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## WITHOUT TEARS

AUTUMN is in some respects an even more auspicious time for new resolutions than the new year. It is certainly true that the approach of the September equinox turns thoughts of many of us towards ways to occupy the longer evenings that lie ahead. Personal needs and fancies vary widely. There are those who prefer a casual, maybe inconsequential, kind of pastime. And there are those who wish to make a commitment to serious study over several months. In the latter case the choice of subject may be hard to resolve. So to the undecided may we put in a word or two in favour of electronics as an appropriate and interesting subject for home study.

The basic principles that govern all electronic circuits are not difficult to learn and can be readily proved by uncomplicated experiments carried out on any table top. With batteries as the source of power and the use of a circuit block for assembling components without any recourse to a soldering iron, the introduction to electronic circuits couldn't be easier.

All this applies to our new twelve-part series Teach-In 84. We commend this home study course to all who would like to understand what makes electronics tick. No previous knowledge of electronic or electrical theory is needed. Anyone of any age should be able to follow these articles and rapidly become accustomed to handling components through the experiments. Furthermore all components needed for the twelve parts of Teach-In 84 can be purchased as a complete kit from a number of our advertisers. All of which goes to reinforce our claim to present "Electronics Without Tears".

## WIDER SCOPE FOR SCHOOLS

Talent for designing and building electronic equipment abounds amongst our secondary school students. This has been amply demonstrated over the past two years through the Schools Electronic Design Award Competition (SEDAC). The projects submitted have been impressive, covering an interesting range of applications within the limitations imposed by the rules which restricted the designs to those having a useful function in a school's science laboratory.

Electronic technology is versatile, and perhaps such limitations ought to be eased in future so that the scope for designs could be extended. This at any rate is the feeling of both sponsors, Mullard and Everyday Electronics: Therefore, the rules have been amended to permit entry of any electronic design having some specific use within a school's premises, or at any external event, indoors or outdoors, in which the school participates.

We confidently expect a large entry of designs and models representative of all aspects of applied electronics, in this coming third year of SEDAC. Scholars and teachers should note the particulars regarding SEDAC 1984 given elsewhere in this issue. Requests for registration forms should be sent in without delay.

## Readers' Enquiries

We cannot undertake to answer readers' letters requesting modifications designs or information on commercial equipment or subjects not published by us. All letters requiring a personal reply should be accompanied by a stamped self-addressed envelope.

We cannot undertake to engage in discussions on the telephone.
Component Supplies
Readers should note that we do not supply electronic components for building the projects featured in EVERYDAY ELECTRONICS; but these requirements can be met by our advertisers.

All reasonable precautions are taken to ensure that the advice and data given to readers are reliable. We cannot, however, guarantee it and we cannot accept legal responsibility for it. Prices quoted are those current as we go to press.

Back Issues
Certain back issues of EVERYDAY ELECTRONICS are available worldwide price $£ 1.00$ inclusive of postage and packing per copy. Enquiries with remittance should be sent to Post Sales Department, IPC Magazines Ltd., Lavington House, 25 Lavington Street, London SE 1 OPF. In the event of non-availability remittances will be returned.
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BY R.A.PENFOLD INTER


While communicating from one room to another in a loud voice can be quite effective, a more civilised and reliable way is to employ a simple intercom such as the one described here. The two stations are connected by a twin screened lead which can be up to 20 metres long if necessary.

The Home Intercom uses the loudspeaker at each station as a microphone as well. Although a loudspeaker does not give high quality output when compared to "proper" microphones, the quality is more than adequate for understandable speech.

Many intercoms use very simple amplifiers which give a reduction in sound quality, and a significant one at that. Furthermore, there is often no volume control so that overloading of the output stage occurs creating severe distortion.

This design achieves good quality reproduction by intercom standards as an integrated amplifier is used, and a volume control enables overloading to be prevented. Slight tailoring of the amplifier's frequency response is used to further improve matters.

## THE CIRCUIT

The full circuit diagram of the Home Intercom is given in Fig. 1. The method
of using the unit and the switching are common with virtually all intercom designs, that is, a high gain amplifier having its input fed from a loudspeaker and its output coupled to a second loudspeaker. The input loudspeaker is used as a moving coil microphone. By using simple send-and-receive changeover switching, the roles of the two loudspeakers can be reversed, and communications in both directions is then possible.

## MAKING A CALL

For someone at the main station to call the remote station, it is necessary for them to switch the unit on, set the SEND/RECEIVE switch (S2) to "send", and talk into the microphone. The unit must also have some way of enabling the remote station to call through to the main one, and this is made possible by having the SEND/RECEIVE switch biased to the "receive" position, and having an extra battery and Call switch in the remote unit.

If the call switch ( $\mathbf{S} 1$ ) is operated, this additional battery (B1) supplies power to the amplifier, and the person at the remote unit is able to talk through to the main station (bearing in mind that the

SEND/RECEIVE switch will be in the "receive" position).

The person at the main unit then switches the unit on and operates the send/receive switch as necessary. The CALL switch is only operated when initially establishing contact with the main station, and during normal operation, power is obtained from the battery (B2) at the main station.

## AMPLIFIER

This design uses a ULN2283 integrated amplifier (IC1) in the output stage, and this gives good quality provided the circuit is not overdriven. The maximum output power is around 100 to 150 mW , and this is sufficient for an intercom where high volume levels are not normally required.

This i.c. has an internal negative feedback circuit which sets the voltage gain of the device typically at 43 dB . If no external feedback resistor is used, as in this application where the full voltage gain of the device is needed, pin 1 is coupled to 0 V by C 6 and no series resistor is used.

The input of IC1 connects direct to the slider of the volume control VR1 and no d.c. blocking capacitor is needed. C5 attenuates the high frequency response of

Fig. 1. Circuit diagram of the two-station Home Intercom. The connecting cable can be up to 20 m long.

the unit and prevents IC 1 from becoming unstable.

Although ICl gives quite a high voltage gain, the output from a high impedance loudspeaker when used as a microphone is extremely small, and will normally be under 1 mV . A considerable boost in gain is therefore needed, and this is provided by TR1 which is used as a conventional high gain common emitter stage.

This gives an increase in voltage gain of around 40 dB . C4 rolls off the high frequency response of the circuit, and this is mainly necessary due to stray capacitance in the SEND/RECEIVE switching which would otherwise give sufficient feedback over the amplifier to produce strong oscillation.

The values of $\mathrm{C} 2, \mathrm{C} 3$, and C 6 have purposely been made rather low so that the low frequency response of the circuit is severely attenuated. This is done to compensate for the excessive bass output of a loudspeaker used as a microphone, which otherwise tends to seriously impair intelligibility.

## COMPONENTS

| Resistors |  |
| :---: | :---: |
| R1 | $1.8 \mathrm{M} \Omega$ |
| R2 | $4.7 \mathrm{k} \Omega$ |
| R3 | 1 k , |
| All $\frac{1}{4} \mathrm{~W}$ ca | boin $\pm 5 \%$ |
| Capacitors |  |
|  |  |
| C1,7,8 | $100 \mu \mathrm{~F} 10 \mathrm{~V}$ elect. (3 off) |
| C2 | 100nF polyester |
| C3 | 47 nF polyester |
| C4 | 3.3 nF cerâmic |
| C5 | 330 pF ceramic |
| C6 | $2 \cdot 2 \mu \mathrm{~F} 63 \mathrm{~V}$ elect. |

## Semiconductors

| TR1 | BC109C silicon non |
| :--- | :--- |
| IC1 | ULN2283 audio power |
|  | amplifier |

## Miscellaneous

| VR1 | $10 \mathrm{k} \Omega$ log. control potentiometer |
| :---: | :---: |
| LS 1,2 | 40 to $80 \Omega$ miniature |
|  | loudspeaker |
| S1 | push-to-make |
|  | momentary action |
|  | push-button |
| S2 | d.p.d.t. miniatur biased toggle |
| S3 | s.p.s.t. miniature toggle |
| B1 | 9 V PP3 battery |
| B2 | 9V PP6 battery |
| Plastic cases with aluminium lids, |  |
| $160 \times 100 \times 60 \mathrm{~mm}$ and $110 \times$ |  |
| $70 \times 50 \mathrm{~mm} ; 0.1 \mathrm{in}$ matrix strip- |  |
| board, 36 holes by 12 strips; |  |
| control knob; battery clip (2 off); |  |
| 20 m twin-core screened cable; |  |
| $7 / 0.2 \mathrm{~mm}$necting wire. p.v.c. sleeved con- |  |
|  |  |



The complete Home Intercom system in matching cases.

The current consumption is about 13 mA under quiescent conditions, but as the ULN2283 has a class A/B output stage the current drain increases substantially when the unit is used at high volume levels.

## MAIN STATION

The main station is housed in a plastic box with an aluminium front panel and outside dimensions $160 \times 100 \times 60 \mathrm{~mm}$.

A pattern of holes are drilled on the left section of the front panel to produce a speaker grille, and the loudspeaker is glued in place behind this. The three controls are mounted on the right-hand side of the front panel, and a small hole for the twin screened connecting cable is drilled in one side of the case.

The amplifier is built on a 0.1 in matrix stripboard using the component layout shown in Fig. 2, and this requires a board


Fig. 2. The stripboard component layout and trackbreak details of the Home Intercom main station. This board slots into the grooves in the plastic case.



The main station of the Home Intercom showing the method of mounting the circuit board. The larger PP6 battery is also housed in this case.


The remote station wiring and loudspeaker mounting details. The smaller PP3 battery in this unit is only used when initially making a call.
having 36 holes by 12 strips. The resistors, capacitors, and link wires are fitted first, and then TR1 and IC1, ensuring that the semiconductor and electrolytic capacitors are inserted the right way round. On IC1, pin 2 and 3 and 5 and 6 are manufactured connected together in pairs.

Note the four breaks in the copper strips on the underside of the board which are made with the appropriate tool or a small drill bit.

## CONNECTING CABLE

The connecting cable is passed through the hole in the case and wired into the circuit. The best wire to use is twin screened type since it is inexpensive, and the screening prevents stray pick-up of mains hum and radio signals in this cable. The capacitance in the lead will be quite high if a long lead is used, but this should not significantly affect the performance of the system as the microphones have a low output impedance. A three-way, nonscreened cable is not recommended for use as the connecting cable, and would
almost certainly give excessive breakthrough of unwanted signals.

The main station is completed by adding the remaining wiring, and slotting the component board into a set of vertical guide rails in the case, so that it fits into the space between the loudspeaker and the controls. The battery fits into the space behind the loudspeaker with foam rubber material to keep it in place when the front panel is fitted.

## REMOTE STATION

The remote station is fitted in a case that matches the one used for the main station, but it has smaller dimensions of about $110 \times 70 \times 50 \mathrm{~mm}$. A similar matrix of holes are drilled to form a speaker grille on the left-hand side of the front panel, as for the main station.

Sl is positioned on the right-hand side of the panel and a small entrance hole is drilled for the connecting cable in one side of the case. Fig. 3 shows the small amount of wiring needed to complete the remote unit.


Fig. 3. The remote station wiring details. The PP3 battery is secured with doublesided adhesive foam pads.


The main station front panel.


The remote station front panel.

## TESTING

A quick way of checking the completed Home Intercom is simply to switch on the unit and turn up the volume with the two stations in the same room, facing each other but spaced some distance apart. This should produce a "howling" sound from the main unit due to accoustic feedback.

Operating the send/receive switch should result in the howling coming from the remote unit. Switching off the Intercom and operating the CALL button on the remote unit should turn the amplifier on again and produce the howling sound from the main unit.

All that remains is to take the two stations into separate rooms and test with normal speech and finally mount the units in the chosen locations.


## BY DAVE BARRINGTON

## Catalogues Received

The latest, over 200-page catalogue from Benningeross Electronics contains such items as soldering irons, smoke detectors, kits, telephone hand-set, case, tools and p.c.b. drills, cassette motors, loudspeakers and an extensive range of semiconductor and i.c. devices.

It is a pity that although the catalogue contains probably one of the largest range of individual items presently listed by an advertiser, the quality of printing leaves a lot to be desired. This is particularly relevant to the reproduction of the photographs. However, at $£ 1$ this 230 -page catalogue is a very welcome addition to the components reference library.

Copies of the Benningcross Components Catalogue may be obtained from: Benningeross Electronics, Dept. EE, 67 Vicarage Road, Watford, Herts WD1 8EJ.

It may, at first glance, seem strange to mention a Mail Order catalogue in these pages. But a quiet revolution has been taking place in the "popular" mail order catalogue business. With the increasing appeal of electronics and computing, particularly in the games area, the amount of stock carried has increased quite considerably.

This is quite evident in the new Argos catalogue where, if you include digital watches, electronic typewriters (computer compatible) and in-car entertainment, nearly 50 -pages of the 284 -pages are devoted to electronics-based products.

## EE PRINTED CIRCUIT BOARD SERVICE

Printed circuit boards for certain EE constructional projects are now available from the EE PCB Service, see list right. These are fabricated in glass-fibre, and are fully drilled and roller tinned. All prices include VAT and postage and packing. Remittances should be sent to: EE PCB Service, Everyday Electronics Editorial Offices, King's Reach Tower, Stamford Street, London SE1 9LS. Cheques should be crossed and made payable to IPC Magazines Ltd.
Readers are advised to check with prices appearing in current issue before ordering.
*Set of four boards.
M.I.T.-Microcomputer Interfacing Techniques, 12-Part Series.

This may seem only a small proportion of the items listed, but when you consider the "state-of-the-art" products carried, it is quite an achievement.

Products stocked include: Silver Reed typewriter, with computer interface for BBC, Dragon, Tandy and Video Genie machines; the Casiotone MT-70 49-key, 4-octave electronic keyboard; Mattel Synsonic Drum Synthesiser and the Aquarius home computer (see New Products); the CBS Coleco Vision Expansion Module 2, car hazard driving video; and the Vectrex video game system with its own built-in monitor screen. The latter four items only released this month.

## Teach-In '84

This month we start the first instalment of a new home-study course for the newcomer to electronics.

This 12-part series entitled Teach-In 84 has been specially commissioned and uses a solderless "breadboard" system so that even the complete novice can tackle the demonstration "exercises".

Arrangements have been made with our advertisers to supply kits of parts, including the Ebbo Block System, for the complete 12 -part series.

For details of suppliers and page number of their advertisement, see page 638 .

## CONSTRUCTIONAL PROJECTS

Short Wave Radio
The variable tuning capacitors used in the Short Wave Radio are stocked by

Ambit, Benningcross, Electrovalue, Rapid, Maplin and Magenta Electronics.

The transistor type BF244B used in the detector/r.f. amplifier stage is listed by Cricklewood and Rapid Electronics.

The aerial tuning coils are available direct from Denco (Clacton) Ltd., Dept EE, 355/9 Old Road, Clacton-on-Sea, Essex CO15 3RH. When ordering specify: Miniature Dual Purpose, Green, Ranges 3, 4 and 5 .

## Immersion Heater Tell-Tale

The thermistor GL16 used in the Immersion Heater Tell-Tale is a miniature bead type. It has a negative temperature coefficient (n.t.c.) and a resistance of $1 \mathrm{M} \Omega$ at $20^{\circ} \mathrm{C}$ and a minimum resistance of $170 \Omega$.

This thermistor is sometimes referred to as types G16 and THB11. It is available from Maplin and most major suppliers

The double socket box and blanking plate should be available from any good electrical store. The LM324N quad op-amp is available from the following: Ambit, Magenta, Rapid and TK Electronics

## Home Intercom

The audio power amplifier, type ULN2283, used in the Home Intercom appears to be only available from Ambit.

## Temperature Transducer Amplifier

The sensing device used in the Temperature Transducer Amplifier is the 590 kH and as far as we are aware is only available from RS Components (Order code $308-809$ ). We would point out however that RS will only supply to recognised dealers and readers will have to order through their local component supplier.

## Microcomputer Interfacing Techniques

All the "special" components called for in this month's constructional projects contained in the Microcomputer Interfacing Techniques series relate to the Digital-toAnalogue Converter (DAC) board.

The s.i.l. 8-commoned resistor pack (Code 140-271), the Darlington driver ULN2803 i.c. (Code 303-438) and the DAC0800 digital-to-analogue converter (Code 309-458) were obtained from RS Components. All these items should be ordered through your local RS stockist.

We do not expect any component purchasing problems for the Model Car Petrol Tank and the Electronic PendulumSEDAC Prize Winner

| PROJECT TITLE | Order Code | Cost |
| :---: | :---: | :---: |
| Eprom Programmer, TRS-80 (June 83) | 8306-01 | £9.31 |
| Eprom Programmer, Genie (June 83) | 8306-02 | £9.31 |
| Eprom Programmer, TRS-80 \& Genie (June 83) | 8306-03 | £1.98 |
| User Port Input/Output M.I.T. Part 1 (July 83) | 8307-01 | £4.82 |
| User Port Control M.I.T. Part 1 (July 83) | 8307-02 | £5.17 |
| Storage 'Scope Interface, BBC Micro (Aug 83) | 8308-01 | £3.20 |
| Car Intruder Alarm (Aug 83) | 8308-02 | £5.15 |
| Electronic Die (Aug 83) | 8308-03 | £4.56 |
| High Power Interface M.I.T. Part 2 (Aug 83) | 8308-04 | £5.08 |
| Pedestrian Crossing Simulation M.I.T. Part 2 (Aug 83) | 8308-05 | £3.56 |
| High Speed A-to-D Converter M.I.T. Part 3 (Sept 83) | 8309-01 | £4.53 |
| Signal Conditioning Amplifier M.I.T. Part 3 (Sept 83) | 8309-02 | £4.48 |
| Stylus Organ (Sept 83) | 8309-03 | £6.84 |
| Distress Beacon (Sept 83) | *8309-04 | £5.36 |
| Distress Beacon Pocket Version (Sept 83) | 8309-05 | £3.98 |
| D-to-A Converter M.IT. Part 4 (Oct 83) | 8310-01 |  |
| High Power DAC Driver M.I.T. Part 4 (Oct 83) | 8310-02 | f5.13 |
| Electronic Pendulum (Oct 83) | 8310-03 | £5:43 |



WITH the present high cost of electricity, many householders are looking for ways of reducing energy costs. One possible area of saving is in the use of immersion water heaters.

