
| TOWERS INTERNATIONAL |  |  |  |
| TRANSISTOR SELECTOR GUIDE. | 1 | $£ 20.00$ | 20P32 |
| BOXES AND CASES |  |  |  |
| Self adhesive brushed aluminium panel. Can be cut |  |  |  |
| with scissors. Very flexible. $16^{\prime \prime} \times 3^{\prime \prime}$. | 1 | £2.00 | 2P283 |
| Amplifier junction box. Metal sloping front. | 1 | £2.00 | 2P256 |
| Black ABS project box $165 \mathrm{~mm} \times 119 \mathrm{~mm} \times 75 \mathrm{~mm}$. | 1 | £3.00 | 3P49 |
| Metal box $8^{\prime \prime} \times 41 / 2^{\prime \prime} \times 4^{\prime \prime}$ louvred ends, ideal psu. | 1 | £3.00 | 3P75 |
| Plastic box with window, ideal for beam switch etc. | 2 | £1.00 | BD132 |
| Metal box $8^{\prime \prime} \times 4^{\prime \prime} \times 1^{\prime \prime}$, slightly sloping. | 1 | £1.00 | BD209 |
| PLastic box with screw on lid $216 \mathrm{~mm} \times 130 \mathrm{~mm} \times 85 \mathrm{~mm}$ | 1 | £4.00 | 4P7 |
| Plastic box with screw on lid $220 \mathrm{~mm} \times 159 \mathrm{~mm} \times 64 \mathrm{~mm}$ | 1 | £4.00 | 4P8 |
| Signal box. 3 lamps on face plate of metal box. | 1 | £3.00 | 3P16 |
| Waterproof case will take 100W transiormer. | 1 | £3.00 | 3 P 15 |
| $4^{\text {"1 }} \times 2^{\text {" }} \times 2^{\text {" }}$ project boxes with slot in top. | 2 | £1.00 | BD780 |
| ABS project box $90 \times 50 \times 25 \mathrm{~mm}$ black. | 1 | £1.00 | CD25 |
| ABS project box $110 \times 56 \times 20 \mathrm{~mm}$ black. | 1 | £1.00 | CD26 |
| ABS project box $100 \times 75 \times 40 \mathrm{~mm}$ black. | 1 | £1.00 | CD27 |
| ABS project box $150 \times 100 \times 55 \mathrm{~mm}$ black. | 1 | $£ 2.00$ | 2P341 |
| BURGLAR ALARMS |  |  |  |
| Ultrasonic intruder detector or alarm 12v op. | 1 | £18.00 | 18P3 |
| As above but with extemal siren and mains adaptor. | 1 | £30.00 | 30P5 |
| CABLE TIES |  |  |  |
| $75 \mathrm{~mm} \times 2.4 \mathrm{~mm}$ nylon white cable ties. | 100 | $£ 1.00$ | BD868 |
| $75 \mathrm{~mm} \times 2.4 \mathrm{~mm}$ nylon white cable ties (1,000 pack). | 1 | £5.00 | 5P181 |
| $142 \mathrm{~mm} \times 3.2 \mathrm{~mm}$ nylon white cable ties. | 100 | £3.00 | $3 \mathrm{P104}$ |
| $142 \mathrm{~mm} \times 3.2 \mathrm{~mm}$ nylon white cable ties ( 1,000 pack). | 1 | £14.00 | 14P6 |
| $385 \mathrm{~mm} \times 5 \mathrm{~mm}$ nylon white cable ties. | 100 | £10.00 | $10 \mathrm{P97}$ |
| Cable tie bases $21 \times 21 \mathrm{~mm}$ self adhesive. | 100 | £5.00 | 5P182 |
| Cable tie bases $28 \times 28 \mathrm{~mm}$ self adhesive. | 100 | £5.00 | 7 P 25 |
| Cable tie gun. Tensions then cuts tie. | 1 | £6.00 | 6P38 |
| Spiral cable wrap for $6-50 \mathrm{~mm}$ bundles ( 10 m length). | 1 | £2.00 | 2P329 |
| Spiral cable wrap for 12.70 mm bundles ( 10 m length). | 1 | £4.00 | 4P74 |
| CAPACITOR BARGAIN PACKS |  |  |  |
| Mixed pack of non electrolytic capacitors. | 100 | £2.00 | 2P286 |
| Mixed pack of electrolytic capacitors. | 40 | $£ 2.00$ | 2 P 287 |
| CAPACITORS AC WORKING |  |  |  |
| Capacitor 8ut 440 v AC $97 \times 45 \mathrm{~mm}$. | 1 | £1.00 | BD632 |
| Capacitor 1 uf 440 v AC $48 \times 38 \mathrm{~mm}$. | 2 | £1.00 | BD633 |
| Capacitor 15 uf 440 v AC $120 \times 75 \times 50 \mathrm{~mm}$. | 1 | £2.00 | 2 P 201 |
| Capacitor 20ut 440 v AC $150 \times 75 \times 50 \mathrm{~mm}$. | 1 | $\varepsilon 2.00$ | 2P200 |
| Capacitor 2 ut 440 v AC $70 \times 35 \mathrm{~mm}$. | 1 | £2.00 | 2P164 |
| Capacitor 12 uf 660 V AC $150 \times 90 \times 45 \mathrm{~mm}$. | 1 | £2.00 | 2P163 |


| DESCRIPTION | $\begin{array}{\|c\|} \hline \text { PACK } \\ \text { OUANTITr } \end{array}$ | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| Capacitor 1.5 f 440 v AC $60 \times 40 \times 25 \mathrm{~mm}$. | 1 | £1.00 | BD279 |
| Capacitor 2.5 uf 440 v AC $73 \times 30 \mathrm{~mm}$. | 1 | $£ 2.00$ | 2P176 |
| CAPACITORS ELECTROLYTIC |  |  |  |
| Capacitor 15,500 uf 10 v can type $110 \times 50 \mathrm{~mm}$. | 1 | £1.00 | BD18 |
| Capacitor $4,700 \mathrm{uf} 25 \mathrm{v}$ axial. | 4 | £1.00 | BD613 |
| Capacitor $3,150 \mathrm{uf} 40 \mathrm{~V}$ can type $110 \times 38 \mathrm{~mm}$. | 2 | £1.00 | BD237 |
| Capacitor 8uf 275 v plastic case $35 \times 15 \mathrm{~mm}$. | 2 | £1.00 | BD573 |
| Capacitor 32uf + 32uf 350v. |  |  |  |
| Capacitor $32 \mathrm{uf} 500 \mathrm{v} 75 \times 25 \mathrm{~mm}$. | 2 | £1.00 | BD609 |
| Capacitor 10 uf 25 v radial. | 20 | £1.00 | CD26 |
| Capacitor 22 uf 25 v radial. | 20 | £1.00 | CD27 |
| Capacitor 47 uf 25 v radial. | 15 | £1.00 | CD28 |
| Capacitor 100 ut 25 v radial. | 12 | £1.00 | CD29 |
| Capacitor 220 uf 25 v radial. | 10 | £1.00 | CD30 |
| Capacitor 470 uf 25 v radial. | 6 | £1.00 | CD31 |
| Capacitor $1000 u f 25 \mathrm{v}$ radial. | 5 | £1.00 | CD32 |
| Capacitor 2200 uf 25 v radial. | 4 | £1.00 | BD856 |
| Capacitor 4700 uf 16 v radial. | 4 | £1.00 | BD857 |
| Capacitor $4,700 \mathrm{16v}$ | 4 | £1.00 | BD858 |
| Capacitor 4.7 uf 63 v . | 10 | £1.00 | BD859 |
| Capacitor 1 uf 63 v . | 12 | £1.00 | BD860 |
| Capacitor 2200uf 6.3v 5.8A can type. | 1 | £1.00 | BD644 |
| Capacitor 4700 uf $100 \mathrm{v} 10 \mathrm{~A} 105 \times 50 \mathrm{~mm}$. | 1 | £2.00 | 2P178 |
| CAPACITORS EHT |  |  |  |
| Capacitor 1000 pf 12 kv ceramic. | 2 | £1.00 | BD439 |
| Capacitor 220pt 8kv ceramic. | 4 | £1.00 | BD440 |
| Capacitor 150pf 8kv ceramic. | 5 | £1.00 | BD441 |
| Capacitor 100 pf 8 kv ceramic. | 5 | £1.00 | BD442 |
| Capacitor 68 pf 8 kv ceramic. | 10 | £1.00 | BD443 |
| Capacitor $2.5 \mathrm{nf} 10 \mathrm{kv} 35 \times 15 \mathrm{~mm}$. | 2 | £1.00 | BD522 |
| Capacitor .265uf 1.5 kv metal case. $110 \times 5565 \mathrm{~mm}$ | 1 | £1.00 | BD362 |
| Capacitor . 001 uf 28 kv plastic $120 \times 60 \times 80 \mathrm{~mm}$. | 1 | £2.00 | 2P124 |
| CAPACITORS VARIABLE |  |  |  |
| Air spaced 2 gang $1 / 4^{\prime \prime}$ spindle 365pf. | 2 | £1.00 | BD36 |
| Solid di-electric AM \& FM sections. Two gang. | 2 | £1.00 | BD37 |
| Solid di-electric compression trimmers. | 10 | £1.00 | BD38 |
| Minature twin tuning condensor 150pi $1 / 4^{\prime \prime}$ sindle. | 1 | $£ 2.00$ | 2 P 237 |
| Twin gang 350pf Jackson $1 / 4^{\prime \prime}$ splndle. | 1 | £2.00 | 2P170 |
| Transmitter tuning condensor 160pf (ex-equip). | 1 | £1.00 | BD424 |
| Trimmer capacitor 3-40pi Muliard. | 1 | £1.00 | BD656 |
| CIRCUIT BREAKERS |  |  |  |
| 10A trip with additional signalling contacts (Belling lee). thermal magnetic type. 2 button $55 \times 20 \times 30 \mathrm{~mm}$. COMPUTER DATA SWITCHES | 1 | £3.00 | 3P103 |


| DESCRIPTION | PACK QUANTITY | PACK PAICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| 2 way RS232 DATA swich (3 female D25). Cased. | 1 | £14.00 | 14P7 |
| 2 way Centronics data switch (3 Centronics fem) cased. CONTACTORS | 1 | £18.00 | 18P9 |
| Heavy duty $24 v$ DC operation 4 pole 25A $95 \times 60 \times 70 \mathrm{~mm}$. | 1 | $£ 1.00$ | BD68 |
| Heavy duty mains operation 4 pole 25A $90 \times 75 \times 90 \mathrm{~mm}$. | 1 | £5.00 | 5P64 |
| COUNTERS |  |  |  |
| Tape deck counter 3 digit. resettable. flywheel/belt op. | 2 | £1.00 | BD26 |
| 7 digit 24v counter panel mount. | 1 | £2.00 | 2P267 |
| 7 digit even numbercounter. mains voltage $47 \times 70 \times 32 \mathrm{~mm}$ | 1 | £1.00 | BD890 |
| Resettable 3 digit. mains operated $45 \times 70 \times 60 \mathrm{~mm}$. | 1 | £2.00 | 2 P 26 |
| 6 digit counter 12v DC operation. | 1 | £2.00 | 2 P342 |
| CRYSTALS |  |  |  |
| 1000 KHZ crystal | 1 | £1.00 | BD866 |
| 5242.880 HZ | 1 | £1.00 | BD867 |
| $2 \mathrm{mhz} 10 \times 12 \mathrm{~mm}$. | 1 | £1.00 | BD658 |
| $6 \mathrm{mhz} 10 \times 12 \mathrm{~mm}$. | 1 | £1.00 | BD659 |
| $8 \mathrm{mhz} 12 \times 5 \mathrm{~mm}$. | 1 | £1.00 | BD937 |
| $12 \mathrm{mhz} 12 \times 5 \mathrm{~mm}$. | 1 | £1.00 | BD938 |
| $16 \mathrm{mhz} 12 \times 5 \mathrm{~mm}$. | 1 | £1.00 | BD939 |
| $18 \mathrm{mhz} 12 \times 5 \mathrm{~mm}$. | 1 | £1.00 | BD940 |
| CONNECTORS |  |  |  |
| 9 way D type male solder. Gold plated. | 3 | £1.00 | BD941 |
| 9 way D type female solder. Gold plated. | 3 | £1.00 | BD942 |
| Plastic hood for 9 way connector. | 3 | £1.00 | BD943 |
| 15 way D type male solder. Gold plated. | 2 | £1.00 | BD944 |
| 15 way D type female solder. Gold plated. | 2 | £1.00 | BD945 |
| Plastic hood for 15 way connector | 2 | $£ 1.00$ | BD947 |
| 25 way D type male solder. Gold plated. | 3 | £2.00 | 2P306 |
| 25 way D type female solder. Gold plated. | 3 | £2.00 | 2 P307 |
| Plastic hood for 25 way connector. | 3 | £2.00 | 2P308 |
| Centronics 36 way cable plug (inc cover). | 1 | £1.00 | BD948 |
| 1 mm banana plug. red. | 4 | £1.00 | BD949 |
| 1 mm banana plug black. | 4 | $£ 1.00$ | BD950 |
| 1 mm banana socket. red. | 4 | £1.00 | BD951 |
| 1 mm banana socket. black. | 4 | £1.00 | BD952 |
| 4 mm banana plug. red. | 4 | £1.00 | BD953 |
| 4 mm banana plug. black. | 4 | £1.00 | BD954 |
| 4 mm banana socket red | 4 | £1.00 | BD955 |
| 4 mm banana socket. black. | 4 | £1.00 | BD956 |
| Crocodile clips (pack of 10 red and 10 black). | 1 | £2.00 | 2 P309 |
| BNC 50 ohm plug. | 3 | £2.00 | 2 P 310 |
| BNC 75 ohm plug. | 3 | £2.00 | 2 P 311 |
| SCART plug. | 1 | £1.00 | BD957 |
| CO-AX plug (TV type). | 5 | £1.00 | BD958 |


| DESCRIPTION | $\begin{aligned} & \text { PACK } \\ & \text { OUANTITY } \end{aligned}$ | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| CO-AX socket (TV type). | 2 | £1.00 | BD959 |
| 2.5 mm jack plug. (Mono). | 5 | £1.00 | BD960 |
| 3.5 mm jack plug. (Stereo). | 3 | £1.00 | BD961 |
| CONNECTOR AND TERMINAL STRIPS |  |  |  |
| 12 way 3A connector strip. One side screw terminals the |  |  |  |
| 12 pole 25A poly connector strip. Screw type. | 2 | £1.00 | BD159 |
| 12 way 5A connector strip. | 4 | £1.00 | BD158 |
| 3 way connectors plug in type. | 4 | £1.00 | BD160 |
| CLOCKS |  |  |  |
| Flip over digital clock with DIY case. | 1 | £3.00 | 3P139 |
| Electric clock malins operated, no case. | 1 | £1.00 | BD211 |
| COILS AND FORMERS |  |  |  |
| Subminature VF transiormers (all the same type). | 100 | £1.00 | BD360 |
| 465 KC VF transformers $1 / 2^{\prime \prime} \times 11 / 2^{\prime \prime}$ high. | 4 | £1.00 | BD40 |
| COMPONENT MOUNTINGS |  |  |  |
| 50 tag component mounting strip. | 2 | £1.00 | BD168 |
| COMPUTER BITS |  |  |  |
| 16K ram pack for $\mathrm{ZX81}$ computer. | 1 | $£ 8.00$ | 8P22 |
| $8^{\prime \prime}$ ex equipment disc drives. | 1 | £8.00 | 8P37 |
| Spectrum sound box with amplifier etc. | 1 | £4.00 | 4P53 |
| BBC joystick with 2 fire buttons. | 1 | £5.00 | 5P159 |
| Quickshot joycard VII for Atari and Commodore. | 1 | £3.00 | 3P85 |
| Quickshot joyball IX for Atari and Commodore. | 1 | £5.00 | 5P164 |
| Trackball for Dragon computers. | 1 | £3.00 | 3P86 |
| Oric/Atmos interface for 2 joysticks with s/ware. | 1 | £1.00 | BD757 |
| Chinnon cased disc drive with leads 3 1/2"360K | 1 | £40.00 | 40P1 |
| Acom data recorder ALF503 with psu and leads. | 1 | £15.00 | 15P43 |
| Coverter to change bbc joystick port to Atari type. | 1 | £2.00 | 2P261 |
| Computer terminals s/hand. mixed makes. | 1 | £15.00 | 15 P 33 |
| Roll of paper for plotter. | 1 | £3.00 | 3P102 |
| Keyboard made for OPD Computer. | 1 | £3.00 | 3 P 27 |
| Commodore 64 games pack (5 different). | 1 | £3.00 | $3 \mathrm{P97}$ |
| Spectrum 48K games pack (5 different). | 1 | £3.00 | 3P96 |
| PC power supply 150W switch on back. Cased. | 1 | £25.00 | 25P18 |
| DELAY SWTCHES |  |  |  |
| 0.25 hour delay switch. | 1 | £1.00 | BD101 |
| Mains motor driven switch 20 secs on or off after push of button. | 1 | £3.00 | 3P138 |
| DIODES |  |  |  |
| OA91 germanium signal diode. | 10 | $\varepsilon 1.00$ | BD976 |
| IN4148 signal diode. | 30 | ع1.00 | BD977 |
| IN4001 50v 1A | 25 | $\Sigma 1.00$ | BD971 |
| IN4002 100v 1A | 25 | £1.00 | BD972 |


| DESCRIPTION | $\left\|\begin{array}{c} \text { PACK } \\ \text { OUANTITY } \end{array}\right\|$ | $\begin{aligned} & \text { PACK } \end{aligned}$ | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| IN4006 800v 1A | 20 | £1.00 | BD973 |
| IN5401 100v 3A | 10 | £1.00 | BD974 |
| IN5408 1000v 3A | 8 | £1.00 | BD975 |
| DISPLAYS |  |  |  |
| 16 character 2 line display. Epsom with data. | 1 | £10.00 | 10P50 |
| Vaccuum flourescent displays. 4 letters or digits. | 2 | £1.00 | BD614 |
| 7 segment displays common cathode . $5^{\prime \prime}$. | 4 | £1.00 | BD466 |
| . $3^{\prime \prime} 7$ segment LED display. Common cathode. | 4 | £3.00 | 3 P 117 |
| . $3^{\prime \prime} 7$ segment LED display. Common anode. | 4 | £3.00 | 3 P 118 |
| .5" 7 segment LED display. Common cathode. | 4 | £3.00 | 3 P 119 |
| .5* 7 segment LED display. Common anode. | 4 | £3.00 | 3 P 120 |
| .56" 7 segment LED display. Common cathode. | 4 | £3.00 | 3 P 121 |
| . $56^{\prime \prime} 7$ segment LED display. Common anode. | 4 | £3.00 | 3 P 122 |
| Clock display $1 / 2$ " figures. | 1 | £1.00 | BD329 |
| 16 character 1 line display no data. | 1 | £6.00 | 6P32 |
| LCD display 4 digit with connection data. | 1 | £3.00 | 3P77 |
| ELECTRICAL ACCESSORIES |  |  |  |
| 13A switched double socket. White. New. | 1 | £4.00 | 4P75 |
| 13A switched single socket. White. New. | 1 | $£ 2.00$ | 2P343 |
| Single gang light switch. White. New. | 1 | £1.00 | CD27 |
| Two gang light switch. White. New. | 1 | £2.00 | 2 P 344 |
| Ceiling rose. White. New | 1 | £1.00 | CD28 |
| Pendant lampholder. White. New. | 2 | £1.00 | CD29 |
| Pendant lampholder with switch. White. New. | 1 | £1.00 | CD30 |
| Batten lampholder. White. New. | 1 | £1.00 | CD31 |
| Battened lampholder angled. White. New. | 1 | £1.00 | CD32 |
| Junction box 15A. White. New. | 1 | £1.00 | CD33 |
| Junction box 30A. White. New. | 2 | £3.00 | 3 P 115 |
| Dimmer switch 630 watt. White. New. | 1 | £4.00 | 4P77 |
| Dimmer switch 630 watt brushed aluminium. New. | 1 | £5.00 | 5P185 |
| White flush light switches. Standard fixing. | 2 | £1.00 | BD5 |
| Double pole 20A switch on standard plate with neon. | 1 | £1.00 | BD531 |
| Double pole 20A white mains switch that fit a standard pattress. | 1 | £1.00 | BD189 |
| 20A double pole switch with neon. Surface mount. | 1 | £2.00 | 2 2P58 |
| Double pole 20A mains brown surface mount switch. | 2 | £1.00 | BD190 |
| White shallow pattress for switches etc. | 10 | £1.00 | BD338 |
| White blanking plates for standard pattress. | 5 | £1.00 | BD358 |
| Mild steel boxes with knockouts $6^{\prime \prime} \times 3^{\prime \prime}$. | 2 | £2.00 | 2P56 |
| 13A panel socket MK ref 735WHI. | 1 | £1.00 | BD161 |
| 5A 3 pin rubber/nylon plug tops. | 3 | £1.00 | BD465 |
| 25A rotary switch surface mounting, cover engraved high, medium, low and off. | 1 | $£ 2.00$ | 2 P 123 |
| 30A rotary switch surface mounting with pointer knob. | 1 | £2.00 | 2 P 122 |


| DESCRIPTION | $\begin{gathered} \text { PACK } \\ \text { CUANTITY } \end{gathered}$ | Pack PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| MK splitter 45A switch $3 \times 15 \mathrm{~A}$ fuses. | 1 | £5.00 | 5P100 |
| Switched spur in meta box. | 1 | £1.00 | BD589 |
| Inflex push switch for table lamp etc. | 3 | £1.00 | BD562 |
| Pull switch celling mount with cord and tassle. white. | 1 | £1.00 | BD528 |
| 13A swiched socket and spur on double plate. Brown. | 1 | £1.00 | BD249 |
| 13A fused spur and socket on double plate. White. | 1 | £1.00 | BD302 |
| 13A ring main spur boxes. | 5 | £1.00 | BD2 |
| Cable clips 2.5 mm flat with hardened nails. | 50 | £1.00 | BD577 |
| Brown ceiling roses. (old style) | 5 | £1.00 | BD292 |
| 5A lighting switches brown | 6 | £1.00 | BD452 |
| ES lampholders, bakelite through panel fixing. | 4 | £1.00 | BD476 |
| 5A 3 pin switched surface mount sockets. | 3 | £1.00 | BD194 |
| 5A 3 pin flush sockets. Brown. | 6 | £1.00 | BD193 |
| BC lampholder adaptors, white. | 6 | £1.00 | BD191 |
| Shaver adapiors for 13A sockets. | 2 | £1.00 | BD617 |
| 13A adaptors to take 2 13A plugs. | 3 | $£ 2.00$ | 2P187 |
| 13A plugs with sloeved pins. white. | 3 | £2.00 | 2P186 |
| 13A adaptor, takes 3 13A plugs. | 2 | £3.00 | 3 P 37 |
| Surface mount oblong 5A switches. brown. | 5 | £1.00 | BD4 |
| EMERGENCY LIGHTING (3 HRS) |  |  |  |
| 3 watt $6^{\prime \prime}$ flourescent. cased. | 1 | £30.00 | 30P3 |
| 8 watt 12" flourescent. cased. | 1 | £40.00 | 40P2 |
| 2.5 watt Xenon bulb. cased. | 1 | $£ 20.00$ | 20P10 |
| FANS AND BLOWERS |  |  |  |
| $41 / 2^{\prime \prime} \times 41 / 2^{\prime \prime}$ Muffin type tan 115v (ex-computer). | 1 | $£ 3.00$ | 3P36 |
| Snail type. $6^{\prime \prime} \times 4^{\prime \prime} 240 \mathrm{v} 270 \mathrm{~mA}$. silent and powerful. | 1 | £5.00 | 5P166 |
| 41/2" $\times 41 / 2^{\prime \prime}$ Muffin type fan 230v (ex-computer). | 1 | £5.00 | 5P40 |
| $33 / 4^{\prime \prime}$ square Papst fan 110 v . | 1 | £5.00 | 5P53 |
| $31 / 4$ " square 12 v brushless fan. $120 \mathrm{~mA} \mathrm{PAPST}$. | 1 | £12.00 | 12 P 12 |
| 85 mm square 240v fans. | 1 | £9.00 | 9 P 10 |
| Brushless Muffin fan 18-28v DC. | 1 | £5.00 | 5P60 |
| $5^{\prime \prime}$ Woods extractor fan (ex-equipment) 230 v . | 1 | $£ 5.00$ | 5P41 |
| $5^{\prime \prime}$ alumimium fan blades to fit $1 / 4^{\prime \prime}$ shaft. | 2 | £1.00 | BD86 |
| Mains motor to suit above blades. | 1 | £1.00 | BD88 |
| 18" long tangential blower with motor at 1 end. 230v. | 1 | £10.00 | 10P89 |
| $14^{\prime \prime}$ blower with motor in the middle. 230 v . | 1 | £10.00 | 10P90 |
| 12v DC fan made by Papst $31 / 2^{\prime \prime}$ square. | 1 | £10.00 | 10P33 |
| Mains operated centrifugal blower $5^{\prime \prime} \times 11 / 2^{\prime \prime}$ output. | 1 | £5.00 | 5P99 |
| $41 / 2^{\prime \prime} \times 41 / 2^{\prime \prime}$ Axial fan. Papst 230v. | 1 | £6.00 | 6P6 |
| Plastic fan blades approx $3^{\prime \prime}$ across. | 2 | £1.00 | BD638 |
| $41 / 2^{\prime \prime} \times 41 / 2^{\prime \prime}$ brushless fan 12 v . | 1 | £8.00 | 8P26 |
| $61 / 2^{\prime \prime}$ powerful fan 240 v 210 cu ft min | 1 | £10.00 | 10P67 |
| FEET |  |  |  |
| 38 mm square self adhesive feet. | 8 | £1.00 | 8D726 |


| DESCRIPTION | $\text { PACK } \mid \text { PAANTITY }$ | PACK PAICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| 20 mm square self adhesive feet. | 8 | $\varepsilon 1.00$ | BD891 |
| 12.5 mm square self adhesive feet. | 12 | £1.00 | BD892 |
| FERRITE POTS ETC |  |  |  |
| 1 1" diameter for chokes etc. | 20 | $£ 1.00$ | BD363 |
| Assorted ferrite shapes. | 5 | £1.00 | BD99 |
| Ferrite core $56 \mathrm{~mm} \times 18 \mathrm{~mm}$ E shaped. | 4 | £1.00 | BD156 |
| FLUORESCENT LJGHTING |  |  |  |
| 12" 8 watt fluorescent tube. | 1 | $£ 1.00$ | BD314 |
| Philips W tube 30 watt. | 1. | £1.00 | BD336 |
| Starter for 40-80 watt tube. | 4 | £1.00 | BD92 |
| Starter holders. Standard type. | 4 | £1.00 | BD407 |
| Terry clip for $11 / 2^{\prime \prime}$ tube. | 5 | £1.00 | BD406 |
| 12v cased flourescent light with on/off switch. 8 w. | 1 | £8.00 | 8P48 |
| DISC DRIVES AND DISCS |  |  |  |
| 3 1/2"' 720K drive by NEC. | 1 | $£ 60.00$ | 60P2 |
| Case for $3^{\prime \prime}$ or $31 / 2^{\prime \prime}$ disc drive. ABS type. | 1 | £4.00 | 4P8 |
| 3 " disc for Amstrad etc. | 1 | £3.00 | 3 P 24 |
| 3 1/2" disc. | 2 | $£ 2.00$ | 2P185 |
| $8^{\prime \prime}$ ex equipment disc drives. | 1 | £8.00 | 8P37 |
| $51 / 4^{\prime \prime}$ disc drives 360K brand new. | 1 | £35.00 | 35P5 |
| $51 / 4^{\prime \prime}$ ex equipment 360 K drive .(condition unknown). | 1 | £18.00 | 18 P 5 |
| $51 / 4^{\prime \prime}$ discs unbranded but good quality. | 10 | £5.00 | 5P168 |
| 3 1/2" discs unbranded but good quality. | 15 | £10.00 | 10P88 |
| Disc box holds $10051 / 4^{\prime \prime}$ discs. Lockable. | 1 | £11.00 | 11 P5 |
| Disc box holds $4031 / 2^{\prime \prime}$ discs or CDs. lockable. | 1 | £9.00 | 9 P 4 |
| FUSES AND FUSEHOLDERS |  |  |  |
| 20 mm quick blow 250 mA . | 12 | £1.00 | BD983 |
| 20 mm quick blow 500 mA | 12 | £1.00 | BD984 |
| 20 mm quick blow 1 A. | 12 | $£ 1.00$ | BD985 |
| 20 mm quick blow 2 A . | 12 | £1.00 | BD986 |
| 20 mm quick blow 4A. | 12 | £1.00 | BD987 |
| 32 mm quick blow 250 mA . | 12 | £1.00 | BD988 |
| 32 mm quick blow 500 mA . | 12 | £1.00 | BD989 |
| 32 mm quick blow 1A. | 12 | £1.00 | BD990 |
| 32 mm quick blow 2A. | 12 | £1.00 | BD991 |
| 32 mm quick blow 4A. | 12 | £1.00 | BD992 |
| 1" plug top fuse 2A. | 10 | £1.00 | BD993 |
| $1{ }^{1 \prime}$ plug top fuse 3A. | 10 | £1.00 | BD994 |
| 1 " plug top fuse 5A. | 10 | £1.00 | BD995 |
| 1 " plug top fuse 13A. | 10 | £1.00 | BD996 |
| 20 mm chassis mount fuseholders. | 20 | £1.00 | BD543 |
| Porcelain fuseholders and fuses. | 4 | £1.00 | BD82 |
| 20 mm chassis mount fuseholders. | 10 | £1.00 | BD144 |
| . 5 A 20 mm fast blow fuses. | 20 | £1.00 | BD542 |

## PRIORITY ORDER FORM



## ULTRA SONIC WIRELESS <br> ALARM <br> SYSTEM <br> 

This system consists of a small, cased ultrasonic movement sensor with a range of 5 metres. This is plugged into a standard 13A socket. The alarm unit is similar in design and is plugged into a 13A socket where you want the alarm to sound. Any movement in the room with the sensor sounds the alarm, or, if required the sensor. Transmmision is via the mains supply so the units will work anywhere in the building. A basic system comes complete with one sensor and one alarm unit but it is possible to have up to five detector units working to the same alarm unit.
Price for the basic system $£ 20.00$ ref 20P34
Additional sensors are $£ 11.00$ ref 11P6

RADIO..CONTROLLED..CARS.
We have secured a large parcel of cars which make ideal Christmas prescents.