Since hot water is less dense than cold, the level of hot water in the storage tank "grows" downwards from the top. If a check is made on the hot-water level then only the amount required need be heated. The traditional method is to check by hand but this is difficult where a modern close-fitting jacket is used or if the cylinder is remote from the kitchen. The result is an excess of hot water; heated with energy which will be dearly paid for.

## L.E.D. DISPLAY

The Immersion Heater Tell-Tale "feels" the tank electronically at four levels and operates a corresponding display of light emitting diodes. If, for example, the tank is half full of hot water, then the top two l.e.d.s will light but the bottom pair will stay off.

The circuit only works when a test button is pressed so normally there is nos drain on the battery. The current consumption is only about 20 mA with all four l.e.d.s on, so the battery should give many months of service.

Apart from the four sensors on the hot-water cylinder and associated wiring,
the project is housed in a standard electrical double socket box with blanking plate. This may be of the surfacemounted or flush (sunk in the wall) type. If the project is on view it will blend with wall'switches and power sockets.

Before commencing construction, it would be a good idea to check that it is possible to run the necessary wires from the cylinder to the proposed position of the unit.

## CIRCUIT

The circuit is based on an LM324 Operational Amplifier i.c. This device contains four separate operational amplifiers and each is made responsible for one particular water level.

Fig. 1 shows the complete circuit diagram. It can be seen that the inverting inputs of the op-amps (marked "-") are connected to a potential divider consisting of two equal value resistors, R5 and R6. This sets the inverting inputs to about one-half of the battery supply voltage (about 4.5 V ). The other inputs, the non-inverting inputs (marked " + "), are each connected to a separate potential divider. One element of which is a thermistor (RTH1) and will be placed in contact with the water tank. The other consists of the fixed and variable resistors, R1 and VR1, connected in series.

## THERMISTORS

Thermistors are heat-sensitive devices which reduce their electricai resistance when warmed (in the case of the negative co-efficient type used here). The voltage at the non-inverting input will rise, therefore, and eventually exceed the fixed voltage at the inverting input.

It is a characteristic of an op-amp that when the voltage at the non-inverting input exceeds that at the inverting input, it switches on and an l.e.d., placed in its output circuit, will light. VR1 can be adjusted so that the l.e.d. lights at the correct temperature. The purpose of R1 is to limit the current flowing through the thermistor should VR1 be set to nearzero resistance. The resistor in series with the l.e.d. limits its current to a safe working level.

All four channels act in an identical manner. As the hot-water level progresses downwards in the tank, the l.e.d.s will light in sequence.

Since this circuit compares two voltages applied to the inputs, it does not matter if the battery runs down to some extent. The only effect is that the l.e.d.s will glow less brightly.

## COMPONENT BOARD

Construction is based on a piece of $0 \cdot 1$ in matrix stripboard, 17 strips by 27

## COMPONENTS

## Resistors

| R1-4 | $10 \mathrm{k} \Omega \cdot(4$ off $)$ |
| :--- | :--- |
| R5.6 | $100 \mathrm{k} \Omega$ (2 off) |
| R7-10 | $1 \mathrm{k} \Omega(4$ off) |
| All $\frac{1}{4}$ or $\frac{1}{2} W$ carbon $\pm 10 \%$ |  |

## Semiconductors

D1-4 TIL220 5 mm red I.e.d. (4 off)
IC1 LM324N quad op-amp

* RTH1-4 GL16 miniature bead. negative temperature coefficient thermistor (4 off)


Miscellaneous

VR1-4
S1 miniature push-button TB1,2 5-way terminal block (2 off)
B1 9V PP3 battery
Stripboard, 0.1 in matrix, 17 strips by 27 holes; double socket box and blanking plate; 14-pin i.c. holder; battery clip; $7 / 0.2 \mathrm{~mm}$ connecting wire ( 5 colours); solid copper wire; mounting hardware.

Approx cost.
Guidance only
holes. Refer to Fig. 2 and begin by soldering the i.c. socket into position. Follow by making the seven links then add the components. Do not insert ICl into its holder at this stage. Make the breaks in the copper tracks as indicated using a spot-face cutter or a small twist drill turned by hand. Note that no breaks are made in tracks $E$ and $I$ between the i.c. pins. Solder 50 mm lengths of connecting wire to each of strips $B, D, J, L$ and $Q$. These will be connected to the l.e.d.s later. Solder wires for S1 and the battery connector.

## TERMINAL BLOCK

Prepare the terminal block TB1 by securing 20 mm long bare copper wires to terminals 1 to 4. Leave TB1/5 unconnected. Best results are obtained by using the stiff wire obtained from the ends of discarded $\frac{1}{2} \mathrm{~W}$ resistors. Shape the leads as necessary and connect them to strips $A$, $F, H$ and $O$ on the circuit panel (see photograph). Trim off excess wire.

Before proceeding, check the circuit panel for errors. In particular, make sure that no accidental "bridges" have been made between adjacent copper strips.


The circuit board, l.e.d.s and terminal block TB1 mounted to the rear of the blanking plate of the double socket box. Note how TB 1 has been bent upwards.

Finally, insert IC1 noting that it must be fitted the correct way round. Take care not to bend the pins.

Fig. 1. The circuit diagram of the Immersion Heater Tell-Tale. D1 indicates the temperature sensed by RTH1 at the top of the tank through to D4 indicating the temperature at the bottom.


## CASE

Attention may now be given to the casing. Double socket boxes and blanking plates to suit are stocked by electrical dealers and D.I.Y. shops. Drill holes in the blanking plate for the l.e.d.s as shown in the photograph. In the prototype these were 5 mm diameter and spaced 12 mm apart. The l.e.d.s may be secured behind these holes to give a neat appearance to the front panel. Alternatively, they could be fitted using proper mounting kits. Drill a hole for the push-button switch (S1) and mount this in position.

Mount the l.e.d.s so that the cathode (k) leads are all to the right-hand side of the circuit panel (see photograph). Use a very small quantity of quick-setting epoxy resin adhesive to secure. When the adhesive has hardened, carefully solder all the cathodes of the l.e.d.s together and connect to the wire already leading to strip $Q$ on the circuit panel. Connect the wires leading to strips $L, J, D$ and $B$ to the anode leads of D1, 2, 3 and 4 , respectively. Solder the wires to the switch. Connect the non-battery side of the switch to TB $1 / 5$ as shown in Fig 2.

## THE THERMISTORS

Thermistors are very delicate components so great care is needed in preparation. They may also be damaged by excessive heat from the soldering iron. The original leads are very short and must be extended to about 500 mm to enable them to be connected to TB2 (see Fig. 2). Begin by taking eight lengths of $7 / 0.2 \mathrm{~mm}$ stranded wire and strip the insulation 5 mm from each end. It is helpful if five colours are available; four pieces of one colour and one each of the others. Each thermistor will then be fitted with one common and one individually coloured lead.

Take one thermistor, slide a short piece of narrow sleeving-insulation stripped from scrap connecting wire-onto each

lead so that about 10 mm of bare wire is left exposed. Slide a 40 mm length of thicker sleeving over two of the 500 mm lengths of wire. Solder the wires to the thermistor leads. Finally, slide the thick sleeving over the joints to insulate them (see Fig. 3).

Ensure that the narrow sleeving is pushed hard up to the body of the thermistor otherwise the leads are likely to twist and cause a short circuit. A little glue may be used to keep the sleeving in position. Repeat the procedure for the other three thermistors and twist the leads for each together.

The thermistors may now be secured temporarily to the copper water cylinder; p.v.c. tape may be used for this. Position RTH1 about one-fifth of the distance
down the tank, RTH2 about two-fifths and so on. Make certain that each thermistor makes good thermal contact with the cylinder. Each common wire connects to TB $2 / 5$. The remaining wires from RTH1, RTH2, RTH3 and RTH4 connect to TB2/1, TB $2 / 2$, TB $2 / 3$ and TB2/4, respectively. A five-way cable connects TB2 at the cylinder to TB1 at the unit.

## TESTING

Start with a cold-water tank. Connect the battery and press the button. It is possible that one or more l.e.d.s will light. If so, adjust the appropriate variable resistor anticlockwise until it goes off.

Heat up the water. When the water at the level of RTH1 is judged to be "hot"

Fig. 2 (above and left). The circuit board layout and interwiring diagram. Note that the connections from the stripboard to TB1 are made with solid wire. All other wiring is carried out with a flexible multi-strand type. The thermistors are secured to the tank (see above right) at one-fifth intervals and terminated at TB2.
Fig. 3 (right). The method of extending the leads of the thermistors and sleeving the joints.

> Note-Since completing the prototype Immersion Heater Tell-Tale, the author discovered that the considerably cheaper VA1067S rod thermistor can be used in place of the GL16 bead thermistor (RTH1 to 4). These are available from Maplin and, if used, must not make metallic contact with the copper cylinder.



BY GEORGE HYLTON

WElcome to Teach-In 84. In this series, which will run for one year, we'll avoid the need for soldering by making use of a commercial electronic breadboard. Components and connecting leads are simply plugged-in.

The recommended breadboard is the EBBO Building Block System. This has a small "tray" into which various modules are slotted. A variety of modules are available for use in the EBBO System. We'll use those which accept individual components and transistors, battery holder modules, integrated circuit modules and power supply (battery) rails.

Certain items (such as earphones) will, for convenience, first be attached to a piece of electrician's terminal block, which will then be connected to the breadboard with long leads.

## COMPONENTS

All components and other items required for the experiments in the entire series are listed in Table 1.3 and can be obtained from a number of retailers as complete kits. See page 638.

## TOOLS

Only a few tools are required: a small pair of long-nosed pliers and a small pair of wire cutters, both preferably with insulated handles; a small screwdriver, $\frac{1}{8}$ in blade.

## TEST METER

A small inexpensive multimeter, to measure volts, milliamperes and ohms, is needed. The multimeter recommended for this series is the KEW7S and is available to Everyday Electronics readers at a special price. See page 657.

If you already have a multimeter with a sensitivity of 1,000 ohms per volt or greater it should be usable, but some of the voltage readings it gives may differ a little from those with the KEW 7S.

It is desirable that your meter should have a d.c. voltage range of less than 10 volts (V) full scale deflection, though you can get by with a range of $0-10 \mathrm{~V}$ for most experiments. Don't attempt to use your meter on its current ranges. (It may then be too susceptible to overload.)

## A.C., D.C., AND SIGNALS

In electronics, most of the circuits carry currents which are mixtures of d.c. (direct or steady current) and a.c. (alternating or time-varying current). Usually the d.c. comes from a battery or the equivalent power supply. It is there to make the transistors in the circuit work. Once working, the transistors are used to handle a.c. signals.

The word "signal" is used very loosely in electronics. Originally, of course, a signal was just that-a beacon or a light or a flag, for example. Then when the electric telegraph was invented it came to mean the electrical equivalents of these visible signals.

Morse Code signals, for instance, are currents in a telegraph line, switched on or off by tapping a contact key. When telephone transmission became possible it was natural to enlarge the meaning of signals to include speech. And so onto radio signals, television signals and digital signals as used in computers.


The EBBO Building Block System used in the Teach-In 84 series. (Left) the Basic Integrated Circuit Starter Pack. (Right) the Basic Discrete Component Starter Pack.


Other equipment required for Teach-In 84. Small Multimeter (Eagle KEW7S), long-nosed pliers, wire cutters, small screwdriver, a piece of Scotchbrite scouring material, and multimeter adaptor lead.

Any voltage or current which carries some sort of information-audible, visible, mathematical or whatever, is now called a signal.

Interesting though some of these signals may be you will never learn how to design or understand circuits for them unless you first have a firm grasp of the principles of d.c. Fortunately many of these principles can easily be extended to a.c. So in learning d.c. circuit design you are laying the foundations for everything else.

## CURRENT

We all know that electric currents flow through wires. An electric current is really a flow of electric charges, carried by electrons or other kinds of chargecarrying particles. But we don't need to go into the physics of the thing. As far as we are concerned, d.c. electric currents are pushed around circuits by batteries. We'll use a 9 V dry battery, or perhaps enough 1.5 V cells to make up 9 V .

Current flows out of the positive terminal of a battery, goes round the circuit, and flows back into the negative terminal. It can only flow if there is a conducting path connecting positive to negative. This path can be a wire, a computer, your own body, anything that conducts electricity.

Just as much current flows back into the negative battery terminal as comes out at the positive. If it didn't, there would be a "traffic jam" of electric charges somewhere along the line and this would itself inhibit current flow.

At first sight, an electric circuit seems like a case of perpetual motion. Surely, the current can go round and round for ever. If as much gets back into the battery as leaves it, then nothing is lost! But you know it can't be so.

## ENERGY

Energy is expended in driving the current. Strictly speaking, the energy is not so much expended as changed from one form to another. What starts life as electrical energy is turned into some other kind, usually heat, which is lost into the air. You can see the process at work in an ordinary electric light bulb, whose filament (a thin conductor) gets intensely hot.

There is a limit to the amount of current which any power source can safely deliver without either internal damage or (in the case of less severe overloads) shortening its life. For a miniature 9 V pocket transistor radio battery this safe current may be only a few thousandths of an ampere, that is, a few milliamperes $(1,000 \mathrm{~mA}=1 \mathrm{~A})$.

## RESISTANCE

We need some means of restricting and adjusting the current we draw from our battery. This is the job of resistance.

In the pioneering days of electrical research, Georg Simon Ohm found that some materials conducted currents less efficiently than others. Long wires were worse than short ones. Thin wires were worse than thick ones. Iron wires were worse than copper ones. By selecting the length, thickness and material of a wire the amount of current could be adjusted.

Looked at another way, some conductors offer more resistance to current flow than others. More battery voltage is needed to push the same amount of current through a high resistance than through a low one.

## RESISTORS

Metal wires were originally used to produce resistance, or, as we'd say now, to make a resistor. For a high resistance a long length of wire was needed. To pack this into a reasonable space often two parallel rows of nails were driven into a wooden baseboard (Fig. 1.1) and the wire threaded through them zigzag fashion.

Nowadays, we are fortunate in being able to buy, for a few pence each, compact resistors in the form of ceramic rods on which are deposited a resistive "track" in the form of a thin film of metal (for the


TABLE 1.1: RESISTOR COLOUR CODE

| Band (ring) Colour | 1 (1st Figure) | (2nd $\stackrel{2}{\text { Figure) }}$ | $3$ <br> (Multiplier) | (Tolerance \%) |
| :---: | :---: | :---: | :---: | :---: |
| BLACK | 0 | 0 | - | - |
| BROWN | 1 | 1 | 10 | 1 |
| RED | 2 | 2 | 100 | 2 |
| ORANGE | 3 | 3 | \$00 | 3 |
| YELLOW | 4 | 4 | 10,000 | 4 |
| GREEN | 5 | 5 | 100,000 | - |
| blue | 6 | 6 | \$000000 | - |
| VIOLET | 7 | 7 | 10000,000 | - |
| GREY | 8 | 8 | 100000000 | - |
| WHITE | 9 | 9 | 1000000000 | - |
| GOLD |  |  | 0.1 | . 5 |
| SILVER |  |  | 0.01 | 10 |


| Examples: |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| (1) BAND | COLOUR |  |  | (2) BAND | COLOUR |  |
| 1 | RED | 2 |  | 1 | ORANGE |  |
| 2 | VIOLET | 7 |  | 2 | ORANGE |  |
| 3 | YELLOW | $\times 10,000$ | $=270.000$ OHMS | 3 | RED | $=3.3 \mathrm{k} \Omega$ |
| 4 | GO10 |  | OR 270k $\Omega$ | ABSEN | OF BAND | $4= \pm 20 \%$ TOLERANCE |


lower resistances) or carbon (for the higher ones) (Fig. 1.2).

The standard range begins at 1 ohm and extends upwards to 10 million ohms (ten megohms, 10M).
Interesting to note the original zigzag is still used as the symbol for resistance on circuit diagrams. You may come across an alternative symbol, a thin rectangle, in certain publications.

## COLOUR CODE

The carbon or metal-film resistance track is insulated by a layer of special paint which gives the resistor its body colour. The electrical value of the resistance (the number of ohms) is usually indicated by an additional colour code. Bands of colour are printed round the body near one end, see Table 1.1.

The first two bands indicate the first two figures of the resistance value. This means that they can stand for anything between " 1 " and " 99 ". Example: Red means " 2 ". If the first two bands from the end of the body are both red, the resistance begins with 22.

The next colour tells you if there are any noughts after the 22. For example, since red means " 2 ", if the third band is also red this means that two noughts

must be added. So the resistance value is 2,200 ohms.

On diagrams this may be written $2,200 \Omega, 2 \cdot 2 k$, or 2 k 2 . Here the Greek capital letter omega ( $\Omega$ ) is the standard symbol for ohms, and $k$ means 1,000 ohms ( 1 kilohm).

What if no noughts need be added? In this case, the third band is black, which means zero-thus zero noughts. So a $22-$ ohm resistor is coded red, red, black. On diagrams this is written $22 \Omega$ or 22 R .