1. A single channel $R / C$ buggy with forward and turn controls, off road tyres,suspension. Fully guaranteed.
$£ 12.00 \ldots$. ref 12 P 40 .
2. A Full function FERRARI TESTAROSSA A true 2 channel radio controlled car with forward, reverse, steering, 2 gears plus turbo boost and working headlights!.

## £22.00....ref 22P6.....

A pair of very low cost walkie talkies with about 100 ' range Yours for $£ 8.00$
ref 8P50


METAL DETECTOR Fun lightweight device for buried treasure!
33" long Tune and fine tune controls. $£ 10.00$ ref 10P101

VERY FAST NICAD CHARGER This very smart plug in unit carries 4 AA cells and plugs into a 13A socket. Charges $\mathbf{4}$ cells in $\mathbf{1 . 8}$ hrs! Supplied with 4 AA nicads.
£16.00
ref 16P8

NICAD BATTERIES AA SIZE 4 FOR $\mathbf{\$ 4 . 0 0}$ REF 4 P44
D SIZE 4 FOR $£ 9.00$ REF 9P12
C SIZE 2 FOR $\{4.00$ REF 4P73
PP3 9V 1 FOR $\mathbf{1 6 . 0 0}^{0}$ REF 6P35 CHARGER.... 56.00 REF 6P36

| DESCRIPTION | $\begin{aligned} & \text { PACK } \\ & \text { QUANTITY } \end{aligned}$ | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| Woods metal 15A thermal cutout. | 5 | $\varepsilon 1.00$ | BD307 |
| Fuse holder $11 / 4^{\prime \prime}$ surface mount. Bulgin. | 10 | £1.00 | BD526 |
| Panel mount fuseholder 20 mm . | 4 | £1.00 | BD618 |
| 1 1/4" fuse holder. Panel mount. | 5 | £1.00 | BD752 |
| GEARS |  |  |  |
| Gearbox kit contains 18 gears, 4 axles and 12 v motor. GOOSENECKS | 1 | £3.00 | $3 P 93$ |
| Chrome finshed $8^{\prime \prime}$ long with standard fittings. | 1 | E2.00 | 2P345 |
| Chrome finished 12" long with standard fittings. | 1 | £3.00 | 3P116 |
| Chrome finlshed 21" long standard fittings. | 1 | £4.00 | 4P76 |
| Goose neck base plate. 3 hole fixing. | 1 | £2.00 | 2P346 |
| HEATSHRINK SLEEVING |  |  |  |
| 1.6 mm dia shrinking to .8 mm . 1.2 metre length. | 1 | £1.00 | BD997 |
| 2.4 mm dia shrinking to 1.2 mm . 1.2 metre length. | 1 | £1.00 | BD998 |
| 3.2 mm dia shrinking to 1.6 mm 1.2 metre length. | 1 | £1.00 | BD999 |
| 4.8 mm dia shrinking to 2.4 mm . 1.2 metre length. | 1 | £2.00 | 2 P331 |
| 6.4 mm dia shrinking to 3.2 mm . 1.2 metre length. | 1 | £2.00 | 2P332 |
| 9.5 mm dia shrinking to 4.7 mm 1.2 metre length. | 1 | £2.00 | 2 P333 |
| 12.7 mm dia shrinking to 6.4 mm 1.2 metre length. | 1 | $\varepsilon 2.00$ | 2 P 334 |
| HEATSINKS |  |  |  |
| TO220 bolt on heatsink 21 deg C/W. | 5 | £1.00 | CD1 |
| TO3 bolt on heatsink. 6.9 deg C/W. | 2 | £1.00 | CD2 |
| TO5 push on heatsink 50 deg C W. | 4 | £1.00 | CD3 |
| 14" long extruded heatsink. | 1 | £8.00 | 8P3 |
| HEATING |  |  |  |
| 1.2 kw min tangential blow heater $70 \times 45 \mathrm{~mm}$ element. | 1 | £6.00 | 6P54 |
| 600 watt coil heaters air or liquid. 4"x3" 10 year life. | 1 | £3.00 | 3P78 |
| Temperature switches 13A. Contacts open at 120 deg C | 3 | £1.00 | BD745 |
| Time and temp module. Displays either in C or F. 1.5 v | 1 | £9.00 | 9P5 |
| Additional sensor for above unit on long lead. | 1 | £3.00 | 3P60 |
| 2.5 kw tangential blow heater $195 \times 45 \mathrm{~mm}$ element. | 1 | £5.00 | 5P62 |
| Triple interlocked rocker switch for heaters. | 2 | £1.00 | BD270 |
| 3 kw tangential blow heater $300 \times 40 \mathrm{~mm}$ element. | 1 | £8.00 | 8P24 |
| 4 gang rocker switch for 3kw heater. | 1 | £1.00 | BD666 |
| 750 watt standard pencil element 220 mm long. | 2 | $\varepsilon 1.00$ | BD377 |
| 1000 watt standard pencll element 232 mm long. | 2 | £1.00 | BD376 |
| Heating pad approx $8^{\prime \prime} \times 5^{\prime \prime} 200$ watt mains voltage. | 1 | £1.00 | BD112 |
| 80 watt brass encased elements for fridges etc. | 2 | £1.00 | BD8 |
| Pull cord swich as fitted to radiant heaters. | 1 | £1.00 | BD299 |
| 15m heating wire, waterproof. Ideal for pipes eic. Mains. | 1 | £5.00 | 5P109 |
| 2 kw tangential heater. $190 \times 65 \mathrm{~mm}$ element. | 1 | £5.00 | 5P127 |
| Quick cuppa 12 v immersion heater. Cigar lighter plug. HEADPHONES AND INSERTS | 1 | £3.00 | $3 \mathrm{P92}$ |

Stereo headphones 8 ohm 1/4" plug. 600 ohm inserts. Useable as speaker or microphone. Stereo walkman type headphones. Inner ear stereo headphones.
Stereo headphone lead (curly) 1 m long with $1 / 4^{\prime \prime}$ plug. Ex GPO earpieces for intercoms etc. 55 ohm.
Crystal earpieces for crystal radios etc.
Dynamic fullsize stereo h/phones 20-20KHZ 32R imp. PROFESSIONAL LOUDSPEAKERS 270 mm bass/mid range 8R $400 \mathrm{w} 65-6 \mathrm{khz}$. 318 mm dia bass $/ \mathrm{mid}$ range 8R 600 w 390 mm dia bass unit 8R 600w $35-6 \mathrm{khz}$ Inline headphone volume control with 5 m lead.
IC SOCKETS
8 way IC socket.
14 way IC socket.
16 way ICsocket.
18 way IC socket.
20 way IC socket.
24 way IC socket.
28 way IC socket.
40 way IC socket.
INDICATORS AND BULBS
12v lilliput bulbs.
1.5 v 300 mA MES bulb.
2.5 v 200mA MES bulb.
3.5 v 300 mA MES bulb.
6.0 v 60 mA MES bulb.
12.0 v 200 mA MES bulb.

24 v MES bulbs 80 mA 2 watt.
Amber indicators with neons 240 v . Oblong.
Amber neon idicators round 240v.
LED holders 3 mm
LED holders 5 mm .
LED 5 mm red.
LED 5 mm red. (1,000 pack).
LED 5 mm green.
LED 5 mm green. (1,000 pack).
LED 5mm yellow.
LED 5 mm yellow. (1,000 pack).
LED 3 mm red.
LED 3 mm red. (1,000 pack).
LED 3mm green.
LED 3 mm green. (1,000 pack).

| PACK | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: |
| 1 | £2.00 | 2P254 |
| 6 | £1.00 | BD139 |
| 1 | £3.00 | 3P51 |
| 1 | £3.00 | 3P56 |
| 2 | £1.00 | BD230 |
| 4 | £1.00 | BD484 |
| 2 | £1.00 | BD529 |
| 1 | £8.00 | 8P33 |
| 1 | £75.00 | 75P3 |
| 1 | £130 | 130P1 |
| 1 | £110 | 110P1 |
| 1 | £1.00 | BD717 |
| 10 | £1.00 | BD773 |
| 7 | £1.00 | BD774 |
| 7 | £1.00 | BD775 |
| 5 | £1.00 | BD776 |
| 5 | £1.00 | BD777 |
| 4 | £1.00 | BD778 |
| 4 | £1.00 | BD779 |
| 4 | £1.00 | BD780 |
| 5 | £1.00 | BD177 |
| 8 | £1.00 | CD10 |
| 8 | $\varepsilon 1.00$ | CD11 |
| 8 | £1.00 | CD12 |
| 8 | E1.00 | CD13 |
| 8 | £1.00 | CD14 |
| 10 | £1.00 | BD694 |
| 3 | £1.00 | BD179 |
| 6 | £1.00 | BD180 |
| 4 | £1.00 | CD33 |
| 4 | £1.00 | BD518 |
| 15 | £1.00 | BD893 |
| 1 | £44.00 | 44P1 |
| 12 | £1.00 | BD894 |
| 1 | £54.00 | 54P1 |
| 10 | £1.00 | CD5 |
| 1 | £64.00 | 64P1 |
| 15 | £1.00 | CD4 |
| 1 | £44.00 | 44P2 |
| 10 | £1.00 | CD6 |
| 1 | ¢54.00 | 54P2 |


| DESCRIPTION | $\begin{gathered} \text { PACK } \\ \text { OUANTITY } \end{gathered}$ | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| LED 3mm yellow. | 10 | £1.00 | CD7 |
| LED 3mm yellow (1,000 pack). | 1 | $£ 64.00$ | 64P2 |
| FLASHING LED 5 mm red 9-12v DC 3 HZ . | 4 | £2.00 | 2 P335 |
| FLASHING LED 5 mm green 9 -12v DC 3 HZ . | 4 | £2.00 | 2P336 |
| FLASHING LED 8 mm red $9-12 \mathrm{v}$ DC 3 HZ . | 2 | £2.00 | 2 P337 |
| FLASHING LED 8mm green 9-12v DC 3HZ. | 2 | £2.00 | 2P338 |
| High power INFRA-RED source 12mW 1.7 v 5 mm . | 3 | £1.00 | CD8 |
| High power INFRA-RED sensor 5 mm . | 1 | £1.00 | CD9 |
| 3 lamps on a face plate of metal box. $140 \times 75 \times 45 \mathrm{~mm}$. | 1 | £3.00 | 3P16 |
| MES bulb holders. Batten type | 4 | £1.00 | 8 D 895 |
| Minature bulb holders for single contact bulbs. | 4 | £1.00 | 8D896 |
| 3 colour LED. | 2 | £1.00 | 8D611 |
| INFRA RED SENSORS ETC |  |  |  |
| IR 5 metre beam 22-26v DC 250 mA switching. INDUCTIVE PROXIMITY SWITCHES | 1 | $£ 25.00$ | 25P15 |
| 98 mm barrel type $10-36 \mathrm{v} 10 \mathrm{~mm}$ range with LED. INSULATORS AND CABLE GRIPS | 1 | £12.00 | 12P19 |
| PVC grommets for $3 / 8^{*}$ cable. | 100 | £1.00 | 8D181 |
| Cable grips for up to $3 / 8^{\prime \prime}$ cable. | 10 | £1.00 | BD431 |
| Cable grips for $3 / 8^{\prime \prime}$ cable. | 20 | £1.00 | 8D432 |
| KETTLE ELEMENTS |  |  |  |
| Old type with 2 round pins and empty ejector. KEY BOARDS | 1 | £1.00 | 8D395 |
| 84 key keyboards uncased with control PC8. | 1 | $£ 3.00$ | 3P89 |
| HI FI SPEAKER UNITS |  |  |  |
| 131 mm dia full range 8 R $60 \mathrm{~W} 63-20 \mathrm{khz}$. | 1 | £12.00 | 12P33 |
| 242 mm dia wooter 8R $60 \mathrm{w} 45-5.5 \mathrm{khz}$. | 1 | £40.00 | 40P9 |
| 154 mm dia dome mid range 8R 100w 300-10khz. | 1 | £25.00 | 25P22 |
| 106 mm dia titanlum tweeter $8 \mathrm{r} 100 \mathrm{w} 2-22 \mathrm{khz}$. | 1 | £14.00 | 14P5 |
| 3 way crossover for above 3 units ( $700 \& 5 \mathrm{khz} 100 \mathrm{watt}$ ) | 1 | £12.00 | 12P34 |
| Set of $2 \times 40 \mathrm{P} 9,2 \times 25 \mathrm{P} 22,2 \times 14 \mathrm{P} 5,82 \times 12 \mathrm{P} 34$ | 1 | £150 | 150P1 |
| LOUDSPEAKERS AND GRILLS |  |  |  |
| Set of 3 speaker grills $140 \mathrm{~mm}, 70 \mathrm{~mm}$ and 28 mm . Black. | 1 | £1.00 | BD737 |
| Pair of white plastic speaker grills $6^{\prime \prime}$ diameter. | 1 | £1.00 | BD904 |
| Flush mounting speaker grill for $8^{\prime \prime}$ speaker. | 1 | £2.00 | 2P281 |
| Tweeter on chrome mounting plate. | 1 | £3.00 | 3P68 |
| $7^{\prime \prime} \times 5^{\prime \prime} 8$ ohm speaker with built in tweeter. 5 watt. | 2 | £3.00 | 3 P 69 |
| $8^{\prime \prime}$ round 8 ohm 60 watt full range speaker. Ali coil. | 1 | £12.00 | 12P14 |
| $5^{\prime \prime}$ diameter full range speaker 30 watt + HF cone 8R. | 1 | £8.00 | 8P49 |
| $3^{\prime \prime}$ diameter full range speaker 20 watt+HF cone. 8R. | 1 | £5.00 | 5P183 |
| Pair of 70 w per channel 3 way car speakers. | 1 | £28.00 | 28P1 |
| Pair of 100 w per channel 3 way car speakers. | 1 | £30.00 | 30P7 |
| $6^{\prime \prime} \times 9^{\prime \prime} 8$ ohm 15 watt speaker. | 1 | £3.00 | 3P76 |