## TOLERANCE

There is a fourth coloured band. This tells you how accurate the resistance value is. Precision is expensive. Fortunately, most circuits will work with resistances which vary from the chosen

Table 1.2

| TERM | SYMBOL | MULTIPLIER | POWER |
| :--- | :--- | :--- | :--- |
| GIGA | $G$ | $1,000,000,000$ | $10^{9}$ |
| MEG | M | $1,000,000$ | $10^{6}$ |
| KIL | $k$ | 1,000 | $10^{3}$ |
| MILLI | m | 0.001 | $10^{-3}$ |
| MICRO | $\mu$ | 0.000001 | $10^{-6}$ |
| NANO | $n$ | 0.000000001 | $10^{-9}$ |
| FEMTO | 1 | 0.000000000000001 | $10^{-15}$ |

## SEMICONDUCTORS:



LIGHT EMITTING DIODES (LED)


VARICAP DIODE


IRANSISTORS


IC's (INTEGRATED CIRCUITS)

value, so long as they don't vary too much.

The permitted amount of variation in value for a given type of resistor is called the tolerance, and is given as a percentage variation from the nominal value.

The Table 1.1 gives the complete colour code. It will seem awkward at first but you'll soon get used to it. The great practical advantage is that since the bands go right round the body of a resistor it is always possible to see them. A printed number (which is also sometimes used) can be obscured if it is facing the breadboard rather than the outside world.

Fig. 1.3. Outline drawings of the component types included in the Teach-In 84 Kit. Characteristic markings indicating values and polarity of lead-out wires are given; also circuit symbols.

This photo shows the full kit of components for Teach-In 84. Starting top left: integrated circuits ('chips') (6) and holders (2) all mounted on conductive foam, potentiometers (2), crystal earphone;

Connecting wire, 12 -way terminal blocks (2). ferrite aerial rod, disc ceramic capacitors (5), light-emitting diodes (3), varicap diodes (1 dual), signal diodes (4), transistors $(5+5)$;
Capacitors: $1 \mu \mathrm{~F}$ polystyrene (3), $10 \mu \mathrm{~F}$ electrolytic (2), $2 \cdot 2 \mu \mathrm{~F}$ electrolytic (1), 10 nF polystyrene (3), $100 \mu \mathrm{~F}$ electrolytic (2), $1,000 \mu \mathrm{~F}$ electrolytic (1), Resistors (100).


## TABLE 1.3

Components required for experiments during the 12 Parts of Teach- $\ln 84$.
Complete kits for all these items may be obtained from retailers advertising in this magazine, see list on page 638.

## RESISTORS

| Quantity | Value | Quantity | Value |
| :---: | :--- | :---: | :--- |
| 10 | $100 \Omega$ | 10 | $33 \mathrm{k} \Omega$ |
| 10 | $330 \Omega$ | 10 | $100 \mathrm{k} \Omega$ |
| 10 | $1 \mathrm{k} \Omega$ | 10 | $330 \mathrm{k} \Omega$ |
| 10 | $3.3 \mathrm{k} \Omega$ | 10 | $1 \mathrm{M} \Omega$ |
| 10 | $10 \mathrm{k} \Omega$ | 10 | $10 \mathrm{M} \Omega$ |

All are carbon film-type resistors, $5 \%$ tolerance, $\frac{1}{3}$ or $\frac{1}{4}$ watt rating

## POTENTIOMETERS

| Quantity | Value |
| :---: | :---: |
| 2 | $10 \mathrm{k} \Omega$ |

Log. law carbon track, plastic shaft 6 mm (dia.) $\times 50 \mathrm{~mm}$ (approx.)

## CAPACITORS

Quantity
$\left.\begin{array}{l}\text { Value } \\ 1000 \mu \mathrm{~F} \\ 100 \mu \mathrm{~F} \\ 10 \mu \mathrm{~F} \\ 2.2 \mu \mathrm{~F}\end{array}\right\}$

Electrolytic type; 16 V or 25 V rating. Very large types (higher voltage ratings) are NOT suitable.
disc ceramic polystyrene polystyrene

Electrolytic and polystyrene capacitors should have axial leads. All wire leads should be about $1 \frac{1}{2}$ in long; short lead-out wires are NOT suitable.

DIODES

| Quantity | Type |  |
| :---: | :--- | :--- |
| 4 | 1N4148 | silicon signal diode |
| 1 | KV1236 | dual varicap diode |
| 3 | - | light emitting diode (l.e.d.) 5 mm , red |

## TRANSISTORS

Quantity

5

5

Type BC107B

BC179
high gain npn transistor (suitable alternatives: BC108B, BC108C, BC109) high gain pnp transistor (suitable alternatives: BC 177 B , BC178B or C)
Both types are encased in metal cans (TO18) and base connection is the centre lead.

## INTEGRATED CIRCUITS

Quantity 2 2 2

## SUNDRIES

Quantity
1
$2 \quad$ 100mm (4in) long
$3 \quad$ Three pieces, different colours, each 1 metre in length
1 One piece (of another colour) 2 metres in length for coil winding. Type
1 Basic Discrete Component Starter Pack
MOSFET op-amp 8-pin d.i.I. $\begin{array}{ll}\text { CA3140E } & \text { MOSFET op-amp 8-pin } \\ \text { 4069UB } & \text { hex inverter 14-pin d.i.l. }\end{array}$ dual 2 -input NOR gate 14-pin d.i.l. 4001 UB

## Description

Ferrite aerial rod $9.5 \mathrm{~mm}\left(\frac{3}{8} \mathrm{in}\right)$ dia.,

## Crystal earphone

12-way 2 -amp terminal block, polythene.
Connecting wire, plastic insulated, single strand ( 0.6 mm ):

Basic Integrated Circuit Starter Pack

## EBBO BREADBOARD MODULES

```
Quantity
``` 1


Fig. 1.4. Chart showing the voltage needed to drive 1 mA through resistances from 100 ohms to 100 kilohms.

\section*{RESISTANCE, VOLTAGE, CURRENT}

The resistance value can be described by saying how many volts are needed to drive a current through a resistance. If a current of one ampere (1A) is driven when 1 V is applied the resistance is 1 ohm.

In electronics we are usually concerned with much smaller currents and much higher resistances. If 1 V drives only 1 mA the resistance is 1,000 ohms \((1 \mathrm{k} \Omega)\). A \(1 \mathrm{k} \Omega\) resistance (commonly in speech "one kay") requires 1 V for every 1 mA that flows through it. If 9 V is applied to \(1 \mathrm{k} \Omega\) the current is 9 mA . More voltage, more current.

The chart (Fig. 1.4) shows how many volts per milliampere for different resistances. To find out how many volts are needed to drive 1 mA through a resistance you read off the voltage opposite the resistance value. Opposite 1 k , for instance, is 1 V . This means that 1 V will drive 1 mA through \(1 \mathrm{k} \Omega\). For \(5 \mathrm{k} \Omega\), on the other hand, 5 V is needed for 1 mA .

For currents other than 1 mA you do a bit of simple arithmetic. Example: What voltage is needed to drive 5 mA through \(1 \mathrm{k} \Omega\) ? The chart says that 1 V drives 1 mA so for 5 mA you need five times the voltage; that is, 5 V . Current is proportional to voltage. To be more precise:

Current \(=\) Voltage \(\div\) Resistance which is one way of stating Ohm's Law, the basic rule which governs all d.c. circuits and many a.c. ones.

\section*{SUPPLIERS OF KITS FOR TEACH-IN 84}

Please refer to advertisement on page stated.
Bi-Pak (page 680)
PO Box 6 .
Ware, Herts.
Greenweld Electronics (page 684)
443 Millbrook Road,
Southampton, SO1 OHX.
Magenta Electronics (page 620)
135 Hunter Street,
Burton-on-Trent, Staffs
DE14 2ST.
TK Electronics (page 682)
11 Boston Road,
London, W7 3SJ.

\section*{EXPERIMENT 1.1}

\section*{LIGHTING A LAMP}

Light-emitting diodes (l.e.d.s) must never be connected directly across a battery. If they are so connected they either do nothing or draw an enormous current which destroys them.

In practice, l.e.d.s are lit from a power supply which delivers more voltage than they need; the excess voltage is "dropped" by including a resistance (Fig. 1.5) which also limits the current.

Suppose we decide to run a red l.e.d at the very modest current of 1 mA , which is just about enough to make it glow dimly. Our battery is 9 V . The chart Fig. 1.4 shows that 9 V drives 1 mA through \(9 \mathrm{k} \Omega\). However, the l.e.d. absorbs about 2 V , so the resistance \(R\) needs only to absorb the remaining 7 V . The chart links 1 mA and 7 V to \(7 \mathrm{k} \Omega\), and this is the required value for \(\mathbf{R}\).

Here the circuit designer is up against a practical problem. Cheap resistors are not usually made to a \(7 \mathrm{k} \Omega\) resistance. The nearest standard values are \(6.8 \mathrm{k} \Omega\) and \(8.2 \mathrm{k} \Omega\). Clearly, \(6.8 \mathrm{k} \Omega\) will allow a little more than 1 mA to flow and \(8.2 \mathrm{k} \Omega\) a little less. Since the l.e.d. can safely pass much more than 1 mA (most l.e.d.s are run at around 10 mA ) we might opt for \(6.8 \mathrm{k} \Omega\) which is the closer standard value.

In fact, we are going to use \(10 \mathrm{k} \Omega\) which is a nice round value, which we'll need for later experiments. How much current will it pass in our circuit? Assuming that the l.e.d. still uses up 2 V , we are

\section*{METER ADAPTOR LEAD}

To use the multimeter with the EBBO Blocks, an adaptor lead is required. This is easily made by cutting a 3 -way section off from one 12 -way terminal block, then connecting a piece of connecting wire about 12 inches long to each outer terminal.

These two leads (preferably one red and one blue or black) should be twisted together to within 2 or 3 inches of the ends. Remove the plastic insulation from the ends, leaving one inch bare as "probes" which will fit the holes on the EBBO blocks.

The multimeter probes are secured to the outer terminals of this block, their colours (red and black) corresponding to the colours of the wire leads already fitted. See photo.
in effect left with a reduced battery voltage of 7 V . The chart shows that 10 V drives 1 mA through \(10 \mathrm{k} \Omega\). Commonsense says that 7 V will drive proportionately less current. Since 7 V is \(7 / 10\) of 10 V , the current will be \(7 / 10\) of 1 mA , which is 0.7 mA .

First make sure you connect your l.e.d. the right way round. Unlike resistors, which pass current equally well in either direction, l.e.d.s work only when the current flows in one particular direction.

A l.e.d. has a "current entry" lead, called the anode (a) and a "current exit" lead called the cathode (k). The anode must always face towards the positive terminal of the battery. This means that if you trace the circuit back from the anode against the direction of current flow you eventually arrive at battery positive.

The cathode lead-out of many l.e.d.s is indicated (none too clearly) by a small flat area near it on the plastic case. It is suggested that you mark the cathode black and the anode red, by slipping short lengths of plastic sleeving (stripped from connecting wire) over the appropriate lead-out wires.

\section*{COMPONENT BLOCK}

Before actually fitting components into the EBBO block read carefully the instruction booklet that comes with the EBBO kit.

Connect up a \(10 \mathrm{k} \Omega\) resistor and l.e.d. using the light blue Discrete Component Block from the EBBO kit. Fig. 1.6 shows the physical arrangement which corresponds with the circuit in Fig. 1.5.

On the diagrams in this Teach-In series, letters and figures with circles round them refer to the labelling of the points on the EBBO blocks. If only a letter is given it means you can use any point in that row. But remember that the discrete component block has two separate areas of terminal points, the lefthand area is not connected to the righthand area. Use one area only for this experiment, as shown in Fig. 1.7.

If component leads are dull and tarnished, brighten them up with a piece of Scotchbrite scouring material before inserting into the component block.

\section*{BATTERIES}

The EBBO kit contains a battery holder which takes a miniature 9 V pocket transistor battery. Rather than use the cheapest form (PP3) choose a calculator type battery which can deliver greater currents.

To use a PP3-type battery clip with the EBBO battery holder connect its red ( + ) lead to \(C(+)\) and its black one to \(C(-)\). You can now connect the battery to the block by plugging leads into the spare positions on the holder, \((+)\) to row A on the block and ( - ) to row G, see Fig. 1.7.

If all your connections and the polarity of the l.e.d. are correct the l.e.d. will glow dimly enough to be seen in a room that


Fig. 1.5. Theoretical circuit diagram for light-emitting diode (I.e.d.) D1 with currentlimiting resistance R. (Encircled letters and numbers refer to holes on component block.)

Fig. 1.7. The circuit of Fig. 1.5 assembled on the EBBO Block. The two left-hand leads go to the battery holder module. (See phofos)


The experiment set-up on the EBBO Block, in accordance with Fig. 1.5 and Fig. 1.7

\section*{CHECK YOUR PROGRESS}

Questions on Teach-In 84 Part 1
Q1.1 What resistance values and tolerances are indicated by these codes?
a. orange, orange, brown, brown
b. brown, black, yellow, gold
c. red, violet, orange, gold
d. green, blue, green, red
e. brown, black, gold, gold
f. yellow, violet, goid, gold

Q1.2. Resistance, voltage and current are related. Fill in the missing one:
a. \(10 \mathrm{~V}, 10 \mathrm{k} \Omega\), (?)
b. \(10 \mathrm{~V}, 10 \mathrm{~mA}\), (?)
c. \(5 \mathrm{~mA}, 2 \mathrm{k} \Omega\), (?)

Q1.3 A l.e.d. is to be run at 10 mA from a 12 V battery. If its running voltage is 2 V what resistance is required to set the current?

Answers next month
isn't brightly lit. Once you have got it working, you can brighten it up by replacing the \(10 \mathrm{k} \Omega\) with a \(1 \mathrm{k} \Omega\).

However, before doing so, set your test meter to 10 V d.c. and measure the voltages as indicated in Fig. 1.8. Note what they are and repeat using \(1 \mathrm{k} \Omega\) instead of \(10 \mathrm{k} \Omega\) for \(R\).

Do not attempt any more measurements. Leave your meter on the 10 V range and switch off the circuit by unplugging the positive battery connection.


Fig. 1.8. Showing where to connect a 10 V d.c. meter to measure three different voltages in the circuit.

Next month: Batteries and Resistors


BY A.P. DONLEAVY

To ADD another hurdle to the competitive aspect of model car racing, each car can be given a petrol tank which is emptied at a rate depending on the speed of the car. With an empty tank, the car comes to a stop until the tank is refilled, which takes a certain amount of time.

A "petrol gauge" gives the state of the tank and the "tank" is in fact, a capacitor. The fuel gauge consists of a l.e.d. which changes colour from green to red as the tank empties.

\section*{CIRCUIT DESCRIPTION}

The circuit diagram is shown in Fig. 1. There is an identical circuit for each track so only the circuit for one track is drawn. The numbers in brackets next to IC2 inputs give the corresponding i.c. pin con-
nections for the other channel and all duplicated component references are prefixed by 100 .

The "tank" is represented by C1, which is full when the capacitor is discharged. The charging up of this capacitor through R3 and IC 1, an adjustable current source, represents the emptying of the tank. The current to charge the capacitor is governed by IC 1 , which in turn depends on the series resistance of R2 and VR1. For a given value of R2 and VR1, the current remains constant even though the voltage across the terminals of IC1 changes as the capacitor charges.

The current flowing to Cl causes a volt drop across R3. This voltage is applied to the non-inverting ( +ve ) input of IC2a (Pin 3). IC2 consists of four operational amplifiers, all of which are used in this project.

The prototype Model Car Petrol Tank for a two-car race track.


The output (Pin 1) of op-amp IC2a is connected directly to the inverting input. This gives it unity gain, that is the output voltage follows the input voltage at the non-inverting terminal.

The output of this op-amp is connected via R4 to the base of TR1, the emitter of which is connected to the base of TR2. It is the collector TR2 which provides the drive current for the racing car. The available current to drive the car from TR2 will depend on the output voltage at Pin 1 of IC2a and the value of R4. The speed of the car depends on the current supplied by TR2.

\section*{OPERATION}

The operation of the unit is then as follows: the capacitor C 1 , initially discharged, is charged up by the current which is adjusted by the setting of VR1. R2 limits the maximum current. The higher the current, the higher the voltage drop across R3, thus the higher the voltage at the output of IC \(2 a\), and the higher the current from TR2, hence the faster the car travels.

The voltage at the \(\mathrm{v}+\) terminal of IC1 continually falls as C 1 charges until it comes to a point where the voltage at the \(\mathrm{v}+\) terminal is insufficient to allow the correct functioning of IC1 (about IV). At this point the current starts dropping, and also the voltage across R3. This in turn leads to a reducing current from TR 1 and TR2.

Another limiting factor is that unless the voltage at Pin 1 IC2a is at least equal to the two base-emitter voltage drops of TR1 and TR2 (about 1.4 V ), these transistors will not be turned on. Hence even though IC1 is functioning at 1.4 V , the output transistors would cease to pass any current, thus stopping the racing car.

In practice, what happens is that at maximum speed the voltage at Pin 3 is about 3 V . This maximum speed can be maintained until the voltage at ICl \(\mathrm{v}+\) has dropped to about 4 V , after which the current through R3 starts dropping, leading to a continuously reducing current from TR2, until it eventually becomes zero.

To refill the tank, S2 switches R1 to C 1 and discharges it at a rate determined by R1. During this refill period, no current flows through R3, so Pin 3 is at ground potential and the car is stopped.

In the normal operation, S1 must be closed, otherwise IC1 sees an infinite resistance between Pin's R and V-, and passes negligible current. S1 is included to allow a sudden stop without needing to adjust the potentiometer VR1.

The petrol consumption will continue to occur whether or not the car is on the track, so the inclusion of S 1 is a petrol saving device to allow for the frequently occurring situation of the cars coming off the tracks.

\section*{FUEL GAUGE}

The "fuel gauge" is contained in the circuitry of IC2b, TR3, TR4 and D1. D1 contains two l.e.d.s, one red and one green in a single package, connected by a common anode. By varying the currents individually through the l.e.d.s it is possible to change the colour of the display from red to orange (or yellow) to green.

The state of the tank is represented by the voltage across the terminals of the
capacitor C 1 . The voltage of the negative terminal of Cl is measured by op-amp IC2b, also connected as a unity gains voltage follower. In the tank's full state, the voltage at Pin 5 IC2b will be at almost supply potential. The output (Pin 7) will also be at this potential, thus TR4 will be turned on, and the green diode in D1 will be alight. R6 controls the maximum current passed by TR4.

TR3 will be turned off, so the overall display colour will be green. As the tank empties, the voltage at Pin 5 and hence Pin 7 of IC 2 b will drop. At some point current will start to flow through TR3, thus turning the red half of D1. As the voltage continues to decrease, the red will get brighter and the green will become extinguished.

When both l.e.d.s are alight, the display will be various shades of orange depending on the currents through each. D2, a green l.e.d. is included to allow the base voltage of TR4 to rise to about 3V before it starts to pass any current. The high forward voltage drop of the diode achieves this. Thus the display colour gives an indication of the charge on the capacitor, that is the state of the tank.

D3 and C2 are intended to protect the supply to the circuitry from the noise generated by the cars and to smooth the rectified AC mains supply. More will be said about the power supply later.

\section*{PRACTICAL ASPECTS}

The circuit and component values in Fig. 1 were designed round a 6 V racing
track. To optimise the performance of the unit for any particular racetrack the constructor must expect to do some experimentation, and if necessary modify some of the component values.

R4 sets the maximum current available from TR2 and this must be chosen to suit the maximum current required by the car. Measure the current for maximum speed using the track's own controller, then by trial and error, find the value of R4, which will give this same output current.

It may well be found that a different value will be required for each track, since the combined gains of TR1 and TR2 will probably be different for each channel. The value of \(R 4\) will be between 100 kilohms and 330 kilohms.

For the gauge circuitry, the values of R5 and R6 determine the calibration of the fuel gauge. For constructors who wish to be fairly exact about this, choose R6 so that the green l.e.d. gives negligible light at about 3 V on Pin 7 of IC 2 b .

\section*{POWER SUPPLY}

The circuitry, except for the output transistors, is designed to operate on a supply voltage between 9 V and 12 V . It could operate on less, but this would reduce the tank capacity. The current delivered by the output transistor TR2 will be approximately the same irrespective of supply voltage.

However, it will only supply the maximum current to the car if the supply voltage is equal to, or greater than, the normal operating voltage for the car. It is

Fig. 1. Circuit diagram of the Model Car Petrol Tank. Only one channel is shown and the hand control, petrol tank and fuel gauge sections are repeated for car 2. Switch S2 is biased towards the RUN position as shown.

envisaged that the existing power supply of the racetrack is to be used and the various different power supplies are as follows:
\(1 \quad 12 \mathrm{~V}\) d.c. rectified mains supply-this is the simplest case, and may be connected directly to the unit as shown in the circuit diagram.
29 V battery or rectified d.c. from a mains unit-again may be connected directly to the circuit. If the tank runs dry too quickly, this can be compensated for by either increasing the value of C 1 , or by reducing the value of \(R 4\) which in turn will require the setting of VRI to be altered to compensate for the speed. This reduces the charging rate of the capacitor but leaves open the possibility of overloading the car.
3 Battery or d.c. rectified supply from a mains unit of less than 9 V --in this case it may be necessary to provide a separate power supply for the circuitry of IC2, TR1, TR3 and TR4 by a 9 V battery or a calculator adaptor, connected by a common ground to the other supply. Fig. 2 shows the circuit for this.


Fig. 2. Power supply modification for low voltage tracks. Note common OV.


Fig. 3. Additional override switch.

\section*{COMPONENTS}

Almost any small signal \(n p n\) transistors could be used for TR1 and TR4 and similarly other small signal pnp transistors for TR3. For TR2, other high current npn types could also be used. A heatsink must be fitted.

The prototype used 2.5 mm banana sockets and plugs to connect track to control box. A 2.5 mm jack plug and socket was used to connect the power supply to the control box.

If at any time it is required to override the tank circuitry and run the car continuously, the addition of an extra switch to connect the \(v+\) input of IC1 to the positive supply will achieve this. Fig. 3 shows the circuit for this. The prototype did not have this facility.

COMPONENTS. .is

Resistors
R1.101 \(22 \mathrm{k} \Omega\) (2 off)
R2,102 \(2.2 \mathrm{k} \Omega\) (2 off)
R3, \(103100 \mathrm{k} \Omega\) (2 off)
R4,104 see text
R5,105 \(22 \mathrm{k} \Omega\) (2 off)
R6,106 68k (2 off)
All \(\frac{1}{4} \mathrm{~W}\) carbon \(\pm 5 \%\)

\section*{Capacitors}

C1, \(101 \quad 470 \mu \mathrm{~F} 16 \mathrm{~V}\) elect. (2 off)
C2 \(220 \mu \mathrm{~F} 16 \mathrm{~V}\) elect.
Semiconductors
\begin{tabular}{ll} 
D1,101 & tri-colour common \\
& anode l.e.d. (2 off) \\
D2,102 & TIL211 green l.e.d. (2 off) \\
D3 & 1N4001 silicon \\
TR1,101 & BC108 npn silicon (2 off) \\
TR2,102 & TIP41 npn silicon (2 off) \\
TR3,103 & BC179 pnp silicon (2 off) \\
TR4,104 & BC109 npn silicon (2 off) \\
IC1,101 & LM334 adjustable \\
& current source (2 off) \\
IC2 & LM324 quad op-amp
\end{tabular}

\section*{Potentiometers}

VR1.101 \(4.7 \mathrm{k} \Omega\) carbon lin (2 off)

Miscellaneous
S1,101 push-to-make momentary action (2 off)
S2,102 s.p.d.t. biased toggle (2 off)
SK1 \(\quad 2.5 \mathrm{~mm}\) mono jack
SK2 \(\quad 2.5 \mathrm{~mm}\) banana red SK3 \(\quad 2.5 \mathrm{~mm}\) banana black SK \(102 \quad 2.5 \mathrm{~mm}\) banana green 0.1 in matrix stripboard, 29 strips by 33 holes; console type case, \(100 \times 160 \times 40\) (front) \(\times 60 \mathrm{~mm}\) (rear); hand-held case, \(85 \times 55 \times\) 35 mm ( 2 off for hand controls); 2.5 mm banana plug ( 3 off); control knob (2 off); l.e.d. clip (2 off); 14-pin i.c. holder; TO220 type heatsink ( 2 off); \(7 / 0.2 \mathrm{~mm}\) wire; mounting hardware; Veropins.

\section*{CIRCUIT BOARD}

The main part of the circuit is constructed on a piece of \(0 \cdot 1\) in matrix stripboard, the layout of which is shown in Fig. 4. The underside view shows all track-breaks. When assembled, the board is secured to the lid of the case using four screws with short spacers to prevent the solder joints shorting on the lid.

The two l.e.d.s, D2 and D102, are soldered directly to the board as their
function in the circuit does not require them to be seen. Note that R4 (and R104) is soldered to a pair of Veropins to facilitate the select-on-test procedure.

Transistors TR2 and TR102 are fitted with heatsinks as these components get quite warm. Any heatsink suited to a T0220 type package will suffice. All usual precautions to be taken when inserting semiconductor and polarised capacitors and all wired connections to be taken to Veropins.

View inside the console case showing the heatsinks on the power transistors. Note that this unit was constructed on Vero \(V-Q\) board and therefore the layout differs slightly from that shown in Fig. 4.


\section*{MODEL CAR 'PETROL TANK'}


Fig. 4. Stripboard layout for both channels of the Model Car Petrol Tank. Note that the TO-220 type heatsinks for TR2 and TR102 are not shown but must be securely fixed to the metal tabs of these transistors.


Fig. 5. Wiring diagram of the hand control unit as used on the prototype model. Some form of cable restraint should be used to prevent the wires from pulling out.

\section*{HAND CONTROLS}

The GO switch S1, and the SPEED control VR1, are mounted into a hand-held box (one for each channel) connected to the control box by a suitable length of twin cored wire. The prototype model uses two small square cases, \(85 \times 55 \times\) 35 mm , but the constructor may wish to use a purpose built hand-held box such as the Vero type 202-21026G with a thumb wheel potentiometer of the type available from Maplin (part number BW06G Edge Control with part number BW09K Edge Knob). See Fig. 5 for layout.

The unit is housed in a console case with a \(157 \times 92 \mathrm{~mm}\) front panel. The FUEL GAUGE l.e.d.s (D1 and D101) and the FILL/RUN switches (S2 and S102) are mounted on the front panel either side of the circuit board assembly.

The 2.5 mm banana sockets to the track and the power input jack socket are mounted in the back and side wall of the case, respectively. Two small holes allow for the wires to the hand controls and some form of strain relief must be used to prevent these wires pulling out. A ratchet type cable tie is ideal.

Some additional ventilation holes are required in the back of the case as the power transistors generate some heat.

\section*{TROUBLE SHOOTING}

The operation of the unit can be tested without connecting it to the track. If the
unit appears completely dead, then check the polarity of the incoming power supply connections. Also check that diode D1 is correctly inserted.
If the operation of VR1 does not regulate the speed of the car, or the car does not work at all, then check the voltage at the base of TR 1 ; this should be about 1.4 V . Check also that the voltage at Pin 1 of IC2 (or Pin 14 for the other track) varies with the setting of VR1 (S1 closed) with the tank at least partially full. If it does, then either TR1 or TR2 must be suspect.

If the voltage at Pin 1 does not vary with VR 1 (S1 closed) check to see if the voltage at Pin 3 (or 12) varies with VR1. If it does then IC2 is faulty, assuming it has been correctly wired.

To investigate any malfunctions of the fuel gauge, check that the voltage at Pin 5 (or 10) follows that of the negative terminal of C1, thus eliminating a wiring error. Then check that the output Pin 7 (or 8) also follows this voltage. If not then IC2b may be faulty, or else the Pin's 6 and 7 (or 8 and 9) are incorrectly wired, possibly shorted to another track. If IC \(2 b\) is functioning correctly, then any fault in the fuel gauge operation will lie either in D1 being faulty or incorrectly wired.

\section*{IN USE}

The existing speed controllers are not needed, as the speed is now controlled by

Rear view of the main control unit. The additional ventilation holes can be seen.



VR1. The hand control is held in one hand while depressing switch \(S 1\), and the speed regulated with VR1 using the other hand.

Should a collision occur or a car leave the track, the petrol consumption will automatically stop if \(S 1\) is released while rectifying the situation on the track. When the fuel tank indicator shows red, the tank may be filled by switching S2 to FILL, until the l.e.d. is green, or for how long the driver considers necessary to finish the race.

If the constructor finds the tank empties too quickly or too slowly, then increase or decrease the value of C 1 . If the tank refilling time is too long or too short, increase or decrease the value of R1, respectively.

\section*{PLFASE TAKENOTE}

\section*{Car Intruder Alarm (August 1983)}

The circuit diagram (Fig. 1, page 520) shows the positive side of the car battery connected via the ignition switch. This has been drawn incorrectly and no connection to the ignition is required.

The circuit diagram also shows the transistor TR1 connected straight to the car horn. TR1 is insufficient to drive the car horn and an automobile relay should be placed in the circuit to switch the horn.
1984 SCHOOLS Electronic Design Award COMP ETITION Build bor the Future aud Reap the Reurards

the twelve finalists WiLl share OVER £3000 AND PARTICIPATE IN AN ALL-EXPENSES-PAID TWO-DAY VISIT TO LONDON DURING JUNE 1984

For the third year running Mullard Ltd.-the largest electronic components company in the UK-and Everyday Electronics join forces to present a rewarding challenge to Secondary Schools within the United Kingdom.

As distinct from the two previous contests, this year the conditions have been somewhat broadened so that entries can relate to electronic equipment having practical use anywhere within the school, or at any external event in which the school participates.

FIRST PRIZE The Sedac Trophy and \(£ 300\)
SECOND PRIZE £200
Plus a selection of components valued at \(£ 200\)
THIRD PRIZE £100
NINE RUNNERS-UP. A selection of components valued at \(£ 100\).
In addition, all twelve finalists will receive a certificate and one year's subscription to EvERYDAY ELECTRONICS-and will enjoy an all-expenses-paid two-day visit to London dúring June 1984.

Science teachers of Secondary Schools are invited to apply for a Registration Form which contains full details of this competition.

Write to: SEDAC Schools Competition, Room 2130, King's Reach Tower, Stamford Street, London SE1 9LS.

Secondary School Pupils-make sure your school accepts this challenge and enters this contest. So bring this announcement to the attention of your science teacher or the head of your school.

Closing date for Registration:
15 November 1983
Closing date for submission of Papers:
31 January 1984

\title{
TEMPERATURE TRANSDUCER AMPLIFIER
}

\author{
BY A. A. CHANERLEY B.Sc.M.Sc.
}


THIS article describes the construction and use of a simple Temperature Transducer Amplifier that provides an output voltage directly proportional to the temperature of the sensor in Celsius. The temperature range capability of the sensor is from -55 to +150 degrees Celsius.

This unit has been specifically designed for use in the laboratory where it is to be connected to a computer via an A-to-D converter to automatically read, store, calculate and graphically display the results from heat experiments in the physics and chemistry laboratories. See Computer Aided Experiments, page 674.

However, it is not restricted to such applications and could be connected to a voltmeter (suitably labelled) to give the temperature reading of say a person's body temperature, room or greenhouse temperature and many other environments, on a Fahrenheit, Celsius or Kelvin scale.


\section*{TEMPERATURE SENSOR}

The circuitry has been designed around the 590 kH temperature transducer, IC 1 .

The temperature transducer itself is a high impedance current source producing a current proportional to the absolute temperature, \(1 \mu \mathrm{~A}\) per Kelvin rise in temperature. The temperature range of this sensor is from -55 to 150 degrees Celsius maximum, being linear to within \(0.8^{\circ} \mathrm{C}\) over this range.

The fact that this is a high impedance current source makes the device ideal for remote temperature sensing, since it is virtually insensitive to voltage drops over long lines.

The simplest configuration for the sensor is in series with a resistor \(R\), as shown in Fig. 1. The \(1 \mu \mathrm{~A}\) per Kelvin is then transformed to a p.d. per Kelvin across the resistor \(R\). So that if \(R=10\) kilohms, the p.d. across it is 0.01 volts per Kelvin. Therefore, on the Kelvin temperature scale, 273 Kelvin produces 2.73 volts
across the resistor; 273 Kelvin is equivalent to 0 degree C on the Celsius scale, that is, at the temperature of melting ice.
Since most people are more at home with the Celsius scale, it is better to produce an output per degree Celsius rather than per Kelvin. This is done using the arrangement of poten tiometers shown in Fig. 2. VR1 provides, after adjustment, the voltage needed to convert from Kelvin to degree Celsius, that is, -2.73 volts, while VR2 zeros the transducer output at the "ice point", that is, 0 degrees C , the temperature of melting ice. The complete arrangement produces an output of 0.01 volts per degree C suitable for feeding into the noninverting input of an op-amp.

\section*{THE AMPLIFIER}

The complete circuit diagram of the amplifier is shown in Fig. 3.

IC2, a 741 op -amp is configured in the non-

Fig. 3. Complete circuit diagram for the Temperature Transducer Amplifier.

inverting mode with feedback to give a closed loop gain of:
\[
\mathrm{A}_{\mathrm{vf}}=(\mathrm{R} 3+\mathrm{R} 2) / \mathrm{R} 2
\]

Substituting in the circuit values gives:
\[
A_{v f}=(16 \mathrm{k} \Omega+2 \cdot 2 \mathrm{k} \Omega) / 2 \cdot 2 \mathrm{k} \Omega=8
\]

The amplifier in this configuration has a very high input impedance given by:
\[
\mathrm{Z}_{\mathrm{inf}}=\mathrm{A}_{\mathrm{v}} \mathrm{BZ}_{\mathrm{in}}
\]
where \(A_{\mathbf{v}}=\) open loop gain \(\bumpeq 200,000\)
\(B=\) feedback fraction \(=R 2 /(R 2+R 3)\)
\(\mathrm{Z}_{\mathrm{in}}=\) input impedance without feedback \(=2 \times 10^{6}\) ohms
\(\mathrm{Z}_{\mathrm{inf}}=\) input impedance with feedback.

Therefore substituting these values gives:
\(\mathrm{Z}_{\mathrm{inf}}=2 \times 10^{5} \times(2.2 \mathrm{k} \Omega / 18.2 \mathrm{k} \Omega) \times\) \(2 \times 10^{6} \bumpeq 10^{11}\) ohms, which is quite a high order of magnitude.

The values of R2 and R3 produce a closed loop gain of \(\times 8\), thus an input of 0.01 volts per degree C from the transducer circuit will produce an output of 0.08 volts per degree Celsius at the amplifier, so that 100 degrees C will produce 8 volts.