| DESCRIPTION | $\begin{array}{\|l\|l\|} \hline \text { PACK } \\ \text { QUATITY } \end{array}$ | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| $6^{\prime \prime} \times 4^{\prime \prime} 16$ ohm 5 watt rating. | 2 | £1.00 | BD243 |
| $61 / 2^{\prime \prime} 4$ ohm speaker 10 watt rating. | 1 | £1.00 | BD137 |
| 2 1/4" 60 ohm speaker. | 2 | £1.00 | BD453 |
| $21 / 4^{\prime \prime} 8$ ohm speaker. | 2 | £1.00 | BD454 |
| $6^{\prime \prime} \times 4^{\prime \prime} 15$ ohm speaker 10 watt rating. | 1 | £2.00 | 2P167 |
| $3^{\prime \prime} 4$ ohm tweeter. | 1 | £1.00 | BD433 |
| 40 watt 3 way crossover. | 1 | £1.00 | BD23 |
| 25 watt cross over for woofer and tweeter. | 2 | £1.00 | BD22 |
| Extension speaker cabinet with back and front ( $5^{\circ \prime}$ ). | 1 | £1.00 | BD118 |
| $41 / 2^{\prime \prime}$ round mid range speker 4 ohm 10 watt. | 1 | $£ 2.00$ | 2P195 |
| 110 db horn/speaker. | 1 | £4.00 | 4P60 |
| Personal mini speaker. Plugs straight into cassette. | 1 | £4.00 | 4P50 |
| TV speakers 3 watt 8 ohm $70 \times 55 \mathrm{~m}$. | 2 | £3.00 | 3P108 |
| TV speakers 5 watt 4 ohm $55 \times 125 \mathrm{~mm}$. | 2 | £3.00 | 3 P 109 |
| Loud speaker wall mounting brackets. (pair). | 1 | £5.00 | 5P152 |
| $5^{\prime \prime} \times 3^{\prime \prime} 16$ ohm speaker 5 watt. | 1 | £1.00 | BD725 |
| Mylar waterproof cone speaker 3 1/4" sq 35R 2 watt. KNOBS ETC | 1 | £1.00 | BD903 |
| Pointer knob for flatted spindie. | 4 | £1.00 | BD295 |
| 1" collet knobs, black. | 4 | £1.00 | BD525 |
| Solid aluminium $11^{\prime \prime \prime}$ dia. Grub screw fixing on $1 / 4^{\prime \prime}$ sft. LASERS | 2 | £1.00 | BD720 |
| Philips 2 mw Helium Neon laser. 260x37mm. | 1 | £40.00 | 40P10 |
| Mains ABS cased power supply kit for laser. | 1 | £20.00 | 20P33 |
| 12v universal supply kit (uncased). | 1 | £25.00 | 25P13 |
| Plastic case with PSU kit big enough to hold tube as well | 1 | £22.00 | 22 P 3 |
| Boxed and built laser. | 1 | £75.00 | 75P4 |
| MAGNETS |  |  |  |
| Flat magnet $1^{\prime \prime} \times 1 / 2^{\prime \prime} \times 1 / 8^{\prime \prime}$. | 6 | £1.00 | BD897 |
| Very powerful magnet. $25 \times 13 \times 6 \mathrm{~mm} u$ shape. MEMORY JOGGER | 2 | £1.00 | BD642 |
| Clockwork ringer, setable 0-60 mins then rings bell. MICROPHONES AND STANDS | 1 | £1.00 | BD77 |
| 600 ohm microphone/speaker inserts. 35 mm dia $\times 10 \mathrm{~mm}$ | 4 | £1.00 | BD139 |
| Dynamic mic inserts. Mylar cone 200 ohm imp. | 2 | £1.00 | BD762 |
| Unidirectional electret condenser mic. top quality 600 r . | 1 | £15.00 | 15P28 |
| Dynamic handheld mic with stand. Cassette type. | 1 | £1.00 | BD305 |
| Hand held dynamic mic with on/off switch. | , | $£ 1.00$ | BD711 |
| Microphone stand. Chrome plated magnetic base. | 1 | $£ 5.00$ | 5P154 |
| FET electret capacitor mic capsule 1.5 v 500 ohm . | 1 | £1.00 | BD646 |
| Crystal microphone 25 mm dia high gain. | 2 | £1.00 | BD729 |
| MICROSWITCHES |  |  |  |
| V3 size 15A 250v c/o button operated. Push on tags. | 3 | £1.00 | BD341 |


| DESCRIPTION | PACK QUANTITY | PACK PRICE | ORDER number |
| :---: | :---: | :---: | :---: |
| V3 size 10A 250v c/o button operated. Push on tags. | 4 | £1.00 | BD340 |
| V3 size 10A $250 \mathrm{c} / 0$ lever operated. Push on tags. | 3 | £1.00 | BD342 |
| Extra thin $11 / 4^{\prime \prime} \times 7 / 8^{\prime \prime} \times 1 / 4^{\prime \prime} 10 \mathrm{~A}$ contacts. | 4 | $\varepsilon 1.00$ | BD403 |
| Large micro switch with $21 / 2^{\prime \prime}$ lever. | 1 | £1.00 | BD591 |
| Subminature microswitches. Assorted. | 5 | £1.00 | BD313 |
| MISCELLANEOUS |  |  |  |
| Musical sounder for cards. | 1 | $£ 1.00$ | BD328 |
| Cabinet locking mechanism with 2 keys. | 1 | £1.00 | BD55 |
| Personal stereo innards. Tape mech and head. | 1 | £1.00 | BD763 |
| Mullard thyristor trigger module. | 1 | £1.00 | BD63 |
| Magnetic brake assembly. Instant stop. | 1 | £1.00 | BD66 |
| Clear plastic lense $11 / 4^{\prime \prime}$ diameter. (phone type). | 25 | £1.00 | BD124 |
| Satety cover for twin 13A outlet. | 6 | £1.00 | BD149 |
| Safety covers for single 13A socket. | 5 | £1.00 | BD150 |
| Hoseclips, collspring type. 3 different sizes. | 20 | £1.00 | BD222 |
| Motor startstop switch, skeleton type with trip. | 2 | £1.00 | BD254 |
| CB coverter converts car radio into AM CB radio. | 1 | £4.00 | 4P48 |
| Push on $1 / 4^{\prime \prime}$ tags with leads. | 50 | £1.00 | BD259 |
| Round pin kettle plug with moulded lead. | 1 | £1.00 | BD316 |
| Through panel cable grips. Adjustable size. | 10 | £1.00 | BD431 |
| Personal stereos with headphones. (customer returns). | 1 | $£ 3.00$ | 3 P83 |
| Therocouple for measuring internal heat. | 1 | £2.00 | 2P137 |
| Chrome disc $21 / 2^{\prime \prime}$ diameter, curved with centre hole. | 10 | £1.00 | BD521 |
| Hoover programmer 3224. | 1 | £1.00 | BD479 |
| Clear lacquer in an aerosol can. | 1 | £1.00 | BD660 |
| Inspection lamp. Mains. | 1 | £4.00 | 4P31 |
| 2764 eprom BBC compatable. | 1 | £3.00 | 3P48 |
| Ceramic insulating beads. Fit 20 swg wire. | 100 | £1.00 | BD690 |
| Insulating tubing made of Paxolin $6^{\prime \prime} \times 1 / 4^{\prime \prime}$. | 6 | £1.00 | BD691 |
| MONITORS |  |  |  |
| 9" monitor, black and white. Uncased. | 1 | $£ 20.00$ | 20P26 |
| $12^{\text {n }}$ amber high res monitor requires 12 v 1.5 A , sep syncs. TTL input. | 1 | £22.00 | 22P2 |
| Metal case for $9^{\prime \prime}$ monitor. | 1 | £12.00 | 12P3 |
| Kit to convert comp video into separate H sync, Vsync. | 1 | £8.00 | 8P39 |
| AC MOTORS |  |  |  |
| $11 / 2^{\prime \prime}$ stack double ended very powerful motor. | 1 | $£ 2.00$ | 2P55 |
| $3^{\prime \prime}$ square shaded pole motor 24 v AC. | 1 | £2.00 | 2 P 266 |
| 240 v AC 25 watt 3000 rpm motor $6^{\prime \prime} \times 4^{\prime \prime} \times 3^{\prime \prime}$. | 1 | £4.00 | 4P54 |
| $3^{\prime \prime}$ square shaded pole motor. 240 v AC. | 1 | £2.00 | 2P265 |
| Mains shaded pole motor $7 / 8^{\prime \prime}$ stack. | 1 | £1.00 | BD85 |
| Papst motor open type cap start. | 1 | $£ 2.00$ | 2P71 |
| EMI tape motor, 2 speed reversible. 93 mm dia 6 mm shatt | 1 | $£ 2.00$ | 2P70 |
| Precision motor for disc or tape. | 1 | $£ 2.00$ | 2P12 |


| DESCRIPTION | $\begin{aligned} & \text { PACK } \\ & \text { QUANTITY } \end{aligned}$ | $\begin{aligned} & \text { PACK } \\ & \text { PAICE } \end{aligned}$ | OROER NUMBER |
| :---: | :---: | :---: | :---: |
| MOTORS WITH GEARBOXES |  |  |  |
| 200 rpm 2 w malns motor. 46 mm dia $\times 18 \mathrm{~mm}$ worm drive. | 2 | £1.00 | BD175 |
| 4 rpm mains motor 2 w (suftable mirror ball etc). | 1 | $£ 2.00$ | 2 P 17 |
| 1 rev per hour 2 watt extra small mains motor. With cog. | 2 | £1.00 | BD500 |
| 500 watt mains motor and 3 speed control. | 1 | $£ 5.00$ | 5P193 |
| 5 rpm 60 watt motor with gearbox. Powerful. $125 \times 60 \times 45$. | . 1 | $£ 5.00$ | 5P54 |
| 16 rpm mains 2 watt motor with gearbox. 50 mm dia $\times 30$. | 1 | £1.00 | BD91 |
| 200 rpm mains 60 watt motor with gearbox. $60 \times 80 \times 90 \mathrm{~mm}$ | 1 | £2.00 | 2 P 38 |
| 6 rpm 60 watt mains motor and gearbox. | 1 | $£ 5.00$ | 5P74 |
| 5 rpm Crouzet type motor 240v. | 1 | £4.00 | 4P63 |
| Micro wave turntable motors. Buill in weight sensor. 240 y . | 1 | £5.00 | 5P169 |
| 60 rpm 60 watt mains motor. | 1 | £5.00 | 5P173 |
| MOTORS DC |  |  |  |
| 1/6th HP 12 v motor. $75 \times 80 \mathrm{~mm}$ spindle $36 \times 8 \mathrm{~mm}$. | 1 | $£ 8.00$ | 8P14 |
| $1 / 10$ th HP 12v motor $1 / 4^{\prime \prime}$ spindle. $75 \times 65 \mathrm{~mm}$. | 1 | $£ 4.00$ | 4 P 22 |
| $1 / 8 \mathrm{th} \mathrm{HP} 12 \mathrm{v}$ motor $75 \times 75 \mathrm{~mm}$ spindle $25 \times 8 \mathrm{~mm}$. | 1 | £6.00 | 6 P 1 |
| 1/3rd HP 12v motor (Sinclair C5 ). 180x100mm 3300 rpm | - 1 | £20.00 | 20 P 22 |
| C5 motor complete with 4:1 reduction box. | 1 | £40.00 | 40P8 |
| Electronic speed controller kit for C5 motor or equiv. | 1 | £17.00 | 17P3 |
| Model motor $1.5 \mathrm{v}-9 \mathrm{~V}$ (speed is voltage dependent). | 1 | £1.00 | BD540 |
| 3 c cassette motor. Very low current. | 1 | £1.00 | BD681 |
| MOTORS STEPPER |  |  |  |
| Stepper motor 7.5 deg step $10-14 \mathrm{v} 27$ ohm $70 \times 40 \mathrm{~mm}$ MOTORS 1-12V | 1 | $£ 5.00$ | 5P81 |
| Model aircraft motors. Spin to start. $12 \times 20 \mathrm{~mm}$. | 10 | £1.00 | BD134 |
| Low current motor for working with solar cells. | 1 | £1.00 | BD681 |
| NOISE FILTERS |  |  |  |
| Chassis mounting noise fitter. $45 \times 30 \times 16 \mathrm{~mm}$. | 1 | £2.00 | 2 P 225 |
| IEC filtered chassis socket. | 1 | £3.00 | 3 P 50 |
| OPTO |  |  |  |
| Sub-min light dependent resistor. | 2 | £1.00 | BD19 |
| Camera flash units. Contains Xenon tube etc 3vop. | 1 | $£ 2.00$ | 2 P 38 |
| Slotted opto photo transistor. | 2 | £1.00 | BD14 |
| Slotted opto interupted switch | 1 | £1.00 | BD545 |
| Light dependent resistor ORP12. | 1 | £1.00 | BD619 |
| PANEL METERS |  |  |  |
| 270 deg movement panel meter. 3 1/2" diameter. | 1 | £3.00 | 3 P 87 |
| 100uA panel meter $11 / 4^{\prime \prime} \times 1 / 2^{\prime \prime}$ with scale illümination. | 1 | £1.00 | BD700 |
| $0-40 \mathrm{v}$ panel meter $80 \times 70 \mathrm{~mm}$. | 1 | £6.00 | 6P24 |
| $0-50 \mathrm{v}$ panel meter $80 \times 70 \mathrm{~mm}$. | 1 | £6.00 | 6 P 25 |
| $0-80 \mathrm{v}$ panel meter $80 \times 70 \mathrm{~mm}$. | 1 | £6.00 | 6 P 26 |
| $0-160 \mathrm{v}$ panel meter $80 \times 70 \mathrm{~mm}$. | 1 | £6.00 | 6 P 27 |
| $0-200 \mathrm{v}$ panel meter $80 \times 70 \mathrm{~mm}$. | 1 | £6.00 | 6P28 |
| $0-10 \mathrm{~A}$ panel meter $80 \times 70 \mathrm{~mm}$. | 1 | £6.00 | 6 P 29 |