\section*{COMPONENTS}

\section*{Resistors \\ \begin{tabular}{ll} 
R1 \(\quad 10 \mathrm{k} \Omega\) \\
R2 & \(2.2 \mathrm{k} \Omega\) \\
R3 & \(16 \mathrm{k} \Omega\) \\
All \(\frac{1}{4} \mathrm{~W}\) & carbon \(\pm 5 \%\) See
\end{tabular} \\ Potentiometers \\ VR1 \(5 \mathrm{k} \Omega\) linear VR2 \(100 \mathrm{k} \Omega\) linear \\ }

\section*{Semiconductors}
\begin{tabular}{ll} 
IC1 & 590kH temperature \\
IC2 & transducer \\
\(741 \mathrm{op}-\mathrm{amp}\)
\end{tabular}

\section*{Miscellaneous}

B1.2 9V type PP3 (2 off)
S1 miniature d.p.d.t. toggle
SK1 4mm panel mounting socket, red
SK2,3 4 mm panel mounting socket, black (2 off)
SK4 4 mm panel mounting socket, green
Stripboard: 0.1 inch matrix size 18 strips \(\times 24\) holes; 8 -pin d.i.l. socket; PP3 battery clips (2 sets); control knobs to suit VR1 and VR2 ( 2 off); 6BA fixings ( 2 sets); stranded connecting wire; case, Vero General Purpose Plastic Box, type 202-21390D size \(120 \times 80\) \(\times 35 \mathrm{~mm}\) or similar.

Approx. cost Guidance only

The op-amp uses a \(\pm 9\) volt split supply, provided by two PP3 batteries, which are switched in and out of circuit by S1, a double pole switch.

\section*{ASSEMBLY}

The construction of the unit is relatively straightforward with the amplifier circuitry assembled on a piece of 0.1 inch matrix stripboard, 18 strips \(\times 24\) holes. Full assembly and wiring details are contained in Fig. 4. The links are made with plastic covered single core wire. A cut-out in one corner of the board is required to accommodate the batteries in the specified case.

Begin assembly by cutting the board to shape, drilling the two fixing holes and then making the necessary breaks in the copper tracks on the underside of the board. There are 12 in all. A d.i.l. socket was used to hold IC2 in the prototype and constructors are advised to do likewise.

Position and solder the components as shown. Attach sufficient lengths of flying lead (stranded type) to comfortably reach the case mounted components when fitted.

Prepare the case lid and base and secure the appropriate components as shown in Fig. 4. Use red and black for the input 4 mm sockets, and green and black for the output sockets, black in each case being the ground ( 0 V ) connection.



Fig. 5. Details of the construction of the sensor "probe". The completed sensor is shown in the photograph above.


The completed temperature amplifier. The supply switches shown in this prototype have been replaced by a single switch.

Fit the completed board in the case using 6BA fixings and then wire up as shown. When complete, insert IC2, paying attention to the orientation of this device to complete the construction. Attach labels to identify the panel controls and sockets.

\section*{SENSOR/CALIBRATION}

The sensor itself is shown in Fig. 5 and in the photograph. Suitable lengths of plastic covered wire are soldered to the ( + ) and ( - ) pins of the sensor and eventually terminated by \(2 \times 4 \mathrm{~mm}\) plugs, red and black respectively, which plug into the input sockets of the amplifier of the corresponding colour.

The temperature transducer is mounted on the end of a glass tube, and both ends are sealed with epoxy resin for water-proofing, so that the can-head protrudes from above the resin. Once it dries, the apparatus is ready for calibration.

Immerse the transducer in a mixture of ice and water ( 0 degrees \(C\) ) and plug the red and black 4 mm plugs with the correspondingly coloured sockets on the Amplifier box. Connect the amplifier output and ground to a d.c. voltmeter ( 10 V full scale) and switch on. Using VR1, zero the output. If the needle doesn't zero completely, then adjust VR2 until the output is 0 volts. The apparatus is now ready for use.


\section*{HOW TO DESIGN AND MAKE YOUR OWN PCB}

\author{
Author \\ Price \\ Size \\ Publisher \\ ISBN \\ R. A. Penfold \\ \(£ 1.95\) Limp edition \\ \(178 \times 110 \mathrm{~mm} .66\) pages \\ Bernard Babani Ltd. \\ 0-85934-096-1
}

FOR a new-comer to electronics as well as to the experienced hobbyist, a need will eventually arise to design and make your own boards.

The book, How To Design And Make Your Own PCB contains three chapters starting with Simple PCB Production, which describes equipment and procedures for producing printed circuit boards. The remaining two chapters are Photographic Methods and Designing and each chapter is easily explained with diagrams to support the text.

Overall, a very useful book for anyone who is contemplating constructing their own boards and the book will also provide some very useful hints for the experienced constructor.
R.A:H

\section*{ELECTRONICS SERVICING PART 2 CORE STUDIES}
\begin{tabular}{ll} 
Author & R. Lewis \\
Price & £6.95 Limp edition \\
Size & \(234 \times 157 \mathrm{~mm} .142\) \\
Publisher & Macmes \\
Pallan Press \\
ISBN & \(0.333-30594-9\)
\end{tabular}

THE approach to the theory is strictly non-mathematical and physical explanations have been used throughout.
The book contains eight chapters starting with LCR Circuits, a chapter which considers the effects of applying direct and alternating voltages to circuits containing one or more of the electronic components, resistors, inductors and capacitors.

The remaining chapters are Transformers, Semiconductor Diodes and Applications, Transistors and other Semiconductor Devices, Voltage Amplifiers, Waveform Generation and Shaping, The Cathode Ray Tube and Simple Mechanisms and finally Power Supplies.

Each of the eight chapters is easily explained with diagrams to support the text and finally there is a set of self-test questions and answers at the back of the book. A good book which will help anybody studying electronics.

\section*{Books in Brief}
R.A.H

The Video Camera Handbook by Peter Lanzendorf (David \& Charles). £9.95. Using simple terms, this book explains the systems available for home video taping, the right tripods, lenses and filters. How to achieve correct lighting, how to make sound recordings and produce script and screenplay are also covered.

\section*{Development News}

Through my post at home I get literally dozens of press releases a day, covering audio, electronics and video. My last postman had to retire with a bad back.

Obviously it's impossible to write about everything I receive. So I read it all, make background phone calls for more information if something looks especially interesting, and then file it away for the future using a Sinclair Spectrum to cross reference related items.

For instance, by making a note of what new integrated circuits are becoming available for designers you can get a clear indication of what new gadgetry is likely to come on the market in a year or so's time. But if you think that being a journalist is just about re-writing press releases, think again.

Last week, Mullard, the British company owned by Philips of Eindhoven, sent me a press release on a new i.c. that integrates the entire circuitry for an \(\mathrm{f} . \mathrm{m}\). radio receiver, from aerial input to audio output. This, said the press release, will let designers build an f.m. radio into a pencil, wristwatch or keyring.

It uses a non-standard intermediate frequency. Usually the i.f. for an f.m. radio is 10.7 MHz , but in the Mullard chip it is 70 kHz . Curiosity aroused, I phoned Mullard to ask how they had overcome distortion problems, and how a low i.f. helps them make the chips smaller

A week, and several phone calls later, ! am still no nearer an answer. Believe it or not, there's no one in Mullard UK who can explain what's written in the company's press release. l've now been promised a call from an expert in Eindhoven who may or may not be able to help.

\section*{Tele News}

Another release is a clear pointer to a new generation of low-cost gadgetry that couples to your telephone and is triggered by an incoming ring. Texas Instruments are now selling chips that are powered by the incoming ring signal to produce a regulated 5 -volt output which connects the phone line to a microprocessor or system that handles voice or data signals.

The chip is ideal for a telephone answering machine, a data modern or facsimile machine. Cost, if you buy in hundreds, is just \(£ 1.40\) each.

Talking of telephones, I've just received news from British Telecom, who proudly announce that they have created 300 new jobs. How? By setting up three Telcare centres. Telcare stands for Telecom Customer Attitude Research, and the centre's sole job is to phone up Telecom subscribers and ask what they think of British Telecom.

Personally, I would have thought that, with so many people already telling BT what they think for free, in letters to the press and to \(B T\), that Telcare is an expensive luxury that BT subscribers can ill afford.

Finally, my favourite press release of the week; a blank sheet of paper with just the PR company's name and address and the words "Press Release" printed at the top.

Perhaps they've just got nothing to say this week.

\section*{Technology Cabaret}

You don't normally associate nightclubs with high technology, especially a Parisian nightclub. But 1 found out differently when I ended up at the Lido, probably the most famous of all Paris nightclubs.

Please understand, I wasri't paying the bill! With drinks at around \(£ 10\) a time on top of the meal, I'm not recommending that anybody else spends their hard-earned money that way either.

The technology at the Lido helps explain the high cost and raises an interesting point on the future of live entertainment. In addition to a fifty strong dance group, and a string of conjurers, jugglers and balancing acts, there's a full orchestra and singers out of sight in the gallery. They keep in touch with the stage through closed circuit television and headphones.

Perhaps inevitably the opening routine show relies on lasers, with several raw beams split into fantail arrays by diffraction gratings. The whole cabaret floor either rises up as a stage, or drops down and slides away to reveal an enormous pit.

Up from this pit, with a throbbing of hydraulics, rises either a swimming pool or a genuine ice-skating rink in a refrigerated tank. At the back of the stage they unveil a miniature Niagara Falls, that pumps real water by the thousands of gallons.

In one dance routine, jungle witchdoctors with scuba gear hidden under their masks, carry flaming torches into the swimming pool. They then stay underwater for the rest of the act with the flaming torches held aloft out of the water. Other fireworks are lit underwater. Don't forget that a firework doesn't need oxygen to burn, it's provided by the chlorate or nitrate in the firework mix.

The star comedian is obviously an ex pick-pocket, whose act involves stealing watches, wallets, socks, ties and braces from the audience without their knowing

His star trick was to sit a dozen or so members of the audience on chairs on the stage, and then give each of them a mild electric shock through the chair seat-l'm still trying to work out how that was done.

\section*{Bad Taste}

High-spot of the evening and pinnacle of bad taste was a re-enactment of the Hindenburg disaster. Half the stage was turned into a mock-up of life on board the airship, with back projected film of the transAtlantic flight from Germany to America.

Then there's a switch to Lakehurst, New Jersey, where the ground crew try to hold the dirigible down by ropes. True to history, it bursts into real flames. Four Lido performers, clad in protective clothing, and drenched in burning paraffin, leap out of the Hindenberg cabin and writhe around on stage, burning to death in distressingly realistic fashion.

We the public have now grown so used to seeing dramatic stunts and spectacles on film and on television, that the only way a live show can compete is to mount spectacles of this kind. But the cost of mounting such a show, with all the necessary technology and as many technicians backstage as performers on stage, is so high that most of us can't afford to go and watch it!

\section*{Post Script}

I finally obtained some technical information on the new Mullard chip, by talking to Philips in Eindhoven. It's a clever design. Normally the i.f. for \(\mathrm{f} . \mathrm{m}\). radio is 10.7 MHz , and at that frequency you can't possibly integrate the necessary coils and capacitors in a chip. But at the low i.f. frequency of 70 kHz you can use \(R C\) filters.

Unfortunately, the low i.f. makes it impossible to achieve the \(\pm 75 \mathrm{kHz}\) deviation normal for f.m. so the designer, a Mr. Kasperkovitz, devised a feedback circuit to keep disfortion down to acceptable limits even at an i.f. deviation of only \(\pm 15 \mathrm{kHz}\).

\section*{Digital War}

The hi fi world is now in the throes of a civil war over digital audio. One faction says that Compact Disc (the laser-read digital audio disc developed by Philips with the subsequent help of Sony), is the best thing to happen to audio since the microgroove LP and stereo.

The more level headed among them admit that the system is suffering from teething problems. Players and discs aren't yet perfect. But who expects any new technology to be perfect from the day it is launched?

The other faction vehemently argues that Compact Disc is inherently imperfect because the coding standards adopted can't ever produce a sound as good as you get from an analogue disc. Even when the players and the discs are working perfectly, and recordings are tailor-made to the new medium, it will still be inadequate.

For the record, if you will pardon the unin tentional pun, I firmly believe that the system
isn't yet right, but it is potentially better than anything an analogue disc can deliver. The best argument I can muster in favour of this, other than gut feeling, is a comparison with the digital audio systems used by the BBC for at least ten years to transmit f.m. stereo radio sound around the country, between v.h.f. transmitters.

Whereas Compact Disc audio is sampled at \(44 \cdot 1 \mathrm{kHz}\), and coded into 16 -bit words, the BBC has for ten years been sampling at 32 kHz and coding into 13 -bit words. So for ten years almost every f.m. stereo radio programme has been piped through a digital chain that is far inferior to Compact Disc.

I wonder why the people who are now complaining about what they call the nasty digital sound of Compact Disc, haven't been complaining over the last ten years about what should be the even more nasty digital sound of BBC v.h.f. radio?-Any comments to the Ed.

\title{
MICROCOMPUTER IITERFACIHG TECHNIQUES INCLUDING MANY USEFUL CONSTRUCTIONAL PROJECTS PART FOUR: DIGITAL-TO-ANALOGUE CONVERSION
}

\section*{BY J. ADA MS B.Sc. M.\$c. \& G. M.FEATHER B. Sc}

Adigital-to-analogue converter (DAC) produces an analogue output that is proportional to a binary code presented at its data inputs.

The reader should refer to Fig. 4.1 which shows a theoretical 2-bit digital-toanalogue converter.

The incoming 2 -bit logic signal is applied to electronic switches, \(S 0\) and \(S 1\), which close on receipt of a logic 1. The closure of a switch connects voltage sources V1 and V2 to the input of an operational amplifier which sums them and thus presents an analogue voltage at its output which will depend upon the binary signal presented to the DAC input.

The truth table for this simple arrangement is given in Table 4.1.

Table 4.1
\begin{tabular}{|c|c|c|}
\hline Digital & Inputs & \multirow{2}{*}{ Analogue Output } \\
\hline D 1 & \(\mathrm{D} \emptyset\) & \\
\hline\(\emptyset\) & \(\emptyset\) & 0 V \\
\(\emptyset\) & 1 & 1 V \\
1 & \(\emptyset\) & 2 V \\
1 & 1 & 3 V \\
\hline
\end{tabular}

The relationship between the digital input and the resulting analogue output voltage can be represented graphically, as shown in Fig. 4.2. In this simple arrangement the smallest increment of the analogue output voltage that can be produced is 1.0 V and this level of resolu-


Fig. 4.1. Theoretical representation for a 2-bit digital-to-analogue converter (DAC). S \(\emptyset\) and S1 are electronic switches operated by logic signals on the data lines.
tion is clearly not adequate for most purposes.

As with ADCs, resolution is a function of the number of bits available, so a DAC offering 8 -bit conversion will produce an

Fig. 4.3. Circuit of a DAC which employs an R-2R resistor network arrangement.


Fig. 4.4. Block schematic diagram for the DACO800 8-bit converter device used in the Digital-to-Analogue Converter.

The DAC0800 is an 8-bit converter and the digital signals applied to the electronic input switches S \(0-\mathrm{S} 7\) of the DAC are derived from the microcomputer user port.

The R-2R ladder network permits an output current to flow according to the combination of electronic switches closed. This is determined by the logic
levels of the eight signal lines from the user port which operate the input switches of the DAC. The overall magnitude of the output current is dependent upon the currents flowing in the network which in turn depends upon the reference current flowing. This is controlled by \(+V_{\text {REF }}\) and \(-V_{\text {REF }}\) and is typically in the order of 2 mA .

The DAC0800 does not posses an internal reference as does, for example, the Ferranti ZN428 and this allows rather more flexibility. A simple Zener diode circuit or some other voltage stabiliser can be used to provide \(+\mathrm{V}_{\mathrm{REF}} ;-\mathrm{V}_{\mathrm{REF}}\) is often grounded.

Two analogue current outputs are provided on the DAC. These are com-

Fig. 4.5. Complete circuit diagram of a Digital-to-Analogue Converter.


\(\frac{\text { Fig. }}{\text { louT. }}\) 4.6. Relationship between \(I_{\text {OUT }}\) and
plementary currents, that is, as one increases there is a corresponding decrease with the other; this is depicted in Fig. 4.6.

The DAC0800 device performs the conversion process very rapidly, 100 ns settling time being typical. Input switching levels are TTL compatible, although they are easily adjustable to other logic levels. The device will function satisfactorily with supply rail voltages between \(\pm 4 \cdot 5 \mathrm{~V}\) to \(\pm 18 \mathrm{~V}\).
are used for this purpose and, for convenience, an 8 -way commoned s.i.l. (single-in-line) resistor network package was employed.

Logic conditions on the input lines are indicated by l.e.d.s D1 to D8, the octal Darlington driver IC3 being used to control these l.e.d.s. This section of the circuitry is optional and may be omitted if desired.

IC1 is a 78 L 05 A voltage regulator which provides both \(+\mathrm{V}_{\text {REF }}(+5 \mathrm{~V})\) and the logic 1 levels when manual switching is used. The DAC0800 positive and negative supplies are derived from the +9 V and -9 V supplies to the board. PP9 batteries are quite adequate for the supply.

Logic threshold levels on IC2 are set by pin 1, which is grounded and gives TTL level operation of the digital input lines. Pin 16 allows for some compensation of the internal reference amplifier and this is provided for by the inclusion of C3 between pin 16 and \(-V_{C C}(-9 \mathrm{~V})\). The positive \(\mathrm{V}_{\text {REF }}\) is, as previously mentioned,
set at +5 V , whilst \(-\mathrm{V}_{\text {REF }}\) is taken to ground by R 10 , a \(4.7 \mathrm{k} \Omega\) resistor.

As with analogue-to-digital converters, DACs can be connected for either unipolar or bipolar operation. The DAC board utilises the bipolar technique which allows the output from the board to go both positive and negative with respect to ground. Bipolar operation is achieved by applying an offset voltage to the output of the DAC. This voltage is adjusted (by VR1) until the output is zero for a digital input of half full range (that is, logic levels 10000000 at the input). Positive and negative full scale outputs are not quite symmetrical about 0 V since the 8 -bit binary signal (representing 255 equal voltage steps) is not divided into two equal sections about 10000000 .

IC4 is a conventional non-inverting operational amplifier circuit which buffers the output from the DAC and provides a small voltage gain.

Table 4:2 shows the relationship between digital signals and the analogue output levels for the DAC board.

\section*{DAC CIRCUIT DESCRIPTION}

The complete circuit of the Digital-toAnalogue Converter (DAC) is shown in Fig. 4.5. TTl logic level inputs to the DAC (IC2) may be derived from the microcomputer user port or selected manually by switches S1a to S1h (an 8way s.p.s.t. d.i.l. switch is used here). With any switch closed, the input line to which it is connected is brought down to 0 V (logic 0). An open switch allows the input line to be pulled up to the +5 V line, forcing it to a logic 1. Resistors R1-R8

Table 4.2. Basic Bipolar Output Operation



\section*{ASSEMBLY DETAILS}

The components for the Digital-toAnalogue unit are accommodated on a single-sided printed circuit board, size \(200 \times 98 \mathrm{~mm}\). The pattern to be etched on the board is shown in Fig. 4.7. This board is available from the EE PCB Service, Order code 8310-01.

Position and solder the components on the board topside as shown in Fig. 4.8. Order of assembly is not important. Use of i.c. sockets for the d.i.l. integrated circuit is advised. In the prototype, Sl was also fitted in a d.i.l. socket. Pay attention to orientation of the l.e.d.s and IC2 and the other i.c.s when inserted in their holders after construction is complete.

The s.i.l. resistor package has its common pin marked on the body with a spot. Make sure this component is fitted the right way round.

The appropriate wires of the ribbon cable from the user port connector are to be soldered directly to the same p.c.b. land areas which accommodate the uncommoned side of switch Sl. The wires


Fig. 4.7. Master pattern (actual size) for the DAC board. This board is available through the EE PCB Service; Order code \(8310-01\).

O = VEROPINS

\(\mathrm{R23}\)
\(-1!1 ?\)




\section*{COMPONENTS}

\section*{Resistors}

R1－8 \(4.7 \mathrm{k} \Omega\) s．i．1．commoned resistor network
R9，10，22 4．7k 2 （ 3 off） R11－18 \(390 \Omega\)（8 off） R19，20 \(10 \mathrm{k} \Omega\)（ 2 off ） R21 \(30 \mathrm{k} \Omega\)
R23 100k \(\Omega\)
R24 \(2.7 \mathrm{k} \Omega\)
Capacitors
C1 100 nF polyester type C280
C2 \(\quad 1 \mu \mathrm{~F} 6 \mathrm{~V}\) elect．radial
C3 10 nF polyester type C280
C4 \(\quad 100 \mathrm{nF}\) polyester type C280
C5 \(\quad 100 \mathrm{nF}\) polyester type C280
Semiconductors
D1－8 TIL209 red l．e．d．（8 off）
IC1 \(78 \mathrm{LO5}\) monolithic 5 V 100 mA voltage regulator
IC2 DACO800 digital－to－ analogue converter IC3 ULN2803 octal Darlington driver
IC4 741 op－amp

\section*{Miscellaneous}

VR1 \(10 \mathrm{k} \Omega \frac{3}{\mathrm{~B}}\)－inch cermet multi－turn preset
S1 8－way s．p．s．t．16－pin d．i．l． switch
TB1 4－way p．c．b．mounting screw terminal block
Printed circuit board：single－ sided size \(200 \times 98 \mathrm{~mm}\) ，EE PCB Service；Order code 8310－01）； Veropins single－sided（ 3 off）； tinned copper wire（for board topside links）；di．I．sockets， 18 －pin （1 off）， 16 －pin（ 2 off）， 8 －pin（ 1 off）： self－adhesive rubber feet（ 4 off）； ribbon cable and connector to suit computer user port．
should be soldered after passing through the holes adjacent to Sl commoned con－ tacts．This arrangement acts as a strain relief mechanism to protect the soldered cable connection．

\section*{SOFTWARE}

Controlling the DAC from BASIC is extremely simple．The eight lines available at the user port must be configured for output by setting all bits in the data direc－ tion register for the port．

Once this has been accomplished then any bit pattern introduced into the I／O register will cause the corresponding out－ put signals to operate the data input switches on the DAC and produce the appropriate output voltage．

The following BASIC program allows the conversion of a digital input into a scaled analogue output．
\begin{tabular}{|c|c|c|c|c|}
\hline Line & BBC & VIC－20 & PET & Commodore 64 \\
\hline 5 & \multicolumn{4}{|l|}{\(<-\)－－－－REM DAC PROGRAM－－－－－－－＞} \\
\hline 10 & ？ \(65122=255\) & POKE37 & F＇OKE59459，25 & POKE56579，255 \\
\hline 20 & \multicolumn{4}{|l|}{\(<-\sim-\)－－－－INPUTX－－－－－－－} \\
\hline 36 & \multicolumn{4}{|l|}{\(<-\)－－－－if \(\bar{x}<\emptyset\) OR \(x>255\) THEN END \(-\cdots-\cdots\)} \\
\hline 46 & \(? 6512 \phi=\mathrm{X}\) & POKE37 & POKE59457， X & POKE56577，X \\
\hline 56 & \multicolumn{4}{|l|}{\(<-\)－二－二－二二ー} \\
\hline
\end{tabular}

BASIC is quite adequate for any slow applications such as the real－time control of mechanical devices or a programmable power supply．Machine code software， however，is often required for waveform generation（music or speech for example） but BASIC can be used to produce alter－ nating waveforms with longer periods．

\title{
HIGH POWER DAC DRIVER BOARD construction
}

\section*{HIGH POWVER DRIVER}

Whilst the DAC board is capable of providing programmable voltages over a wide range, the current that may be drawn from it is small because of the limited power capability of the buffer amplifier. Fig. 4.9 shows the circuit of an add-on power amplifier (the DAC Power Driver board) which will provide programmable voltages at currents of several amperes when connected to a suitably rated power supply.

The analogue output from the DAC board is applied to the non-inverting input of IC 1, a 741 operational amplifier, the output of which drives the complementary Darlington power amplifier stages TR1/TR3 and TR2/TR4. Overall negative feedback is applied to the inverting input of the operational amplifier which reduces the voltage gain of the circuit to unity. An offset null provision is

\section*{COMPONENTS}

Resistors
\[
\begin{aligned}
& R 1 \quad 12 \mathrm{k} \Omega \quad \mathrm{R} 2 \quad 390 \Omega \\
& \text { All } \frac{1}{4} \mathrm{~W} \text { carbon } \pm 5 \%
\end{aligned}
\]

Capacitor
C1 470pF ceramic
Semiconductors
TR1 BC107 silicon npn
TR2 BCY71 silicon pmp
TR3 2N3055 silicon npn
TR4 MJE2955 silicon pnp

\section*{Miscellaneous}

VR1 \(22 \mathrm{k} \Omega\) horizontal skeleton preset
TB1 2-way p.c.b. mounting screw terminal block
Printed circuit board: singlesided size \(177 \times 96 \mathrm{~mm}, E E P C B\) Service, Order code 8310-02; single-sided Veropins ( 5 off ); 8 pin d.i.1. socket; heatsinks, TO-3 type \(14^{\circ} \mathrm{C} / \mathrm{W}\) or better ( 2 off); self-adhesive rubber feet ( 4 off); 4BA fixings ( 4 sets).

Approx. cost
Guidance-only
necessary and this is achieved by the inclusion of VR1. Input impedance is set at \(12 \mathrm{k} \Omega\) by R1.

Output transistors TR3 and TR4 are high power ( 115 W ) devices and must be provided with adequate heatsinks if large load currents are envisaged. The prototypes employed have heatsinks rated at \(14^{\circ} \mathrm{C} / \mathrm{W}\) and these were adequate for currents up to about 2 A ; larger currents will necessitate the use of larger heatsinks.

The board is quite capable of driving small d.c. motors whose speed and direction can thus be programmed.

\section*{ASSEMBLY}

The High Power Driver circuitry of Fig. 4.9 is assembled on a single-sided printed circuit board, size \(177 \times 96 \mathrm{~mm}\). The pattern to be etched on this board is shown in Fig. 4.10. The black regions represent the copper tracks to remain after etching. This board is available through the EE PCB Service, Order code 8310-02.

The components are positioned on the p.c.b. topside as shown in Fig. 4.11. The order of component assembly is unimportant. Input and power supply connections in the prototype were made by soldering to board fitted Veropins, with the output connection accessible via a screw terminal block. Spacing of the Veropin location points allow p.c.b.-type screw terminal blocks to be fitted if desired.

The power transistors TR3 and TR4 need to be fitted on heatsinks and these are sandwiched between board and device and held by 4BA screws and nuts. Use heatsink compound or silicone grease on touching surfaces to increase thermal conductivity from device to heatsink.

Screw fix these assemblies before soldering the two leads from the transistor to the board.

The collector connection is made at the case of these transistors, via the fixing nut and screw, the heatsinks must not be allowed to come into contact with one another.
Next Month: Stepper Motor Control.

Fig. 4.9. Circuit diagram of an add-on High Power Driver.



Fig. 4.10. Master pattern (actual size) for the High Power Driver. This board is available through the EE PCB Service; Order code 8310-02.


Fig. 4.11. Layout of components on the topside of the p.c.b. The heatsinks for the two power transistors are indicated by the tone areas


Input sensitivity (D.C. and A.C. ranges): 2,000 ohms per volt.
Ranges: A.C. \(0-10 \mathrm{~V}, 0-50 \mathrm{~V}, 0-250 \mathrm{~V}, 0-1,000 \mathrm{~V}\).
D.C. \(0-2.5 \mathrm{~V}, 0-10 \mathrm{~V}, 0-50 \mathrm{~V}, 0-250 \mathrm{~V}\), \(0-1,000 \mathrm{~V}\)
\(0-500 \mu \mathrm{~A}, 0-5 \mathrm{~mA}, 0-250 \mathrm{~mA}\).
Resistance: \(0-5,000\) ohms, \(0-500,000\) ohms
Diode and fuse protection incorporated.
Overall dimensions: \(950 \times 600 \times 340 \mathrm{~mm}\).
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}

\title{
CAR ON/OFF TOUCH SWITCH
}

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An alternative to the normal on/off mechanical switches used in cars. May also be adapted for any application where on/off switching is required.

\title{
RADIO WORLD
}


BY PAT HAWKER G3VA

\section*{All Change}

Changes in the regulation of amateur radio licences in the UK have been coming thick and fast. First, following the General Election, the entire Radio Regulatory Department in Waterloo Bridge Road was transferred lock, stock and barrel, from the responsibility of the Home Secretary to the Secretary of State for Trade and Industry.

Then, on 27 July, DoTl published the final report of the "Independent Review of the Radio Spectrum ( \(30-960 \mathrm{MHz}\) )" by the three wise men chaired by Dr James Merriman recommending the setting up of an entirely new ministerial department to be responsible for telecommunications, including all radio regulatory matters but (as a third choice) RRD to be with DoTI rather than Home Office. But the report had virtually nothing to say about amateur radio or CB.
The following day, DoTI announced in Parliament the transfer on 19 September of amateur radio licensing (including the issue and renewal of licences) back to its old home, the Post Office, who already issue CB licences. Unlike CB, however, amateur radio licences will not be issued over the counter but dealt with by postal application at the PO headquarters. (The address will be: Radio Amateur Licensing Unit, Chetwynd House, Chesterfield, Derbyshire S49 1PF, telephone Chesterfield (0246) 207555 who will issue the necessary application forms, etc).

A plus point is that the Post Office is guaranteeing a turn round in normal conditions of five days and at peak times of ten, and are to put all records on computer. The long delays of dealing with Waterloo Bridge House should soon be over.

The minus points? These are difficult to predict and it may be only my natural pessimism that makes me feel that more and more amateur radio in Europe is being lumped in with CB by people who fail to see that the two are essentially different.

This is not to deride CB, a useful leisure pastime and a low-cost communications facility which I was advocating many years before the Home Office finally gave way. I have always believed that not only is amateur radio a more fulfilling hobby but also has genuinely contributed to selftraining, technical education and the advancement of the radio sciences.

\section*{First Impressions}

The impression I have is that the authorities, whichever ministry is involved, nowadays are concerned only with reducing the administrative burden of catering for the 48,000 (and still expanding) UK radio amateurs. Unlike the USA, USSR and Japan, the UK has consistently declined to
introduce any form of "incentive" licensing, or any form of "novice" licence.

But then little official encouragement is given these days to British licensees to progress from the Class B (v.h.f.-only) to the Class A (v.h.f./h.f.) licence. Even the current president of the R.S.G.B. (a former senior member of the Radio Regulatory Department) is reported to have emphasised recently that "the Class B permit is in no way to be regarded as inferior to the Class \(A\) version but simply reflects an interest in v.h.f./u.h.f. techniques rather than in h.f.-bands communication"

This is hardly a positive encouragement to Class \(B\) licensees to prepare for a 12 words per minute Morse test! A very different attitude from the American Radio Relay League which continues to support graded Morse tests for all amateur radio licences!

\section*{The Merriman Report}

As someone with an interest to declare in both radio communications and broadcasting I would have liked to feel that the three wise men of the Independent Review committee have been truly impartial in their recommendations. But the fact is that, both in the interim and final reports, broadcasting has been clobbered unmercifully by a committee on which broadcasting was not represented at all, although accounting, at the outset, for the use of 45 per cent of the spectrum between 30 and 960 MHz . This percentage may seem excessive until you remember that virtually every single person in the UK benefits from broadcast services of entertainment, information and education; relatively few from their own mobile communications.

On the other hand, most of us have a suspicion that the very large parts of the spectrum reserved for Defence purposes are vastly under-used in peacetime and are unlikely ever to be filled even in a major war. Yet the Merriman report saw little scope for any reduction in Defence spectrum, a conclusion with which one presumes at least one of the three wise men-a former Assistant Chief of Defence Staff (Signals) and now a member of a firm prominent in Defence communications equipment-must have been in full agreement.

One recalls that no attempt has been made to re-establish the old Television Advisory Committee. The committee incidentally that presupposed the re-engineering and Bands 1 and 111 for 625-line television.

There are many other curious features of the Merriman report. The insistence that UK frequency block allocations remain "an official secret". The idea that an American FCC-type regulatory body could not work
because only a government-body can deal with Defence frequencies. Why not leave detailed Defence assignments to MoD in the American manner?

The more one reads the report the more one gains the impression that the "independent review" was set up with two ideas in mind: (1) to wrest RRD away from the Home Office; and (2) to clobber the broadcasters. It appears to have succeeded in both.

\section*{In The Background?}

There is a deep-rooted belief that complaints to broadcast engineers about the quality of or faults in broadcast transmissions tend to be met with a single stock phrase: "It was all right leaving me"'

In practice this is often an entirely justified statement: a networked broadcast signal goes through such a complex path after it leaves its originating studio that there is almost unlimited scope for the gremlins to get to work.

An example, fortunately of a rare type, occurred recently on one of my recordings for the weekly IBA's Engineering Announcements that go out on Tuesdays (9.15 a.m.) on Channel 4. As luck would have it I was describing the new automatic test equipment (MATE) becoming available to the 50 IBA mobile maintenance teams that spring into action when anything goes wrong at a transmitter, most often these days due to the effects of lightning on overhead mains supply lines or on equipment. But my words were accompanied by background music that seemed determined to become "foreground"

A first reaction was that the producer or someone in master control had decided to liven up the programme but had overdone it a bit. When 1 enquired, everyone firmly denied having any part in jazzing me up and assured me it was all right leaving them.

Finally, it was discovered that the music was entering the system at BT's National Television Switching Centre in London. But the mystery remains-whose music was it? Where was it originated? How did it get mixed into network television? And for what purpose was it going along highquality music lines? Only British Telecom knows the answers!

\section*{Signals From Space}

Although there are still sensitive receivers attached to giant aerials searching (so far vainly) for radio signals that could denote the existence of extra-terrestrial intelligence, enormous scientific knowledge has been gained over the past 50 years from random noise and also from the curious pulsar signals, first observed in the 1960 s and initially thought to come from those elusive little green men.

It was exactly 50 years ago that Karl Jansky of Bell Laboratories, first announced that he had found that radio signals were coming from space, while investigating atmospherics. And almost 45 years since an American radio amateur, Grote Reber, W9GFZ, built the world's first steerable radio telescope.

It is also 41 years since British radar expert 'J. S. Hey recognised that what was thought to be deliberate enemy jamming was the result of an enormous solar flare and became "the father of British radioastronomy," inspiring Sir Bernard Lovell, Sir Martin Ryle and J. L. Pawsey to begin intensive investigations that developed into the science of radioastronomy and so opened a new window on space.

\section*{WORLD'S SMALLEST COLOURTV}

Pictured here is the world's smallest colour TV from National Panasonic Division of Matsushita Electric of Japan which had its first public showing at this year's Chicago Consumer Electronics Show

Designed around a conventional, but newly-developed, 1.5 in screen, miniature picture tube, the colour "Travelvision" is equipped with video input terminals, making it convenient as a video monitor as well as a portable TV. The set can be powered from a.c., d.c. car battery, rechargeable and dry batteries.

Before you rush out to try and purchase a set, it should be pointed out that Travelvision is only suitable for the American TV broadcast system. To-date, no plans have been tabled for the UK and European systems.

Designed specially for the American market, it looks as though Travelvision has stolen a march on the Sinclair flat-screen TV. As the Americans would say, "Eat Your Heart Out Clive".


\section*{COMPUTER FAIR}

The second London Computer Fair, sponsored by Practical Computing and Your Computer, was held at Earls Court from 16 June for four days. Last year, this event saw nearly 38.5 thousand visitors through its gates and this year, the figure has increased to nearly 48 thousand.