| DESCRIPTION | $\begin{aligned} & \text { PACK } \\ & \text { OUANTITY } \end{aligned}$ | $\begin{aligned} & \text { PACK } \\ & \text { PAICE } \end{aligned}$ | OROER NUMBER |
| :---: | :---: | :---: | :---: |
| 0-5A panel meter $80 \times 70 \mathrm{~mm}$. | 1 | $£ 6.00$ | 6P30 |
| 45-55 HZ frequency Indicator. | 1 | £15.00 | 15P19 |
| $55-65$ vibrating reed frequency indicator. | 1 | £15.00 | 15P18 |
| 200uA panel meter $43 / 4^{\prime \prime} \times 2$ 1/2". | 1 | £10.00 | 10P24 |
| 1 mA panel meter $6^{\prime \prime} \times 31 / 2^{\prime \prime}$. | 1 | £10.00 | 10P41 |
| 0-100uA panel meter. Scale separate. | 1 | £4.00 | 4P32 |
| 100-0-100uA panel meter. Scale separate. | 1 | £4.00 | 4P67 |
| 1 mA panel meter. Scale separate. | 1 | £4.00 | 4P68 |
| VU meter $11 / 2^{\prime \prime}$ square. | 2 | £1.00 | BD366 |
| High quality 50uA panel meter. Moveable scales! | 1 | £3.00 | 3 P81 |
| PANEL METERS $60 \times 45 \mathrm{MM}$ |  |  |  |
| 0-50A 2K3 internal resistance. | 1 | £6.00 | 6P39 |
| 0-100A 1 K2 Internal resistance. | 1 | £6.00 | 6P40 |
| 0-500A 360R internal resistance. | 1 | £6.00 | 6P41 |
| 0-1mA 100R internal resistance. | 1 | $£ 6.00$ | 6P42 |
| 0-10mA 60R internal resistance. | 1 | £6.00 | 6 P 43 |
| 0-100mA 0.6R intemal resistance. | 1 | £6.00 | 6P44 |
| 0-1A | 1 | £6.00 | 6P45 |
| 0-25v DC 25K internal resistance. | 1 | £6.00 | 6P46 |
| 0-30v DC 30K intemal resistance | 1 | £6.00 | 6P47 |
| PCB PRODUCTION EQUIPMENT |  |  |  |
| PHOTO ETCH PCB (UV sensitive) |  |  |  |
| $100 \times 160 \mathrm{~mm}$ single sided. | 1 | £3.00 | 3P133 |
| $203 \times 144 \mathrm{~mm}$ single sided. | 1 | £4.00 | 4P78 |
| $100 \times 160 \mathrm{~mm}$ double sided. | 1 | £4.00 | 4P79 |
| $203 \times 144 \mathrm{~mm}$ double sided. | 1 | £5.00 | 5P189 |
| Ferric chloride etchant (makes 1 litre). | 1 | £3.00 | 3P134 |
| Developer crystals (makes 1 litre). | 1 | £2.00 | 2 P 348 |
| Polypropylene trays. $325 \times 225 \times 50 \mathrm{~mm}$. | 1 | $\Sigma 2.00$ | 2 P 349 |
| Brass wire brush. | 1 | £1.00 | BD837 |
| Transfer starter pack 12 different sheets. | 1 | £10.00 | 10P100 |
| Light box kit for PCB layout etc. | 1 | £16.00 | 16P7 |
| PLUGS AND SOCKETS |  |  |  |
| 3.5 mm mono jack sockets. Chassis or PCB mount. | 8 | £1.00 | BD697 |
| Chassis mount BNC socket. (4 hole fixing). | 2 | £1.00 | BD851 |
| Bulgin 3 pin plug and socket $11 / 2^{\prime \prime}$ diameter. | 2 | £1.00 | BD715 |
| Earphone distribution panel ( $3 \times 3.5 \mathrm{~mm}$ sockets+led). POTS-VARIABLE RESISTORS | 1 | £1.00 | BD727 |
| Standard size 500 K with DP switch $1 / 4^{\prime \prime}$ shaft. | 4 | £1.00 | BD245 |
| Minl volume controls. 1/4" shaft. Assorted valus. | 10 | £1.00 | BD109 |
| Slider pots. Various values. | 10 | £1.00 | BD110 |
| 10 turn 100 ohm 3w 1/4" shaft. | 1 | £1.00 | BD291 |
| 100k multiturn knobs with Integeral knob. | 5 | £1.00 | BD143 |
| Wirewound pots (4 different values with knobs) | 4 | £1.00 | BD71 |


| DESCRIPTION | $\begin{array}{c\|} \text { PACK } \\ \text { OUNVTITY } \end{array}$ | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| $500 \mathrm{hm} \mathrm{3w} \mathrm{wirewound}$. | 4 | $£ 1.00$ | BD73 |
| 8 ohm 25 w 1/4" shaft. | 2 | $£ 1.00$ | BD69 |
| 1000 ohm 25 w $1 / 4^{\prime \prime}$ shaft. | 2 | $\Sigma 1.00$ | BD70 |
| Minature pot $1 / 4^{\prime \prime}$ shatt 50K. | 4 | £1.00 | BD781 |
| Minature pot $1 / 4^{\prime \prime}$ shatt 200 K . | 4 | £1.00 | BD782 |
| Minature pot $1 / 4^{\prime \prime}$ shaft 3 M . | 4 | $\Sigma 1.00$ | BD783 |
| Stereo gang pot $1 / 4^{\prime \prime}$ shaft 50K. | 3 | £1.00 | BD784 |
| Stereo gang pot 1/4" shaft 200K. | 3 | £1.00 | BD785 |
| Quad pot $1 / 4^{\prime \prime}$ shaft 50K. | 4 | £1.00 | BD447 |
| POWER SUPPLIES |  |  |  |
| AZTEC SWITCHED MODE |  |  |  |
| +5v at 3.75A, +12v at $1.5 A$ and $-12 v$ at 0.4A |  |  |  |
| PCB size is $80 \mathrm{~mm} \times 165 \mathrm{~mm}$. Brand new. | 1 | £12.00 | 12 P 39 |
| Case for above PSU with IEC filtered inlet \&switch. | 1 | £5.00 | 5P190 |
| Customer retumed switched mode PSU's for repair. | 1 | £2.00 | 2 P292 |
| Switched mode PSUs ex-equipment $+5,+12$ \& $-12 v$. | 1 | 88.00 | 8 P 36 |
| In car PSU $12 v$ in $3,4.5,6,7.5,9,12 v 800 \mathrm{~mA}$ out, Cig plug | 1 | £5.00 | 5P167 |
| 24 v with sep channels for stereo use. Max 20 watt. | 1 | ¢2.00 | 2 P 4 |
| Switched mode PSU $+12 \mathrm{v},-12 \mathrm{v},+5 \mathrm{v},-5 \mathrm{v}$. 100 v input. | 1 | $\underline{1} 9.00$ | 9 P 2 |
| 4.5100 mA PSU in case with lead. | 1 | £1.00 | BD104 |
| 6 v 200 mA PSU in case with lead. | 1 | £1.00 | 8D103 |
| 6 V 700 mA PSU in case. | 1 | £1.00 | BD899 |
| 8-12v variable PSU. | 1 | £2.00 | 2P3 |
| 12 v 200 mA PSU in case for 13A socket. | 1 | £2.00 | 2 P 114 |
| 9 v 500 mA PSU in case for 13A socket. | 1 | $£ 2.00$ | 2 P 134 |
| 9 v 350 mA AC PSU in case for 13A socket. | 1 | $£ 1.00$ | BD566 |
| $24 v$ PSU chassis with all components. | 1 | £2.00 | 2P150 |
| 9 v AC 100 mA in case with lead. | 1 | £1.00 | BD733 |
| 9 v AC 100 mA in case for 13A socket. | 1 | £1.00 | BD900 |
| 15 v 500 ma DC power supply. | 1 | £2.00 | 2P289 |
| 9 v AC 300 mA in case for shaver socket. | 1 | £1.00 | BD901 |
| PRESSURE SWITCHES |  |  |  |
| Brass pressure switch set for 8psi but adjustable to 15 psi | 1 | £2.00 | 2P92 |
| Switch with 3 different operating pressures. Can be mouth operated. $85 \times 30 \mathrm{~mm}$. | 1 | $£ 1.00$ | BD67 |
| PROJECTS AND KITS |  |  |  |
| Electronic spaceship kit responds to sound complete |  |  |  |
| with wheels and body. | 1 | £10.00 | 10 P 81 |
| Surface mount kit. Makes mini micro amp. | 1 | £7.00 | 7 P 15 |
| Surface mount solder. | 1 | £12.00 | 12 P 18 |
| FM bug kit with PCB embedded coil and sub min mic. | 1 | £5.00 | 5P158 |
| Stabilized PSU kit 3-30v 2A. Excluding case. | 1 | £20.00 | 20P25 |
| Solar powered helicopter kit. | 1 | $\varepsilon 9.00$ | 9P6 |
| Solar powered aeroplane kit. | 1 | £9.00 | 9P7 |


| DESCRIPTION | PACK OUANTITY | PACK PAICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| Solar powered gramophone kit. | 1 | $£ 9.00$ | 9P8 |
| 12-220v inverter kit 15 watt. | 1 | £12.00 | 12P17 |
| 12-220v inverter kit 100 watt. | 1 | £20.00 | 20P27 |
| Microwave tester 9v. | 1 | £6.00 | 6P19 |
| Mini radio module. Superhet with aerial $2^{\prime \prime}$ square. 9v. | 1 | £1.00 | BD716 |
| Airband receiver with 250 page aircraft guide. | 1 | £14.00 | 14P1 |
| Claudivs speech box ( 64 phrases or words) | 1 | $£ 30.00$ | 30P10 |
| Telephone autodialler (state no reqd)BT approved. | 1 | £12.00 | 12P23 |
| 25 watt stereo amplifier IC plus diagram. STK043. | 1 | £4.00 | 4P69 |
| Geiger counter kit 9v operation. | 1 | £39.00 | 39P1 |
| Educational kits, Ohms law and chemical balance kits. | 2 | £1.00 | BD902 |
| Powerful ionizer mains operated kit with case. | 1 | £18.00 | 18P2 |
| 3 channel sound to light kit 750w /channel with case. | 1 | £15.00 | 15P6 |
| Drill speed controller kit. Max 750 watt. PUMPS | 1 | £4.00 | 4P40 |
| Drill operated pump. Fits any drill. | 1 | £3.00 | 3P140 |
| WashIng machine pump. mains. | 1 | £5.00 | 5P18 |
| 350 psi oil pump. Heavy duty. | 1 | £10.00 | 10P2 |
| RECTIFIERS AND DIODES |  |  |  |
| 5A stud rectifiers 250v. | 5 | £1.00 | BD78 |
| 1A 1000v diodes. | 10 | £1.00 | BD471 |
| Minature glass diode for low voltage use. (IN4148). | 15 | £1.00 | BD576 |
| Bridge rectitier 600v 3A. | 1 | £1.00 | BD546 |
| Rectifier 35A 60v. | 1 | £2.00 | 2P179 |
| RELAYS |  |  |  |
| PCB mount relay 5 v coil $2 \mathrm{C} / \mathrm{O} 2 \mathrm{~A}$ contacts. $26 \times 18 \times 17$. | 1 | £1.00 | BD665 |
| Mains operated relay. Single 8A C/O . $45 \times 30 \times 32 \mathrm{~mm}$ | 1 | £1.00 | BD486 |
| Mains operated relay. 48 A C/O contacts. $60 \times 50 \times 30 \mathrm{~mm}$ | 1 | $£ 2.00$ | 2P144 |
| 48 v low profile relay. Twin C/O contacts. $27 \times 22 \times 10 \mathrm{~mm}$ | 2 | £1.00 | BD225 |
| 24 v relay in transistor package $2 \times$ C/O contacts. | 1 | £1.00 | BD748 |
| 3 v reed relay kit. 4 coils and 4 reeds. | 1 | £1.00 | BD145 |
| 6 v reed switch relays. Normally open contacts. | 2 | £1.00 | BD48 |
| Utra small 12 v relay. 3A contacts. $10 \times 17 \mathrm{~mm}$. | 1 | £1.00 | BD205 |
| 12 v water resistant relay. | 1 | £1.00 | BD154 |
| 12 v sealed relay 2 CIO contacts. $20 \times 10 \times 22 \mathrm{~m}$. | 1 | £1.00 | BD311 |
| 12 v minature relay 700 ohm coil. $2 \mathrm{C} / \mathrm{O}$ contacts. | 1 | £1.00 | BD51 |
| 12v minature relay. $4 \mathrm{C} / \mathrm{O}$ contacts. 160 hm . | 1 | £1.00 | BD52 |
| $12 v$ DC or $24 v$ AC plug in relay. 3 C/O contacts. | 2 | £1.00 | BD50 |
| 24 v minature relay. 4 CIO 5 A contacts. $28 \times 20 \times 32 \mathrm{~mm}$. | 2 | £1.00 | BD580 |
| 24 v minature relay. 2 C/O 5A contacts $28 \times 25 \times 10 \mathrm{~mm}$. | 2 | £1.00 | BD850 |
| 1.5 v relay 16 ohm $2 \mathrm{C} / \mathrm{O}$ contacts. $23 \times 18 \times 30 \mathrm{~mm}$. | 1 | £1.00 | BD512 |
| 2 v reed relay. Normally closed contacts. $30 \times 12 \times 9 \mathrm{~mm}$ | 2 | £1.00 | BD549 |
| 3.5 v mini PCB mount relay. $2 \mathrm{C} / \mathrm{O} 3 \mathrm{~A}$ contacts. | 1 | £1.00 | BD548 |
| Assorted motor start relays. | 4 | £1.00 | BD535 |


| DESCRIPTION | $\begin{array}{\|c\|} \text { PACK } \\ \text { OUANTITY } \end{array}$ | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| 50 v AC ex-equip relay. Plug in. 3 C/O 8A contacts. | 4 | £1.00 | BD578 |
| 50 v latching relay. $2 \mathrm{C} / \mathrm{O}$ 5A contacts. | 1 | £1.00 | BD501 |
| Mains relay 4 pole C/O 8A. Complete with base. | 1 | £2.00 | 2P241 |
| RESISTORS $1 / 4$ WATT $2 \%$ METAL FILM |  |  |  |
| $10 \times 50$ values (500 resistors in all). | 1 | £5.00 | 5P170 |
| RESISTORS 2.5 WATT WIRE WOUND |  |  |  |
| 2.5 watt . 068 ohm | 15 | £1.00 | BD830 |
| 2.5 watt . 075 ohm | 15 | £1.00 | BD831 |
| 2.5 watt . 33 ohm | 15 | £1.00 | BD832 |
| 2.5 watt . 44 ohm | 15 | £1.00 | BD833 |
| 2.5 watt 1 ohm | 15 | £1.00 | 8D834 |
| 2.5 watt 1.8 ohm | 15 | £1.00 | BD835 |
| 2.5 watt 2 ohm | 15 | £1.00 | BD836 |
| 2.5 watt 6.9 ohm | 15 | £1.00 | BD838 |
| 2.5 watt 16 ohm | 15 | $£ 1.00$ | BD839 |
| 2.5 watt 51 ohm | 15 | £1.00 | BD840 |
| 2.5 watt 150 ohm | 15 | £1.00 | BD842 |
| 2.5 watt 620 ohm | 15 | £.1.00 | BD844 |
| 2.5 watt 820 ohm | 15 | £1.00 | BD845 |
| 2.5 watl 4K7 ohm | 15 | £1.00 | BD847 |
| 2.5 watt 6 K 8 ohm | 15 | £1.00 | BD848 |
| 2.5 watt 8K2 ohm | 15 | £1.00 | BD849 |
| RESISTORS 5 WATT WIRE WOUND |  |  |  |
| 5 watt 1 ohm | 13 | £1.00 | BD883 |
| 5 watl 6.8 ohm | 13 | £1.00 | BD884 |
| RESISTORS 6 WATT WIRE WOUND |  |  |  |
| 6 watt . 09 ohm. | 12 | £1.00 | BD786 |
| 6 watt . 15 ohm | 12 | £1.00 | BD787 |
| 6 watl . 24 ohm | 12 | £1.00 | BD788 |
| 6 watt . 27 ohm | 12 | £1.00 | BD789 |
| 6 watt .33 ohm | 12 | £1.00 | BD790 |
| 6 watt .5 ohm | 12 | £1.00 | BD791 |
| 6 watl .68 ohm | 12 | £1.00 | BD792 |
| 6 watt 1 ohm | 12 | £1.00 | BD793 |
| 6 watt 1.2 ohm | 12 | £1.00 | BD794 |
| 6 watt 1.8 ohm | 12 | £1.00 | BD795 |
| 6 watt 2 ohm | 12 | ¢1.00 | BD796 |
| 6 watt 2.7 ohm | 12 | £1.00 | BD797 |
| 6 watt 3.3 ohm | 12 | £1.00 | BD798 |
| 6 watt 3.9 ohm | 12 | £1.00 | BD799 |
| 6 watt 12 ohm | 12 | £1.00 | BD800 |
| 6 watt 18 ohm | 12 | £1.00 | BD801 |
| 6 watt 75 ohm | 12 | $£ 1.00$ | BD803 |
| 6 watt 100 ohm | 12 | £1.00 | BD804 |