The growth of the Computer Fair reflects the enormous growth in the personal computer market, for it is users at this end of the market that the show caters for. Machines ranging from the diminutive Sinclair ZX81 at \(£ 40\) to powerful business microcomputers costing more than \(£ 3000\) will account for an astonishing \(£ 1.4\) billion worth of business in 1983 .
Such is the success of this event that similar shows have been organised for the North (Belle Vue, Manchester, 24-26 November, 1983), the Midlands (Bingley Hall, Birmingham, 3-6 May, 1984) and back in London for a special Christmas Fair at the Wembley Conference Centre (15-18 December, 1983).


All the major manufacturers were at the Fair although few had anything really new to display. However, with lots of working machines on their stands, potential purchasers can get "hands-on" experience and talk to the experts all under one roof.

Software to try out was also in abundance, mostly in the form of arcade type games (and the imagination of some of these games compilers is stretched to incredible limits!), but also some home and business management systems.

The Acorn stand was devoted entirely to the BBC microcomputer range and the forthcoming Electron computer was noticeable only by its absence. However, they were demonstrating the Econet local area network, which can link up to 254 BBC micro stations in a data exchanging system.

Exhibiting at their first UK exhibition, Oric Products International had on display the 16 K and 48 K machines and the recently launched Oric four-colour printer. This retails at \(£ 170\) and has full alphanumeric and graphics capabilities.

Maplin Electronic Supplies introduced Cashtel (Computer Aided SHopping by TELephone) at the Computer Fair, a system whereby customers can interrogate Maplin's stock levels and place orders directly using a variety of computers at home. Also displayed was their extensive range of hardware, software and add-on kits.

Also worthy of a mention is the new Texas Instruments CC-40 portable computer. With a liquid crystal display and measuring only \(240 \times 145 \mathrm{~mm}\), the CC- 40 offers true briefcase portability. With 6 K of on-board ram expandable to 18 K and low cost peripherals including a printer/plotter, Wafertape data storage and RS232 interface, it forms a versatile system.

A company more known for their high quality add-ons for the ZX81, Memotech had their first public showing of the all new MTX-500 computer, suitable for home, business or educational use. Standard features include 32 K of user ram (expandable to 512 K ), 16 K resident rom, 16 colours and a 79 -key keyboard.

In all, there were over 130 exhibitors, 23 of these in the special "Sinclair Village", a separate portion of the show dedicated exclusively to companies producing a whole range of goodies for the ZX81 and Spectrum. Finally, a "Club Avenue" for regional and national computer clubs completed the scene, some of which now produce very professional magazines for members.

\section*{electronics}

\section*{LIVING WITH ROBOTS}

The British Engineerium, Off Nevill Road, Hove, East Sussex has launched a new exhibition called Living With Robots and the official opening took place on 29 July, with Patrick Moore, the television presenter presiding over the ceremony.

The display incorporates many of the more famous robots of television fame. A replica of K9, the Dr Who robot was on display but it did not seem too impressed with its surroundings and would not work. Star Wars was also represented with replicas of models used in the film on show and last but not least the industrial robots like the 600-Fanuc, a programmable robot welder.

\section*{London Earth Station}

A satellite earth station in the heart of London's dockland is to be built by British Telecom International.

The earth station, London's first, will be operational early next year. It will transmit television programmes to cable TV systems in the UK and in Europe, and provide the capital's businesses with advanced telecommunications services.

The earth station will be located at Pier Road, North Woolwich, close to the old King George V Dock. Initially two dish aerials, 13 m ( 43 ft ) in diameter, will be ready for service early in 1984. One will operate through the European Communications Satellite (ECS) and the other through an INTELSAT satellite.

The Western Australian Government is to establish a Technology Directorate as a focal point for its drive to encourage technological development in the area.

\section*{EVENTS}

\section*{SHUTTLE POWER}

A contract worth more than \(\$ 100,000\) has been placed by NASA with Gould Micro Power Products Division to provide zinc-air power packs for use on board the Space Shuttle.

The power packs, which each contain 60 zinc-air power cells, will be used to operate an in-cabin radio system that enables astronauts to communicate with each other and with the ground, while maintaining freedom of movement inside the orbiter.

\section*{RADIO TRANSFER}

The Radio Regulatory Department, including the Directorate of Radio Technology, has been transferred from the Home Office to the Department of Trade and Industry.
The Broadcasting Department of the Home Office is not being transferred. Its responsibilities include: constitutional aspects of broadcasting; cable programme services; direct broadcasting by satellite (DBS); BBC and IBA finance; TV licensing; technical planning policy; and local and community radio.

The Great Home Entertainment Spectacular takes place from 17 to 25 September at Olympia, London.

The 10 th International Broadcasting Convention-IBC 84, will be held at the Metropole Conference \& Exhibition Centre, Brighton from 21 to 25 September. Technical Sessions will be run throughout the show.

Sponsored by Jeffries Hi Fi, the South Coast Hi Fi Show is to be held at the Royal Albion Hotel, Old Steine, Brighton, from 11 to 13 November, 1983.

The international cable and satellite television industry is to have a major exhibition at its European focal point in Basel, Switzerland. "Eurocast ' \(8 \mathbf{4}\) " will take place in the halls of the Swiss Industries Fair from 5 to 9 May next year.


HRH Prince Philip talking to Mr. \& Mrs. Frank Chorley after the award ceremony.

\section*{Prince Philip Medal}

The 1983 Prince Philip Medal of the City and Guilds of London Institute was presented to Mr. Frank Chorley by HRH Prince Philip in the Music Room at Buckingham Palace. Mr. Chorley is Deputy Chairman and Managing Director of Plessey Electronic Systems Ltd.

Since 1962, the City and Guilds of London Institute has conferred the Prince Philip Medal, its highest award, on an individual who has received an Institute qualification and subsequently shown outstanding achievement in technology or in service to industry. Only one medal is awarded each year and the final choice is made by HRH Prince Philip who is the Institute's President.

\section*{British Lead}

British standards are being used in ninety-eight per cent of viewdata and teletext sets throughout the world. This is revealed by a survey undertaken for British Telecom, Mullard Ltd. and the Department of Trade and Industry.
The survey, which identified more than \(2 \frac{1}{4}\) million viewdata and teletext sets in 26 countries, was carried out by the US-based consultancy, CSP International Inc.

All amateur radio licences will be issued and renewed by the Post Office headquarters in Chesterfield on behalf of the Secretary of State for Trade and Industry.

The annual salary survey of pay and other benefits in computing, conducted by the National Computing Centre, once again shows that, on straight salary terms, salaries tend to be higher in the public sector than in private industry.

\section*{NO FAIR THIS YEAR}

\section*{Electronic Hobbies Fair Victim of Recession}

An announcement in this issue explains the reasons behind the decision not to proceed with this year's Electronic Hobbies Fair. The news will be a great disappointment to thousands who enjoyed their visit to last year's Fair and had been looking forward to this year's event. Additional special attractions had been lined up to extend the overall interest of the Fair, and a number of national organisations had declared their intention of participating. But in the end the whole exhibition depended upon the retail trade that serves the hobbyist and home constructor.

Trade is still in the doldrums and despite frequent optimistic predictions from economists, the eagerly awaited revival still seems as far off as ever. So, for the present, we can only hope for a general improvement before too long, and for a future Electronic Hobbies Fair that will signal this change in fortunes in an unmistakable manner.

\section*{Two-Tone}

Sir-Seeing Mr. Johns' item in the July '83 Circuit Exchange (page 423) about his Two-Way Doorbell reminded me of the modification I made to the Modulated Tone Doorbell from the March '81 issue of Everyday Electronics
This involves inserting a "steering diode" between the two halves of the circuit. An additional switch, S 2 , will apply 9 V to both halves, giving a frequency modulated tone as normal. The original switch, S1, now only applies the power to IC 1 b and associated components, giving a single, clear tone
A. B. Wilson,

Bradley, Yorkshire.

To effect Mr. Wilson's modifications, the diode is placed in the 9 V line between IC1a and IC1b with the cathode ( \(k\) ) towards IC1b (referring to Fig. 1, page 180, March '81 issue). It is important that the power connection to IC1 (pin 14) is transferred to the IC1b side of the circuit. Pins 4 and 10 remain as shown.

The additional push-button switch (S2) is connected between the anode (a) of the diode and the positive ( + ) of the battery, B1. A 1 N4148 diode is a suitable device. Some alteration of stripboard layout is necessary.

\section*{Key Question}

Dear Paul Young-With reference to your piano keyboard question in Counter Intelligence in the August issue of EE .

Established musicians will quite naturally be against typewriter or computer keyboards. I can imagine that after having spent years in becoming proficient on a traditional keyboard, the thought of spending a long period adjusting to an "alien" keyboard, trying to rid yourself of deeply ingrained techniques would come as a deep shock. 1 think that it is thus quite understandable that musicians will not like the idea of new-type keyboards.

On the other hand there is one vast disadvantage of computer keyboards. Piano keyboards are essentially "analogue" devices. They produce a sound whose loudness is proportional to their velocity. In other words, hit the keys harder and you get a louder noise. Computer keyboards, on the other hand, are "digital" devices. They are either "on" (pressed) or "off" (not pressed). This lack of expression due to using electronic keyboards is a serious handicap to a musician's creativity. This lack of expression is not such a handicap as one may suppose, as good recording and re-mixing facilities can work wonders. It does, however, mean that "live" playing is severely restricted.

The ideal solution is to produce a musical instrument which can be "pre-programmed" by a computer keyboard or played "live" by a traditional keyboard, with "dynamic expression" (the harder you hit the keys, the louder the sound produced).

The Fairlight computer musical instrument possesses all these (and a great many more) abilities, its only real drawback being its price. At around \(£ 14,000\) each I don't think I will be giving many as Christmas presents this year.

I have noticed that from time to time you have expressed an interest in the paranormal . One method of testing for telepathy is achieved by allowing a computer to secretly choose one of five "Zener cards". The subject under test then has to guess which card the computer chose. This process continues for a set number of guesses after which the computer prints the percentage of correct guesses.

You will be asked to submit 100 guesses by the computer. The arithmetic behind the results is as follows:

In five guesses you may score one by pure chance.
In 100 guesses you may score 20 by pure chance.
Hence any score significantly above, say 25 per cent (a statisticaliy adjusted figure to take into account the small number of guesses) may, if repeatable, indicate precognition.

For more information about current thoughts and research in this field consult the book "Mind Over Matter" by Kit Pedler, published by Thames/Methuen. This book provides a layman's guide to the subject and, perhaps most important of all, is completely non-sensationalistic: a definite antidote for an overdose of the VonDaniken's!

In conclusion, may I say that I have read and enjoyed your articles in EE since 1974 and have enjoyed them all. Please keep it up, your article is the first thing I read when I receive EE.

Christopher Stone, Hull,
North Humberside.

\section*{Soft Apple}

Sir-As a rule, I do not purchase an electronics magazine unless an initial perusal reveals an interesting article likely to be completed without too much difficulty.

Accordingly, the article Real-Time Clock Apple // in Everyday Electronics (Vol. 12, No. 5, May 1983) caught my attention and was read with interest.

However, a closer study prompts me to believe that the article is incomplete, as generally throughout the article and specifically on page 274, reference is made to enclosed programs for initialisation, testing, setting and reading the clock when installed on the Apple II and these appear to have been omitted from the above edition. Your comments would oblige.
D. Elliott,

Brisbane, Australia.

In reply to your comments concerning the omission of the program listings for this article, it is not our normal policy to print the software funless it is a very short routine-a test program for example) as we are primarily an electronics magazine and therefore more concerned with the hardware. All programs are available from the EE Software Service, either in cassette form or as a listing. The order codes for this project (Apple II version) are \(T 002\) and LOO2 respectively, and the current price is \(£ 2.95\) inclusive.

We regret any confusion arising from the wording of this article.

\section*{Oric Interface?}

Sir-l read with interest the article Microcomputer Interfacing Techniques by Mr. Adams and Mr. Feather, in your July 83 issue.
I am an avid reader of your publication and found the article most enlightening; however, I noticed you select four specific microcomputers and provide abundant details with relation to those particular machines. However, as the owner of an Oric 164 K computer, I am at a complete loss to know how to utilise the information contained in the last edition.
I would be obliged if it were possible for you to supply details in order that the circuitry in question can be interfaced with the Oric and look forward to hearing from you at your earliest convenience.
N. R. Fineman, Southgate, London.

Unfortunately your Oric 1 computer does not have the necessary user port to carry out the interfacing techniques described in this series. Although the Oric uses the 6502 microprocessor las used by the four computers described in the series), it does not have the 6522 Versatile Interface Adaptor integrated circuit on board so an additional user port add-on is required.

At present, we do not have a suitable circuit on file, but you may be able to obtain one from an outside supplier. However, we have noted the need for such a project and a design may well be featured in the not too distant future.

\section*{Gen On The Generator}

Sir-l have followed the Test Gear '83 series with zest, making each project as soon as your magazine comes out. So what has happened to the Frequency Meter?

Incidentaliy, the value of the tantalum bead capacitor (C15) in the Pu/se Generator (July ' 83 issue) cannot be obtained. It is listed as a \(100 \mu \mathrm{~F}, 25 \mathrm{~V}\) component, but a \(100 \mu \mathrm{~F}, 10 \mathrm{~V}\) tantalum capacitor is readily available and works well.

The VHF Aerial Pre-Amplifier by Hamid Reza Tajzadeh (Circuit Exchange, August '83) was found to be of little use as all modern TVs are UHF but it worked well on my car radio. Also, is there more "gen" on the Pulse Generator testing and use?
L. R. B. Turner,

North End,
Portsmouth.
With regard to Mr. Tajzadeh's VHF PreAmp, the TV system in Iran is still VHF lunlike our own, which has recently switched exclusively to UHF) but we felt that this circuit would be useful in other applications, for example, f.m. radio.

As far as more information on testing and using the Pulse Generator, it is our usual policy to only publish the details supplied by the author if we deem these to be sufficient. However, if you have any particular problems, we would be happy to answer them or forward your query to the author.

Concerning the Frequency Meter, the finat unit in the series, unfortunately due to circumstances beyond our control this instrument will not now be featured in this magazine. We hope this has not inconvenienced too many readers.

\section*{cIrcult EXCHANGE}

This is the spot where readers pass on to fellow enthusiasts useful and interesting circuits they have themselves devised. Payment is made for all circuits published in this feature. Contributions should be accompanied by a letter stating that the circuit idea offered is wholly or in significant part the original work of the sender and that it has not been offered for publication elsewhere.

\section*{F.E.T. DARKROOM TIMER}
\(\mathbb{M Y}^{Y}\) circuit comprises a pair of f.e.t. - transistors wired as a monostable, a transistor interface switch, and oscillator circuit that drives an l.e.d. on and off (flashing) when the set time is reached.

The timing monostable is made up of f.e.t. transistors TR1 and TR2 wired in the source (emitter) coupled mode with capacitor C2 and R3, R5, R6 and potentiometer VR1 being the time determining components.

The advantage of using f.e.t. transistors is obvious as we all know that these are voltage operated devices and as such does not impose limitations on the values of timing resistors; thence an extended time is possible. \(S 1\) is the time start push-button switch, which when depressed pulls TR2 gate to earth potential, reducing TR2 conduction, thus reducing the voltage created by resistor R8. TR1 then begins to conduct and brings down its source voltage across R2.

Toh Eng Kiong, Singapore.


This circuit is capable of monitoring four occupants in a house; of course my basic design could be expanded to register more than this number.

When the main switch is on, l.e.d.s D12-D15 are lighted, indicating that the four occupants are OUT with switches S1 to S4 opened. L.e.d.s D1-D4 are fitted in the occupants' rooms and will only be lighted when their respective switches are closed, to indicate that the occupant of that particular room is IN. For example,

S1 for l.e.d. D1, S2 for l.e.d. D2, etc. When this happens any of the related l.e.d.s D 12 to D 15 will switch OFF. When all the four occupants are IN each closing their respective switches S1 to S4, l.e.d.s D1 to D4 light up, whilst D12 to D15 shut off and D11 will glow to indicate that all four occupants are IN. The master indicator l.e.d. D11 and individual room indicators D12 to D15 are contained on the main control panel installed in the Monitor's room.

This system is effective to eradicate the problems of determining the presence of the four occupants especially in the night when the main house door has to be closed, thus ensuring that none of the occupants are shut out. Of course this system can only operate effectively if the occupants are made to remember to switch on their respective switches when in their rooms and to shut off (open) the respective switches when leaving.

Toh Eng Kiong, Singapore.

\title{
SHORT WAVE RADIO
}

\author{
BY R.A.PENFOLD
}

\section*{Covers the entire short-wave spectrum from 1.5 MHz to 30 MHz in three tuning ranges}

DESPITE increasing competition from other fields of amateur electronics, short-wave radio remains very popular, and deservedly so. Commercial communications receivers tend to be rather expensive, but it is possible to learn a great deal about short-wave radio and have just as much fun using an inexpensive home-constructed receiver.

The short-wave radio described here covers the entire short-wave spectrum in three tuning ranges which are (approximately) as follows:

\section*{Range \(3 \quad 1.5 \mathrm{MHz}\) to 4.8 MHz \\ Range \(4 \quad 4 \cdot 6 \mathrm{MHz}\) to \(14 \cdot 5 \mathrm{MHz}\) \\ Range \(5 \quad 9 \mathrm{MHz}\) to 30 MHz}

Ready-made coils are used and thè range numbers used above are those of the coil manufacturer (Denco). Plug-in band changing is used, and this simply entails plugging-in the appropriate coil for the desired tuning range. This is a common method of band changing in simple short-wave sets, and helps to minimise the cost and complexity of the set.

The output of the set is intended for high impedance ( 4000 or 2000 ohm) headphones, but medium impedance
types also produce good results. A crystal earphone gives acceptable quality at minimal cost. A low impedance earphone is not suitable.

An external aerial is required, but this does not need to be a large or elaborate type, and even a few feet of wire mounted indoors will give quite good reception.