BULL ELECTRICAL 0273-203500 20

## PRIORITY ORDER FORM

TO BULLELECTRICAL 250 PORTLAND ROAD HOVE SUSSEX BN3 5QT QUANTITY DESCRIPTION

FROM


PERSONAL SAFETY ALARM WITH BUILT IN TORCH. A useful device for home, office or travel! Requires 3 AA batteries. Price is $£ 3.00$ ref $\mathbf{3 P 1 3 5}$


## HELPING HANDS

A useful device for any workshop. complete with 2 'hands' and a magnifier. $\mathbf{£ 4 . 0 0}$ ref $\mathbf{4 P 8 0}$


WEIR SMM300S SWITCHED MODE POWER SUPPLY.
+5 @ 40A, +12 to +24v (adj) 6A +12 to $+15 \mathrm{v}(\mathrm{adj}) 12 \mathrm{~A}$
+5 tol5v(adj)3A. £30.00 ref 30P14

$5^{\prime \prime}$ SIDE CUTTERS
Top quality cutters capable of cutting wires as fine as a hair.
Pressed steel construction, hardened and tempered with sprung jaws.
Price is $£ 2.00$ a pair. ref $\mathbf{2 P 3 5 0}$


### 0.5KG RESIN CORED 22SWG SOLDER

 £4.00 A REEL ref 4P7212 VOLT SOLDERING IRON COMPLETE WITH CIGAR LIGHTER PLUG AND LEAD. 15 WATT. £3.00 REF 3P136

| DESCRIPTION | $\left\lvert\, \begin{gathered} \text { PACK } \\ \text { OUANTITY } \end{gathered}\right.$ | $\begin{aligned} & \text { PACK } \\ & \text { PRICE } \end{aligned}$ | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| 6 watt 400 ohm | 12 | £1.00 | BD810 |
| 6 watt 560 ohm | 12 | £1.00 | BD811 |
| 6 watt 1K ohm | 12 | £1.00 | BD812 |
| 6 watt 1K15 ohm | 12 | £1.00 | BD813 |
| 6 watt 1K5 ohm | 12 | £1.00 | BD814 |
| 6 watt 3K9 ohm | 12 | £1.00 | BD815 |
| 6 watt 15K ohm | 12 | £1.00 | BD816 |
| 6 watt 19K ohm | 12 | £1.00 | BD817 |
| 6 watt 20K ohm | 12 | £1.00 | BD818 |
| RESISTORS WIRE WOUND 7 WATT |  |  |  |
| 7 watt 6.8 ohm | 10 | £1.00 | BD880 |
| 7 watt 80 hm | 10 | £1.00 | BD881 |
| 7 watt 12 ohm | 10 | £1.00 | BD882 |
| RESISTORS WIRE WOUND 9 WATT |  |  |  |
| 9 watt 12 ohm | 8 | £1.00 | BD819 |
| 9 watt 820 ohm | 8 | £1.00 | BD821 |
| 9 watt 1K ohm | 8 | £1.00 | BD822 |
| 9 watt 68k ohm | 8 | £1.00 | BD823 |
| RESISTORS WIRE WOUND 11 WATT |  |  |  |
| 11 watt 6.0 ohm | 6 | £1.00 | BD877 |
| 11 watt 6.8 ohm | 6 | £1.00 | BD876 |
| 11 watt 11 ohm | 6 | £1.00 | BD878 |
| RESISTORS WIRE WOUND 12 WATT |  |  |  |
| 12 watt 2.2 ohm | 5 | £1.00 | BD824 |
| 12 watt 10 ohm | 5 | £1.00 | BD825 |
| 12 watt 12 ohm | 5 | $\underline{1.00}$ | BD826 |
| RIBBON CABLE |  |  |  |
| 10 way $100{ }^{\text {reel grey } 71.127 \mathrm{~mm}}$ | 1 | £7.00 | 7P24 |
| 16 way 100' reel | 1 | £12.00 | 12P32 |
| 20 way 100' reel | 1 | £14.00 | 14P4 |
| 26 way 100' reel | 1 | $\varepsilon 18.00$ | 18 P 8 |
| 34 way 100' reel | 1 | £24.00 | 24P6 |
| 40 way 100' reel | 1 | £28.00 | 28P2 |
| SEMI CONDUCTORS |  |  |  |
| LM317T Variable voltage reg 1.5A. | 2 | £1.00 | BD905 |
| Bargain pack of 20 different ICs. | 1 | £1.00 | BD906 |
| TIP3055 power transistor. | 1 | £1.00 | BD655 |
| 100 watt mostet pair 2S. 99 and 2SK343. | 1 | £4.00 | 4P51 |
| Power mosfet 2SJ77. | 1 | £2.00 | 2P285 |
| J111 JFET. | 10. | £1.00 | BD864 |
| Large power transistors. (unknown spec). | 4 | £1.00 | BD210 |
| 3A SCR 50 v stud mounting. | 4 | £1.00 | BD575 |
| SERVICE AIDS |  |  |  |


| DESCRIPTION | $\begin{array}{\|c\|} \hline \text { PACK } \\ \text { OUANTIIY } \end{array}$ | ${ }_{\text {PACK }}$ | OROER NUMBER |
| :---: | :---: | :---: | :---: |
| Switch cleaner $\mathbf{2 2 6 g}$ aerosol with applicator. | 1 | £2.00 | 2P321 |
| Aero clene precision cleaner. Aerosol. | 1 | £2.00 | 2 P 322 |
| Circuit freezer for fault finding. 226g aerosol. | 1 | £2.00 | 2 P 323 |
| Foam cleanser (powerful). 370 g aerosol. | 1 | £2.00 | 2 P 324 |
| Silicone grease for waterprooting. $\mathbf{2 2 6 g}$ aerosol. | 1 | $£ 2.00$ | 2 P 325 |
| Anti static spray mist. 150 g aerosol | 1 | £2.00 | 2 P 326 |
| Plastic seal. 145 g aerosol. | 1 | £2.00 | 2 P 327 |
| Aero duster. 200 g aerosol. | 1 | £2.00 | 2 P 328 |
| Quick fix mains connector. | 1 | £7.00 | 7 P 18 |
| Stylus cleaning brush. | 2 | £1.00 | BD730 |
| Carbon fibre record cleaning brush 1,000,000 fibres. | 1 | £1.00 | BD707 |
| 500 ml of precision gear oil with applicator tube. | 1 | £2.00 | 2 P 269 |
| Disc drive head cleaner for $51 / 4^{\prime \prime}$ double sided. | 1 | £2.00 | 2P250 |
| Soldering Iron stand complete with sponge. | 1 | £3.00 | 3 P 66 |
| Etch resist pen for making PCBs. | 1 | £1.00 | BD699 |
| 30 watt mains soldering iron. Wooden handle. | 1 | $£ 3.00$ | 3 P 65 |
| Antex 15 watt iron with 1 mm bit. 250 v . | 1 | £10.00 | 10P86 |
| 1/2kg solder $60 / 40$ multicore resin solder 18swg. | 1 | £6.00 | 6P9 |
| Instrument case with handle $5^{\prime \prime} \times 4$ "x2". | 1 | £1.00 | BD742 |
| Portasol gas soldering Iron (uses lighter gas) 10-60w. | 1 | £18.00 | 18P4 |
| Loctite metal adhesive. | 1 | £2.00 | 2 P 215 |
| 10kv extension for Multi tester. | 1 | £2.00 | 2P211 |
| Minl pocket screw driver. | 4 | £1.00 | BD436 |
| $1 \times 4.2 \mathrm{~mm}$ drill bit, 3 allen keys and 15 mm tap cutter. SOLAR CELLS | 1 | £1.00 | BD863 |
| Solar battery charger. Takes 4 AA cells. | 1 | $£ 6.00$ | 6P3 |
| 700 mA .45 v solar cell. $95 \times 65 \times 7.5 \mathrm{~mm}$. | 1 | £3.00 | 3 P 42 |
| 400 mA .45 v solar cell. $75 \times 45 \times 7.5 \mathrm{~mm}$. | 1 | £2.00 | 2P199 |
| 100mA . 45 solar cell. $45 \times 26 \times 7.5 \mathrm{~mm}$. | 1 | £1.00 | BD631 |
| SOLENOIDS |  |  |  |
| Mains solenoid, spring loaded $11 / 2^{\prime \prime}$ square $1 / 4^{\text {" }}$ travel. | 1 | £1.00 | BD701 |
| Mains solenoid. $38 \times 25 \times 32 \mathrm{~mm} 25 \mathrm{~mm}$ travel. | 1 | £1.00 | BD300 |
| 115 v solenoid. $50 \times 35 \times 38 \mathrm{~mm} 20 \mathrm{~mm}$ travel | 1 | £1.00 | BD199 |
| Mains solenoid (push or pull). $35 \times 25 \times 32 \mathrm{~mm} 10 \mathrm{~mm}$ trave | 1 | £1.00 | BD387 |
| Flat solenoids. | 2 | £1.00 | BD59 |
| 12 v solenoid with plunger. $30 \times 12 \times 12 \mathrm{~mm} 20 \mathrm{~mm}$ travel. | 1 | £1.00 | BD232 |
| $70 v$ DC solenoid $21 / 2^{\prime \prime} \times 1^{\prime \prime}$. Powerful. | 1 | £2.00 | 2 P 271 |
| 12v DC solenoid $13 / 4^{\prime \prime}$ square. Powerful. | 1 | £2.00 | 2P272 |
| SPECIAL LIGHTING EFFECTS |  |  |  |
| Motor driven switch 6 or more 10A c/o contacts. | 1 | £2.00 | 2 P 19 |
| Sound to light kit. 750 watts per channel(3). Cased. | 1 | £20.00 | 20 P 35 |
| Mains motor driver flasher 1000 watt. | 1 | £2.00 | 2 P 25 |
| Strobe light, cased, 240v AC adjustable speed. STRIP BOARD | 1 | £25.00 | 25P23 |


| DESCRIPTION | $\begin{gathered} \text { PACK } \\ \text { QUANTITY } \end{gathered}$ | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| $5^{\prime \prime} \times 4^{\prime \prime} .1$ spacing copper clad strip board. | 1 | £1.00 | BD736 |
| 17 " $\times 4^{\text {" }} 1$ spacing copper clad strip board. STRIPPERS | 1 | £4.00 | 4P62 |
| Component board full of diodes etc. | 1 | £1.00 | BD120 |
| Component board full of ICs transistors etc. | 1 | $£ 2.00$ | 2P282 |
| Magnetic read write unit. | 1 | £1.00 | BD381 |
| Ex GPO boards containing many assorted components. SUPPRESSORS | 4 | £1.00 | BD620 |
| Mains filter (inductance and capacitance) with leads 13A. | 1 | $£ 1.00$ | BD248 |
| Auto noise eliminator inline for car use. | 2 | £1.00 | BD751 |
| Noise suppressor 2A inline mains. Chassis mount. | 1 | £1.00 | BD570 |
| Noise suppressor 3A $1^{\prime \prime} \times 3 / 8^{\prime \prime}$. Chassis mount. | 1 | £1.00 | BD698 |
| SURVEILLANCE PRODUCTS |  |  |  |
| FM bug bullt Inside a 13A mains adapter. The adapter |  |  |  |
| still functions but transmits even the slightest wisper. | 1 | £26.00 | 26P2 |
| FM bug kit with embedded coil. | 1 | $£ 5.00$ | 5P158 |
| Built and tested superior minature FM bug. | 1 | £14.00 | 14P3 |
| Built and tested phone bug fits inside phone and is powered by the phone. | 1 | $£ 20.00$ | 20P28 |
| As above but built into a secondary BT socket. | 1 | £23.00 | 23 P 10 |
| As above but built into a master socket. | 1 | £24.00 | 24P5 |
| Handheld built and tested bug detector. | 1 | £50.00 | 50P4 |
| Airband receiver. | 1 | £25.00 | 25P17 |
| FM bug built into a pen | 1 | £75.00 | 75P1 |
| FM bug buit into a calculator. | 1 | £75.00 | 75P2 |
| SWITCHES ROTARY |  |  |  |
| 4 pole 3 way. | 3 | £1.00 | BD870 |
| 3 pole 4 way. | 3 | £1.00 | BD871 |
| 2 pole 6 way. | 3 | £1.00 | BD872 |
| 1 pole 12 way. | 3 | £1.00 | BD873 |
| Rotary mains on/off switch. | 2 | £1.00 | BD456 |
| 12 pole 2 way. | 2 | £1.00 | BD874 |
| Minature edgewise 4 pole 1 way with knob. | 1 | £1.00 | BD496 |
| Ceramic wave change switch 12 pole 3 way. $1 / 4^{\prime \prime}$ shaft. SWITCHES PUSH | 1 | £1.00 | BD303 |
| Normally on heavy duty metal switches. | 4 | £1.00 | BD176 |
| Oblong push switches 5A 220v AC. | 2 | £1.00 | BD263 |
| Key board style push switch with knob. | 8 | £1.00 | BD201 |
| Sub minature push switch DPDT. | 1 | £1.00 | BD650 |
| Panel mount push switch. 5A 250v. | 4 | £1.00 | BD670 |
| 12 banks of switches (100 switches). | 1 | £1.00 | BD481 |
| Push mains switches. | 4 | £1.00 | BD494 |
| Pushon push off mains table lamp switch. | 4 | £1.00 | BD121 |

## DESCRIPTION

Peavy duty illuminated swich 2 pole CIO at 16A.

## SWITCHES SLIDE

Minature slide switch with chrome dolly. DPDT. Sub minature slide switch 2 pole c/o.
Slide switch single pole C/O chassis mount. Mini slide switch DPDT 5A.
SWITCHES TOGGLE
2 pole C/O. Toggle springs back to c/off position. $15 \times 15 \times 45 \mathrm{~mm}$ non latching.
Standard size toggle switch.
4 pole cetre off c/o toggle switch. 10A 250v.
Sub minature toggle switch $8 \times 4 \times 7 \mathrm{~mm}$ SPST.
SWICHES ROCKER
Rocker switch DPST.
Rocker switch panel mount single pole 10A. White.
Double pole rocker with built in 3A trip. 240v.
Rocker switch 2 pole C/O 10A.
Rocker switch SPST centre off. 10A 250v.
Rocker switch SPST. 13A.
Spring loaded 10A rocker switch. (car window etc).

## SWITCHES VARIOUS

30A panel mount toggle switch 250 v .
PCB mount 16 way BCD output 6 mm square rotary. Push swich with protective shoulder. For motor start etc. Motorized stud switch (ex equipment).
Key operated switch.
Key operated switch good quality Yale.
Glass reed switch.
Mercury switch. Mains rated.
Humidity swich.
Low voltage dolls house type switch.
Thumb wheel switch minature BCD.
Thumb wheel switch standard size normal contacts.
Double pole on off leaf switch.
TAGS AND CONNECTORS
Push on $1 / 4^{\prime \prime}$ tag connectors.
Push on 1/4" tag connectors (right angled).
Soldercon terminals (make your own IC sockets).
Push on tags for $1 / 8^{\prime \prime}$ spindle.
Tags with hole to fit 6BA bolt.
TAPE BITS
Double micro cassette deck. Complete with motors. Assorted tape heads.
Stereo tape heads (cassette type).

| DESCRIPTION | PACK QUANTITY | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| Tape splicing block with cutter 45 and 90 deg 2 tape size | 1 | $£ 1.00$ | BD723 |
| TELEPHONE BITS |  |  |  |
| 20 metres of 4 core white phone cable. | 1 | £2.00 | 2P98 |
| 100 metre reel of white 4 core cable. | 1 | $£ 8.00$ | 8P15 |
| 500 pack of cable clips for above cable. | 1 | £2.00 | 2P99 |
| 10 digit switch pad for telephone etc. | 1 | £1.00 | BD200 |
| Pushbutton 10 memory phones (customer returns). | 1 | £6.00 | 6P16 |
| Frequency changer unit. Reduces250v 50hz to 50v 25 hz . | 1 | £10.00 | 10P37 |
| 3 metre phone leads new style plug. Black. | 2 | £1.00 | BD639 |
| 3 mtre phone leads new style plug. White. | 1 | £1.00 | BD705 |
| Device to enable any phone to be converted to new plug | 1 | £2.00 | 2P249 |
| Twin telephone bells AC. | 1 | £1.00 | BD600 |
| 5 pole GPO plug and wall mount socket. | 1 | £1.00 | BD534 |
| 5 metre new style telephone extension lead. | 1 | £3.00 | 3P70 |
| TERMINALS |  |  |  |
| Screw down terminal posts. Also take 4mm banana plug. | 6 | $£ 1.00$ | BD264 |
| Minature panel mount terminals. Screw tops. | 4 | £1.00 | BD755 |
| THERMOSTATS |  |  |  |
| Danfloss cased wall mount heating thermostat. | 1 | £4.00 | 4P4 |
| Wall mounting low voltage thermostat. | 1 | £1.00 | BD115 |
| Capillary type fridge thermostat. | 1 | £1.00 | BD96 |
| Panoset. Keeps saucepans at pre set temperature. | 1 | $£ 1.00$ | BD252 |
| Oven thermostat with tep calibrated knob. | 1 | £2.00 | 2P158 |
| Appliance thermostat. Spindle type adjustment. | 2 | £1.00 | BD582 |
| Washing machine thermostat. | 1 | £1.00 | BD458 |
| 3 level water thermostats. | 4 | £1.00 | BD537 |
| THERMISTORS |  |  |  |
| 5mm dia NTC 40R@100 deg, 300@25R. | 4 | £1.00 | BD978 |
| 5mm dia NTC 80R@100 deg, 1K@ 25R. | 4 | £1.00 | BD979 |
| 5 mm dia NTC 380R@100 deg, 5K@ 25R. | 4 | £1.00 | BD980 |
| 5 mm dia NTC 1K8R@100 deg, 30K @ 25R. | 4 | £1.00 | BD981 |
| 5mm dia NTC 5K1@100 deg, 100K@ 25R. | 4 | £1.00 | BD982 |
| TIMERS AND TIME SWITCHES |  |  |  |
| 25A electrical programmer (ex equipment). | 1 | £3.00 | 3P106 |
| Microwave control panel with digltal clock and relay o/p. | 1 | £6.00 | 6P18 |
| Plug in 24 hour time switch. | 1 | £10.00 | 10P54 |
| Mains motor driven. Switches on or off for 20 secs. | 1 | £2.00 | 2P22 |
| 10 min clockwork time switch 15A 230v. | 1 | £1.00 | BD579 |
| 90 min time switch engraved in mins 15A 230 v . | 1 | £2.00 | 2 P 90 |
| 2 hour time switch. Clockwork 15A. | 1 | £2.00 | 2P89 |
| 12 hour time switch. Clockwork 15A 240v. | 1 | £2.00 | 2P24 |
| Smiths time and set switch for heating etc. 15A. | 1 | £2.00 | 2P9 |
| Omron Industrial timer. | 1 | £5.00 | 5P55 |
| 100A time switch. 1 on 1 off per 24 hours. | 1 | £10.00 | 10P14 |


| DESCRIPTION | PACK OUANTITY | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| Plug in Venner solar dial including case with window. | 1 | £7.00 | 7P17 |
| Adaptor kit for above to give 12 on/offs. | 1 | £3.00 | 3P105 |
| 24 hour time switch. (ex equipment). | 1 | $£ 1.00$ | BD45 |
| 12 hour timer. | 1 | £1.00 | BD478 |
| Mains driven electric clock. (uncased). | 1 | £1.00 | BD211 |
| TOOLS ETC |  |  |  |
| Extra thin screw driver. | 4 | £1.00 | BD129 |
| $5^{\prime \prime}$ electricians pliers. | 1 | £2.00 | 2P253 |
| Multimeter 16 ranges .2K OPV with leads. | 1 | £7.00 | 7P10 |
| Multimeter 20K OPV 10A AC ample ranges. | 1 | £15.00 | 15P30 |
| 4BA spanner. | 10 | £1.00 | BD142 |
| Screw drivers. | 10 | $£ 1.00$ | BD322 |
| Top pocket screw drivers. | 2 | £1.00 | BD436 |
| $7{ }^{\text {² }}$ electricians pliers. | 1 | £3.00 | 3P25 |
| $6^{\prime \prime}$ diagonal side cutters. | 1 | £2.00 | 2P161 |
| Set of 10 minature flat blade screwdrivers in case. | 1 | £2.00 | 2 P311 |
| Set of 10 minature crosspoint screwdrivers in case. | 1 | £2.00 | 2P312 |
| Set of 5 minature nut spinners in case (metric). | 1 | $£ 2.00$ | 2 P 313 |
| 7" Mole grip style wrench. | 1 | £3.00 | 3P114 |
| Minature electronics vice, table mounting. Metal. | 1 | £5.00 | 5P180 |
| Minature PCB drill 9-16v DC 1.5A. With case. | 1 | £15.00 | 15P41 |
| CHROME VANADIUM SCREWDRIVERS |  |  |  |
| $75 \times 3 \mathrm{~mm}$ flat blade. | 1 | £1.00 | CD15 |
| $100 \times 3 \mathrm{~mm}$ flat blade. | 1 | £1.00 | CD16 |
| $150 \times 5 \mathrm{~mm}$ flat blade. | 1 | $£ 1.00$ | CD17 |
| $200 \times 5 \mathrm{~mm}$ flat blade. | 1 | £1.00 | CD18 |
| 75 mm 0 size pozidrive. | 1 | £1.00 | CD19 |
| 75 mm 1 size pozidrive. | 1 | £1.00 | CD20 |
| 100 mm 2 size pozidrive. | 1 | £1.00 | CD21 |
| 150 mm 3 size pozidrive. | 1 | £2.00 | 2P339 |
| TRANSFORMERS COUPLING |  |  |  |
| Minature driver transformer 20K to 1 K (centre tapped). TRANSFORMERS TOROIDAL | 1 | $£ 1.00$ | BD653 |
| 24v 5A plastic encapsulated 4" diameter. | 1 | £5.00 | 5P34 |
| 6 v 1 A upright mounting. | 2 | $£ 1.00$ | BD9 |
| $8 \mathrm{v} 1 / 2 \mathrm{~A}$ double upright mounting. | 1 | £1.00 | BD212 |
| 9v 1/2A upright mounting. | 1 | £1.00 | BD266 |
| 12v 1/2A upright mounting. | 1 | £1.00 | BD10 |
| 15 V 2 A upright mounting. | 1 | £3.00 | 3 P 88 |
| 15 v 4 A upright mounting. | 1 | $£ 4.00$ | 4P56 |
| 30 v 1 A upright mounting. | 1 | £2.00 | 2P270 |
| 18v 500VA transformer. | 1 | £20.00 | 20P9 |
| 24 V 5A transformer in waterproof box for outside use. | 1 | £8.00 | 8P17 |


| DESCRIPTION | PACK | PACK PRICE | OROER |
| :---: | :---: | :---: | :---: |
| 9-0-9v. 4 A PCB mounting. | 1 | £1.00 | BD661 |
| 12-0-12v 1/2 A PCB mounting. | 1 | £1.00 | BD636 |
| $20-0-20 \mathrm{v} 2.5 \mathrm{~A}$ upright mounting. | 1 | £4.00 | 4P24 |
| 24-0-24v plus 12-0-12v 150 watt. | 1 | £8.00 | 8 P 11 |
| 50 v 2 A plus 6.3 v upright mounting. | 1 | £3.00 | 3 P 10 |
| 90 v plus 90 v 220 mA (large transformer). | 1 | $£ 5.00$ | 5P128 |
| 100 v plus 100 v 1 A upright mounting. | 1 | £5.00 | 5P86 |
| $250-0-250 \mathrm{v} 60 \mathrm{~mA}$ plus 6.3 v at 5A. | 1 | £4.00 | 4P41 |
| 20 V 1.5 A upright mounting. | 1 | £2.00 | 2P214 |
| 440 v to 240 v stepdowm transformer. | 1 | £5.00 | 5P67 |
| EHT transformer 8kv 3mA. | 1 | £10.00 | 10P56 |
| 10 v plus 17v 1A transformer. | 1 | £1.00 | BD492 |
| CHASSIS MOUNT TRANSFORMERS |  |  |  |
| 6-0-6 2A 59x50x54mm 24 VA . | 1 | $£ 5.00$ | 5P184 |
| 9-0-9 2A 59x59x50mm 36 VA . | 1 | $£ 6.00$ | 6P48 |
| 12-0-12 2A 68x57x55mm 48VA. | 1 | $£ 7.00$ | 7P27 |
| 20-0-20 2A 78x65x66mm 80VA. | 1 | £10.00 | 10P98 |
| $30-0-30 \mathrm{v} 2 \mathrm{~A} 78 \times 65 \times 66 \mathrm{~mm} 120 \mathrm{VA}$. | 1 | £12.00 | 12P36 |
| $6-0-6 \mathrm{~V} 4 \mathrm{~A} 68 \times 57 \times 54 \mathrm{~mm} 48 \mathrm{VA}$. | 1 | £7.00 | 7P28 |
| $6-0-6 \mathrm{~V} 8 \mathrm{~A} 78 \times 65 \times 72 \mathrm{~mm} 96 \mathrm{VA}$. | 1 | £12.00 | 12 P 37 |
| $9-0-9 \mathrm{v} 4 \mathrm{~A} 68 \times 57 \times 66 \mathrm{~mm} 72 \mathrm{VA}$. | 1 | £10.00 | 10P99 |
| 12-0-12v 4A $78 \times 65 \times 70 \mathrm{~mm} 96 \mathrm{VA}$. | 1 | £12.00 | 12P38 |
| 12-0-12v 8A 96x83x82mm 192VA. | 1 | £22.00 | 22P4 |
| $0-153 \mathrm{~A} 68 \times 58 \times 62 \mathrm{~mm} 45 \mathrm{VA}$. | 1 | £6.00 | 6P49 |
| AUTO TRANSFORMERS |  |  |  |
| Electronic auto transformer 1 KW resistive loads only. | 1 | $£ 5.00$ | 5P157 |
| 100 watt auto transformer. | 1 | £2.00 | 2P6 |
| TRANSISTORS |  |  |  |
| AC128 | 3 | £1.00 | BD909 |
| AC127 | 3 | £1.00 | BD910 |
| BC108 | 6 | £1.00 | BD911 |
| BC109 | 6 | £1.00 | BD912 |
| BC140 | 4 | £1.00 | BD913 |
| BC141 | 4 | £1.00 | BD914 |
| BC142 | 3 | £1.00 | BD915 |
| BC143 | 3 | £1.00 | BD916 |
| BC149 | 6 | £1.00 | BD917 |
| BC158 | 6 | £1.00 | BD918 |
| BC160 | 3 | £1.00 | BD920 |
| BC182L | 12 | £1.00 | BD921 |
| BC183L | 10 | £1.00 | BD922 |
| BC184L | 10 | £1.00 | BD923 |
| BC212L | 10 | £1.00 | BD924 |
| BC213L | 10 | £1.00 | BD925 |


| DESCRIPTION | PACK QUANTITY | $\begin{aligned} & \text { PACK } \\ & \text { PRICE } \end{aligned}$ | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| BC2378 | 10 | £.00 | BD926 |
| BC238B | 10 | £1.00 | BD927 |
| BC337 | 10 | £1.00 | BD928 |
| BC338 | 10 | £1.00 | BD929 |
| BCY70 | 4 | £1.00 | BD930 |
| BCY71 | 4 | £1.00 | BD931 |
| BFY50 | 3 | £1.00 | BD932 |
| BFY51 | 3 | £1.00 | BD933 |
| BU208 | 1 | £2.00 | 2P298 |
| MJ2955 | 2 | £2.00 | 2 P 299 |
| MJE3055 | 2 | £2.00 | 2P300 |
| TIP29A | 6 | £2.00 | 2P301 |
| TIP30A | 6 | £2.00 | 2 P302 |
| TIP31C | 5 | £2.00 | 2 P 303 |
| TIP32C | 5 | £2.00 | 2P304 |
| ZTX300 | 5 | £1.00 | BD934 |
| ZTX500 | 5 | £1.00 | BD935 |
| 2N2905 | 4 | £1.00 | BD936 |
| 2N3055 | 3 | £2.00 | 2P305 |
| TRIACS AND THYRISTORS |  |  |  |
| TIC206D 3A 400v triac. | 4 | $£ 2.00$ | 2 P 314 |
| TIC226D 8A 400v triac. | 3 | £2.00 | 2 P 315 |
| TIC246D 16A 400v triac. | 2 | £2.00 | 2 P 316 |
| TIC106D 4A 400v thyristor. | 5 | £2.00 | 2 P 317 |
| TIC116D 8A 400v thyristor. | 3 | £2.00 | 2 P 318 |
| TIC126D 12A 400V thyristor. | 2 | $£ 2.00$ | 2P319 |
| TV BITS |  |  |  |
| Flyback EHT unit ITT ref 17ACC79. | 1 | £2.00 | 2 P 111 |
| Varicap push button tuner. | 2 | £1.00 | BD155 |
| 8kV GEC line output transformers. | 1 | £2.00 | 2 P 262 |
| Yoke assemblies for TV tubes (contain a lot of wire). | 1 | £1.00 | BD747 |
| Linearity coils ref BKM10-7. | 2 | £1.00 | BD462 |
| TV sound receiver box 240 V 7 channel. | 1 | £12.00 | 12 P 22 |
| 75 ohm low loss co-ax cable.(10 metres). | 1 | £2.00 | 2 P 236 |
| 75 ohm low loss co-ax cable ( 100 metres). ULTRASONICS | 1 | £15.00 | 15P31 |
| Ulirasonic intruder detector. | 1 | £20.00 | 20P11 |
| UNISELECTORS |  |  |  |
| Minature 50 v unselector with circuit ideas. | 1 | £1.00 | BD56 |
| VALVES AIR AND FLUID |  |  |  |
| 24v DC air or water valve. Threaded couplings. | 1 | $£ 10.00$ | 10P73 |
| Air or gas shut off valve. Clock operated. | 1 | £1.00 | BD152 |
| Air or gas shut off valve. Temp operated. | 1 | £1.00 | BD153 |
| 110 v mains operated air valve. | 1 | £1.00 | BD202 |


| DESCRIPTION | $\begin{aligned} & \text { PACK } \\ & \text { OUANTITY } \end{aligned}$ | PACK PRICE | ORDER NUMBER |
| :---: | :---: | :---: | :---: |
| 230 v mains operated air valve. | 1 | £2.00 | 2P34 |
| 230 v mains water valve. | 1 | £1.00 | BD370 |
| High pressure mains gas or water valve 1/2" thread. | 1 | £5.00 | 5P171 |
| High pressure mains gas or water valve 1 " thread. VIDEO TAPES | 1 | £5.00 | 5P172 |
| Blank 3 hour top quality video tape. (VHS). | 1 | £3.00 | 3P63 |
| Blank 3 hour top quality video tapes. (VHS). | 5 | £11.00 | 11P3 |
| VOLTAGE REGULATORS |  |  |  |
| 5 V 1A TO220 voltage regulator. 7805 | 3 | $£ 1.00$ | BD962 |
| 12v 1A TO220 voltage regulator. 7812 | 3 | £1.00 | BD963 |
| -5v 1A TO220 voltage regulator. 7905 | 3 | £1.00 | BD964 |
| -12v 1A TO220 voltage regulator. 7912 | 3 | £1.00 | BD965 |
| 5 v 2 A TO220 voltage regulator. 78 S 05 | 2 | £1.00 | BD966 |
| -12v 2A TO220 voltage regulator 78 S 12 | 2 | £1.00 | BD967 |
| WHEELS |  |  |  |
| $13^{\prime \prime}$ C5 spoked wheel with tyre \& tube. Cycle bearing | 1 | £6.00 | 6P10 |
| $16^{n}$ C5 spoked wheelwith tyre \& tube. $1^{\prime \prime}$ centre hole. WIRE AND CABLE | 1 | £6,00 | 6P11 |
| 2 core pvc covered cable 5A. (18 metres). | 1 | £2.00 | 2P218 |
| 2 core pvc covered cable 8A. (15 metres). | 1 | £2.00 | 2P219 |
| 3 core pvc covered cable 5A. (15 metres). | 1 | £2.00 | 2P189 |
| 3 core pvc covered cable 8A. (14 metres). | 1 | £2.00 | 2P220 |
| 3 core pvc covered cable 10A. (12 metres). | 1 | $£ 2.00$ | 2 P 221 |
| 3 core pvc covered cable 13A. (10 metres). | 1 | £2.00 | 2P190 |
| IBM printer lead (D25 to Centronics plug). Parallel 2 m . | 1 | £5.00 | 5P186 |
| IBM printer lead (d25 to Centronics plug). Parallel 3m. | 1 | £6.00 | 6P50 |
| RS232 data cable D25 male to D25 male 2Metres long. | 1 | £5.00 | 5P187 |
| RS232 data cable D25 female to D25 male. 2M long. | 1 | £5.00 | 5P188 |
| Centronics cable. Plug to plug 2 metres long. | 1 | £6.00 | 6P51 |
| RS232 D25 gender changers. Male to male. | 1 | £3.00 | 3P124 |
| RS232 D25 gender changers. Female to female. | 1 | £3.00 | 3P125 |
| D9 gender changer male to male. | 1 | £3.00 | 3P126 |
| D9 gender changer female to female. | 1 | £3.00 | 3P127 |
| Centronics gender changer male to male. | 1 | £6.00 | 6P52 |
| Centronics gender changer female to female. | 1 | £6.00 | 6P53 |
| D9 male to D25 female adaptor. (IBM compatable). | 1 | £3.00 | 3P128 |
| D9 female to D25 male adaptor. (IBM compatable). | 1 | £3.00 | 3P129 |
| IEC lead fitted with IEC socket. 2 metres. | 1 | £2.00 | 2 P347 |
| IEC lead fitted with IEC socket and 13A plug. 2M. | 1 | £3.00 | 3P130 |
| Equipment wire PTFE coated and silver plated. (5m). | 1 | £1.00 | BD564 |
| Single core screened cable $7 \times .02$ conductors. (10m). | 1 | £1.00 | BD668 |
| 4 core cable $7 \times .2 \mathrm{~mm}$ grey. (100 metres). | 1 | £8.00 | 8P19 |
| Garden tool extension cable 2 core. (20m). | 1 | £2.00 | 2P20 |
| 2 core screened cable. (10 metres). | 1 | £1.00 | BD122 |



## SAB9 RADIO RECEIVER

First class radio that receives MW/LW/FM/AIR/MARINE bands (108-176)mhz.
Fine tuning control on the FM band. Complete with earphone. $£ 25.00$ ref 25P17


RADIO RECEIVER MBR 7
A fully featured multiband that receives MW 5351630khz,LW 150-300kh FM 87.5-108,AIR 108136, MARINE 136-176
SW1 7-23MHZ, SW2 3-7.2MHZ AFC, direction finder, fine tuning mains or battery.
£64.00 ref 64P3
LOW COST RECEIVER 8342
Covers AIR 108-145MHZ, PB 145176MHZ, WB 162.5MHZ, TV 5487MHZ, FM 88-108MHZ, CB CHANNELS 1-80.
Receiver measures $100 \times 200 \mathrm{x}$ 50 mm . Supplied complete with squelch, control and strap.
Battery operated.
£19.00 ref 19P14
Mains adaptor $£ 5.00$ ref 5P191


12v Solar cell 200 mA output ideal for trickle charging etc. \&15.00 each ref 15P42


SOLAR CELLS
600Ma .45V £3.00 REF 3P42 400Ma . 45 V £2.00 REF 2P199 100 Ma .45 V £1.00 REF BD631 new 1 AMP . 45 circular solder type only $\mathbf{1 5 . 0 0}$ ref 5P192


SOLAR KIT
Comprises of solar cell, Special DC motor, plastic fan and turntables, various cutout shapes and a 20 page book about solar energy.
Price is $£ 8.00$ ref $\mathbf{8 P 5 1}$


## FIBRE OPTIC CABLE

Stranded optical fibres sheathed in black PVC
1 metre length $£ 2.00$ ref $\mathbf{2 P 3 5 1}$ 5 metre length $£ 7.00$ ref 7 P 29

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FM CORDLESS MICROPHONEThis device is a small handheldmicrophone that transmits to anordinary FM radio. It is batteryoperated (PP3) and has twotransmit power levels. Tuneablefrom $90-105 \mathrm{MHZ}$ max imumrange is $500^{\prime} £ 15.00$ ref 15 P 42
MINATURE RADIO TRANSCEIVERS

A pair of walkie talkies with a range of up to 2 Kilometres. Units measure $22 \times 52 \times 155 \mathrm{~mm}$ and come complete with cases, aerials and earpieces. Price for the pair is $£ 30.00$ ref 30P12.


CAR STEREO AND FM RADIO


Low cost car stereo system giving 5 watts per channel. Signal to noise ratio < 45db wow and flutter less than .35\%. neg earth. Retail price $£ 49.95$, ours.. $£ 25.00$ ref 25P21
|WHISPER 2000 LISTENING AID You may have seen these aids advertised in the national press at $£ 8.95$. If you haven't its a device that enables you to hear sounds that would otherwise be inaudible.
Our Price is $£ 5.00$ ref 5P179
10 BAND COMMUNICATIONS RECEIVER Cmprehensive receiver covering 7 shortwave bands FM, AM and LW. Tuning eye DXAocal switch mains or battery. Complete with shoulder strap \& mains lead. £34.00 ref 34P1

## VIDEO SENDER UNIT

These units are designed to transmit audioand video signals from either a Video, a camera or a computer to a standard TV set. Max
 range is 100 ft . 12 v operation, Price $£ 15.00$ ref 15P39 12v psu $£ 5.00$ ref 5P178.


## PASSIVE INFRA RED MOTION SENSOR

Built in day light sensor. Adjustable lights on time (8 sec-15mins) $50^{\prime}$ range with a 90 degree coverage. Manual over ride facility.
Complete with wall brackets bulb holders etc Brand new and guaranteed OUR PRICE $£ 25.00$ ref 25P24

# Bull Electrical 250 Portland road Hove <br> Sussex BN3 5QT Phone 0273-203500 Fax 027323077 


[^0]:    Carbon Film resistors $1 / 2 \mathrm{~W} 5 \%$ E24 series 0.51 R to 10 MO
    100 off per value $-75 p$, even hundreds per value totalling 1000 Metal Film resistors \%WW 1OR to 1 MO 5\% E12 series - 2p. 1\% E24 series Mixed metal/carbon film resistors \%W E24 series 1 RO to 10MO. Linear Carbon pre-sets 100 mW and $1 / 4 \mathrm{~W} 100 \mathrm{R}$ to 4 M 7 E 6 series Miniature polyster 100 mW and $/ 4 W \mathrm{~W} 100 \mathrm{R}$ to 4 M 7 E series 015, .022, .033, .047, .068-4p, 0.1-5p.0.12, 0.15, 0.22-6p.0. Mylar (polyester) capacitors 100 V working $\mathbf{E 1 2}$ series vertical mounting 1000 p to 8200 p-3p. 01 to $068-4 \mathrm{p} .0 .1-5 \mathrm{p} .0 .12 .0 .15,0.22-6 \mathrm{p} .0 .47 / 50 \mathrm{~V}-8 \mathrm{p}$ Submin ceramic plate capacitors 100 V wkg vetical mountings. E12 series $2 \% 1.8 \mathrm{pf}$ tp $47 \mathrm{pf}-3 \mathrm{p} .2 \% 56$ pf to 330pt - 4p. 10\% 390p-4700p.
    Disc/plate ceramics 50 V E12 series 1 PO to 1000 P , E6 Series 1500 P to 47000 P Polystyrene capacitors 63 V working E12 series long axial wires 10 pf to $820 \mathrm{pf}-3 \mathrm{p} .1000 \mathrm{pf}$ to $10.000 \mathrm{pf}-4 \mathrm{p} .12 .000 \mathrm{pf}$.
    4mos 4001-20p. 4011 Timer
    ALUMINIUM ELECTROLYTICS (Mfds/Volts)
    1/50, 2.2/50, 4.7/50, 10/25, 10/50.
    $22 / 16,22 / 25,22 / 50,47 / 16,47 / 25,47 / 50$
    100/16, 100/25 7p; 100/50 12p; $100 / 100 \ldots .$.
    $220 / 168 p ; 220 / 25,220 / 5010 p ; 470 / 16,470 /$
    $1000 / 2525$ p; 1000/35, 2200/25 35p; 4700/25
    Submin, tantalum bead electrolytics (Mfds/Volts)
    $0.1 / 35,0.22 / 35,0.47 / 35,1.0 / 35,3.3 / 16,4.7 / 16$
    $2.2 / 35,4.7 / 25,4.7 / 35,6.8 / 1615 \mathrm{p} ; 10 / 16,22 / 6$
    $33 / 10,47 / 6,22 / 1630 p ; 47 / 1035 p ; 47 / 1660 \mathrm{p} ; 47 / 35$
    VOLTAGEREGULATORS
    $1 \mathrm{~A}+$ or $-5 \mathrm{~V}, 8 \mathrm{~V}, 12 \mathrm{~V}, 15 \mathrm{~V}, 18 \mathrm{~V} \& 24 \mathrm{~V}-55 \mathrm{P} .100 \mathrm{~mA} .5,8,12,15, \mathrm{~V}+$ DIODES (piv/amps
    75/25mA N41482p. 800/1A 1N4006 41/2p. 400/3A 1N5404 14p. 115/15mA OA 100/1A 1 N4002 31/2p. 1000/1A 1 N4007 5p. 60/1.5A S1M1 5 p. 100/1A bridge Zener diodes E24 series 3 V 3 to 33 V 400 mW - 8 p . 1 watt
    Battery snaps for PP3 - $6 p$ for PP9
    L.E.D.s 3 mm . \& 5 mm . Red, Green, Yellow - 10 p . Grommets $3 \mathrm{~mm}-2 \mathrm{p}$, 5 mm Red flashing L.E.D.'s require 5 V supply only
    Mains indicator neons with 220 k resistor.
    Mains indicator neons with 220 k resistor
    20 mm fuses 100 mA to 5 A . Q blow $5 \mathrm{p} . \mathrm{A} /$ surge 8 p . Holders, chassis, mounting High speed pc drill $0.8,1.0,1.3,1.5,2.0 \mathrm{~m}-30 \mathrm{p}$. Machines 12 V dc HELPING HANOS 6 ball joints and 2 croc clips to hold awkward jobs Glass reed switches with single pole make contacts - 8 p . Magners $0.1^{\prime \prime}$ Stripboard $21^{1 / 2^{\prime \prime}} \times 1^{\prime \prime} 9$ rows 25 holes -20 o. $3^{3} 4^{\prime \prime} \times 21_{2}^{\prime \prime} 24$ rows 37 holes Jack plugs $2.5 \& 3.5 \mathrm{~m}$ Sockets Panel Mig 2.583 .5 m

    ## TRANSISTORS

    BC107/8/9-12p, BC547/8/9-8p, BC557/8/9-8p, BC182, 182L, BC183, 183L, BC184 184L, BC212, 212L-10p
    BC327, 337,337L-12p, BC727, 737-12p, BO135/6/7/8/9-25p, BCY70-15p, BFY50/51/52-20p
    BFX88-15p, 2N3055-50p. TIP31, 32-30p. TIP41,42-40p, BU208A-£1.20, BF195, 197-12p All prices are inclusive of VAT. Postage 30p (free over E5). Lists Free.