\section*{THE CIRCUIT}

The circuit is a tuned radio frequency (t.r.f.) type, and although this is a basic type of receiver, it can nevertheless give very good results if used skilfully. Fig. 1 shows the circuit diagram of the receiver.

The aerial signal is coupled to the primary winding (L1) of r.f. transformer T 1 . The main secondary (L2) of T 1 forms a parallel tuned circuit with variable capacitors C1 and C2. This tuned circuit acts as a simple filter which selects signals over a narrow band of frequencies, and attenuates signals outside this passband. Cl is the main TUNING or bandset control, and C2 is used for fine tuning and is called the bandspread control.

The output of the tuned circuit is directly coupled to the input of a common
source amplifier based on TR1. This stage provides detection in addition to r.f. amplification. Detection is produced because TR1 tends to amplify one set of half cycles more than the other, giving an inefficient form of rectification. C3 is the r.f. filter capacitor and C4 couples the audio output signal to the next stage of the receiver.

A third winding (L3) on T1 is included in the drain circuit of TR1, and this provides positive feedback over the circuit. In this way, some of the signal is coupled back to the input of the circuit to be amplified for a second time. This is known as regeneration, and VR1 shunts the L3 winding on T1 so that the amount of regeneration can be controlled. The higher the resistance through VR1, the less it shunts the feedback winding L3 and the greater the amount of regeneration that is obtained.

Apart from boosting the gain of the circuit, regeneration has two other useful effects. Firstly, there is more feedback at frequencies towards the centre of T1's passband than there is towards the edges of the passband. This tends to boost gain more at the centre than at the edges, shar-

Fig. 1. The circuit diagram of the Short Wave Radio. T1 is the interchangeable Denco tuning coil.


pening the response of the filter. Thus the "selectivity" of the receiver is increased so that it is more able to pick out the desired station from the crowded shortwave bands.

Secondly, the efficiency of the rectification provided by TR1 is boosted by the regeneration, giving a substantial improvement in detection efficiency.

There is a limit to the amount of regeneration that can be used, and if this limit is exceeded, the detector stage goes into oscillation and proper reception becomes impossible. For optimum results, the level of regeneration should be set just below this threshold.

As the audio output from TR1 is at a very low level, it is necessary to have a large amount of audio amplification to produce a strong enough output for sufficient volume from the set. Therefore, two stages of audio gain are used, and the first of these is a high gain, common emitter stage built around TR2.

The signal is then coupled by C 5 to volume control VR2 and then to a second common emitter amplifier, TR3. The full gain of TR3 is not needed and is therefore used to introduce negative feedback which reduces the voltage gain of this stage. C8 reduces the high frequency response of the output stage and aids the stability of the circuit.

The current consumption of the circuit is about 4 mA , and a small 9 V battery (PP3 size) is more than adequate to supply such a small current.
threaded holes in the front plate of the component, or by a single 4BA screw passing through the bottom panel of the case and into a threaded hole in the base plate of the capacitor.

The second method is the simpler of the two, but spacers must be used over the mounting screw to bring Cl to the correct height. With either method, it is essential to use short screws that will penetrate Cl by no more than 3 or 4 mm , as the threads might otherwise damage the plates.

\section*{COIL BRACKET}

Tl is mounted on an " L " shaped aluminium bracket made from 16 or 18 s.w.g. aluminium, and this is detailed in Fig. 4. The lower part of the bracket is fixed to the base panel of the case using 6BA fixings, and a B9A valve holder (SK3) is mounted on the upper part of the bracket, again using 6BA fixings. The positions of the two small mounting holes for the valve holder can be marked on the bracket using the holder as a template, once the main 19 mm diameter hole has been made using a chassis punch.

The coils have a base which plugs into the B9A valve holder, and the bracket is mounted in a position that enables the
coils to be plugged into the holder through a 19 mm diameter hole punched in the rear panel of the case. This enables the coils to be changed over without having to remove the outer cover of the case.

\section*{COMPONENT BOARD}

The component panel is a 0.1 inch matrix stripboard having 31 holes by 16 copper strips, and this is illustrated in Fig. 2. First cut out a board of the specified size using a hacksaw, and then drill the two 6 BA clearance ( 3.3 mm diameter) mounting holes. Then make the single break in the copper strips before fitting and soldering the components into place, leaving the semiconductors until the end.

Veropins are used at points where offboard components will be connected, and these are tinned with solder. The finished board is then mounted on the base panel of the case between T1 and VR1. Use spacers over the 6BA mounting screws to hold the connections on the underside of the board clear of the metal case.

Fig. 3 shows the wiring needed to complete the receiver, and this is all done using ordinary \(7 / 0.2 \mathrm{~mm}\), p.v.c. insulated connecting wire. The leads carrying r.f. signals, particularly those connecting to \(\mathrm{C} 1, \mathrm{C} 2\), and T 1 should be kept as short as possible, but in other respects this wiring is straightforward.

\section*{AERIAL AND EARTH}

A long-wire aerial is required, and this is simply a length of aerial wire connected to SK1. Ideally, an outdoor aerial about 10 or 20 metres long should be used, and the wire should be mounted as high as possible, well clear of buildings or other large obstructions. It should also be insulated from the aerial supports to prevent the signals received from being leaked away to earth.

View inside the chassis showing the general layout of components and the method of

\section*{COMSTRMCTION}

\section*{METAL CASE}

A metal instrument case measuring about \(230 \times 130 \times 64 \mathrm{~mm}\) (or any similar metal case) is used to house the receiver. It is necessary to use a metal case as some of the connections are carried through the case. The general layout of the unit can be seen from the photographs, and it is strongly recommended that this should not be radically altered as it is important to keep some of the leads in the r.f. part of the circuit reasonably short.

Mechanically, the receiver is quite straightforward apart from the mounting of Cl and T1. C1 can either be mounted using three countersunk 4BA screws passing through the front panel and into
inserting the tuning coils through a hole in the rear panel.


\section*{SHORT WAVE RADIO}

- ALL DIMENSIONS IN mm

Fig. 4. The coil mounting bracket for holding the valve base (SK3) parallel to the rear panel.


\section*{COMPONENTS}

Fig. 2. The component layout of the Short Wave Radio circuit board.
\begin{tabular}{|c|c|c|c|c|c|}
\hline \multicolumn{2}{|l|}{Resistors} & C5,9 & 1 1 F 63 V elect. (2 off) & SK1,2 & 4 mm banana socket \\
\hline R1 & \(1.5 \mathrm{k} \Omega\) & C6 & 3.3 nF ceramic 330 nF polyester & & \begin{tabular}{l}
(2 off) \\
B9A 9-pin valve holder
\end{tabular} \\
\hline R2 & \(1.8 \mathrm{M} \Omega\)
\(4.7 \mathrm{k} \Omega\) & C7
CB & 330 nF polyester 10 nF polyester & \[
\begin{aligned}
& \text { SK3 } \\
& \text { SK4 }
\end{aligned}
\] & \begin{tabular}{l}
B9A 9-pin valve holder \\
3.5 mm jack socket
\end{tabular} \\
\hline R3
R4, & \(4.7 \mathrm{k} \Omega\)
\(2.2 \mathrm{k} \Omega\)
(2 off) & C8
C 10 & 10, & T1 & Denco Green D.P. \\
\hline R4,7
R5 & \(2.2 \mathrm{k} \Omega(2 \mathrm{off})\)
\(1 \mathrm{M} \Omega\) & & -00」 & & coils (3 off, ranges \\
\hline R6 & \(100 \Omega\) page 629 & \multicolumn{2}{|l|}{Semiconductors} & & 3,4,5) \\
\hline All \(\frac{1}{4} \mathrm{~W}\) ca & bon \(\pm 5 \%\) & TR1 & BF244B n-channel & 81 & 9V PP3 battery \\
\hline \multicolumn{2}{|l|}{Capacitors} & TR2 & i.f.e.t.
BC109 npn silicon
BC179 pnp silicon & Stripbo by 16 & 0.1 in matrix, 31 holes ips; metal case \(230 \times\) \\
\hline C1 & 365pF Jackson type 0 air spaced variable & TR3 & BC179 pnp silicon & 130 alumin & 64 mm ; \(18 \mathrm{~s} . \mathrm{w} . \mathrm{g}\) \(70 \times 40 \mathrm{~mm}\); high im- \\
\hline & capacitor & \multicolumn{2}{|l|}{Miscellaneous} & pedanc & eadphones with 3.5 mm \\
\hline \multirow[t]{3}{*}{C2} & 25pF Jackson type & VR1 & \(1 \mathrm{k} \Omega\) linear carbon & jack pl & 4 mm banana plug (2 \\
\hline & C804 air spaced & & potentiometer & off, fo & th and aerial); \(7 / 0.2 \mathrm{~mm}\) \\
\hline & variable capacitor & VR2/S 1 & \(10 \mathrm{k} \Omega \mathrm{log}\). carbon & conne & wire; 6BA and 4BA \\
\hline C3 & 22 nF polyester & & potentiometer with & moun & hardware; battery clip; \\
\hline C4 & 470nF polyester & & s.p.s.t. switch & contro & ob (4 off). \\
\hline
\end{tabular}

It is not essential to use an outdoor antenna though, and about 10 metres of aerial wire strung around a room or in a loft will give quite good results, although a proper outdoor aerial will give significantly better reception when propagation conditions are poor.

An earth connection will improve matters, especially on the low frequency bands (that is, when using the range 3 coil). An earth connection can be made to a metal plate or pipe buried in the ground, and the plate or pipe should preferably have an area of about 0.25 square metres or more. The lead which connects it to SK2 can be made from aerial wire and should be no longer than necessary.

\section*{USING THE SET}

VR2/S1 is a combined on/off switch and volume control, and in normal use it is best to advance VR2 to maximum, with this control only being backed-off if a strong station produces excessive volume. C1 is used to tune the set to the band which is to be searched for stations, but fine tuning is accomplished using C 2 . This covers only a small part of each tuning range, and consequently tuning is much easier using this control.

VR1 is the regeneration control, and it is essential to use this control properly if good results are to be achieved. With this control fully backed-off (set in a fully anticlockwise direction) it is likely that few, if any, stations will be received. Advancing VR1 should result in improved results with greater sensitivity and selectivity, but if it is advanced too far, the set will go into oscillation.

With the set in this state, received stations will be accompanied by an audio tone which will vary in pitch as the tuning is adjusted, and it will be impossible to receive stations properly. For best results, VR1 must be carefully set just below the point at which oscillation occurs and it may be necessary to slightly reset VRI each time the tuning controls are shifted by more than a small amount.
When first trying out the set, it is probably best to use the range 4 coil as several of the popular broadcast bands are within the frequency coverage of this range, and these will usually provide a


A plan view of the prototype Short Wave Radio showing the routing of the wiring. Connections to the tuning coils and variable capacitors should be kept as short and direct as possible.
great many strong stations at any time of the day or night. The high frequency bands are included in the coverage of the range 5 coil, and are excellent for long distance reception. However, these are mainly daytime bands, and normally fade-out during the hours of darkness.

There is only one main broadcast band within the coverage of the range 3 coil, and this is quite small. The 160 -metre amateur band will be found towards the low frequency end, and this may provide a.m. transmissions from local "hams".

Single sideband (s.s.b.) and Morse (c.w.) are commonly used on the amateur
short-wave bands in preference to ordinary a.m., and the receiver can resolve this type of transmission if VR1 is advanced slightly beyond the threshold of oscillation. However, the tuning controls must be adjusted very carefully in order to give a proper audio output from an s.s.b. transmission.

The three coils are supplied with their adjustable cores fully screwed down into the coil former, and it will probably be found that correct coverage is obtained with the cores in this position. If the coverage is found to be incorrect, the cores can be adjusted to rectify this.


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Perhaps the humble switch is the most basic electronic component of all; a mechanical device for interrupting the flow of an electric current in a circuit. The most obvious example being the on/off switch fitted to almost all domestic appliances.

\section*{SWITCH CONFIGURATIONS}

The simplest switch form is the Single Pole Single Throw (abbreviated to s.p.s.t.) and this type has a single "pole" with one other terminal, connection being made between the two when the switch is operated. Some examples of these simple on/off switches are given in Fig. 2.

The Single Pole Double Throw switch (s.p.d.t.) has only one "pole" but with two other terminals and activating will switch the connection to the pole from one terminal to the other. For this reason it is also known as a Single Pole Change-Over switch.

The Double Pole Single Throw switch (d.p.s.t.) is really two on/off type switches in the same package and activated simultaneously by the same mechanism. That is, they are mechanically linked but electrically isolated from each other.

In similar fashion, the Double Pole Double Throw (d.p.d.t.) switch is two change-over switches mechanically linked and operated by the same mechanism. Also referred to as a Double Pole



Change-Over switch. See Fig. 1.
A latched switch is one which, having been operated, will stay in that state until it is operated again.

Alternatively, the non-latching or momentary action switch will change its state when depressed but return to its original state when the actuating force is removed.

Miniature push-button switches are examples of non-latching (see Fig. 4) and fit into a further two categories: the push-tomake switch, being off in the natural state and on when pressed (normally open contacts); and the push-to-break switch, being on in the normal state and off when pressed (normally closed contacts).

\section*{OTHER SWITCHES}

The rotary wafer switch (see Fig. 3) consists of a rotating mechanism onto the spindle of which is placed wafer switch elements. The pole of the switch is connected to a wiper contact which sweeps around the wafer making contact with the terminals of the switch. Each position is latched and an adjustable stop enables the switch to be custom designed for any number of "ways".

The microswitch is a useful device (see Fig. 2) usually being a single pole changeover switch operated by a very small push-button and requiring very little actuating force.


Photograph above shows groups of various types of switches, including: (1) slide switches; (2) rotary switches, including the wafer type and a control potentiometer with integral switch; (3) key operated on/off switch; (4) four different sizes of toggle switch; (5) a selection of push-button switches with momentary action types, a microswitch and left, a latching version.
Fig. 1. The circuit symbols of the four basic switch configurations plus the two types of simple push-button switch.


Fig. 1
Fig. 2. Four types of s.p.s.t. (on/off) switch, one of which is mechanically coupled to a control potentiometer. A microswitch is also shown.
Fig. 3. Two rotary switches and the typical circuit symbol. The wafer switch can be exterided by adding more wafers. The maximum number of "ways" for both types is usually 12 .
Fig. 4. Two push-button switches, one with screw terminals (for table lamps and so on), and a miniature non-latching switch.


\title{
TESTED \\ In the cut-throat world of consumer electronics, one of the
}
 questions designers apparently ponder over is "Will anyone notice if we save money by chopping this out?" In the domestic TV set, one of the first casualties seems to be the sound quality. Small speakers and no tone controls are common and all this is really quite sad, as the TV compan ies do their best to transmit the highest quality sound. Given this background a compact and independent TV tuner that connects direct to your \(\mathrm{Hi}-\mathrm{Fi}\) is a must for quality reproduction. The unit is mains operated
This TV SOUND TUNER offers full UHF coverage with 5 pre-selected tuning controls. It can also be used in conjunct

\section*{ion with your video recorder. Dimensions: \(11 \frac{3}{4^{\prime \prime}} \times 8 \frac{1}{2^{\prime \prime}} \times 31 / 4^{\prime \prime}\) \\ 5.1.e \(+£ 2.00 \mathrm{pRp}\). ion with your video}

\section*{PRACTICAL ELECTRONICS STEREO CASSETTE RECORDER}


\section*{MONO MIXER AMP}

Ideal for Church
halls \(\&\) Club houses.
\(\mathbf{E 4 5 . 0 0}+£ 2.00 \mathrm{p} \& \mathrm{p}\) 50 WATT Six individually mixed inputs for two pick ups (Cer or mag.), two moving coill microphones and two auxiliary for tape tuner, organs, etc. Eight slider controls six for level and two for master bass and treble, four extra treble controls for mic. and aux. inputs. Size:
 (cont.) for use with


Matching AKG Microphone to suit (with speech and music filter). Complete with lead. ONLY \(\mathbf{f 9 . 9 5}\) plus 75 p p\&p

\section*{VHF STEREO TUNER KIT}


This easy to build 3 band stereo AM/FM tuner kit is designe in conjunction with Practical Electronics (July 'B1 issue). For ease of :onstruction and alignment it incorporates three Mullard modules and an I.C. IF. System.
FEATURES: VHF MW, LW Bands, interstation muting and AFC on VHF. Tuning meter. Two back printed PCB's. Read made chas s and scale. Aerial: AM - ferrite rod, FM - 75 or 300 ohms. statalised power supply with 'C' core mains trans former. Ai : nponents supplied.are to strict P.E. specificatcon, Fron! : die size: \(101 / 2^{\prime \prime} \times 21 / 2^{\prime \prime}\) approx. Complete with diagram and instructions.
*SPEじiAL OFFER! * Self assembly simulated wood £13.95 (w/out case) Finish size: \(1114^{\prime \prime} \times 81 /^{\prime \prime \prime} \times 314^{\prime \prime \prime}\) Plus \(£ 2.50\) p \& p. \(£ 3.50\) Plus \(£ 1.50 \mathrm{p}\) \&

\section*{HI-FI TWEETER BARGAIN}

GOODMANS TWEETERS 8 ohm soft dome radiator, tweeter (37. sq.) for use in up to 40 W £3.95 each ( \(\mathrm{p} \& \mathrm{p} \mathrm{E} 1\) ) or
\(\mathbf{~} \mathbf{6 . 9 5}\) pair ( \(p \& p \mathrm{E} 1.50\) )


\section*{AUDAX 8" SPEAKER}

HIGH QUALITY 40 WATTS RMS BASS/MIDRANGE

date their products without notice. All enquiries send S.A.E.

ALL CALLERS TO: 323 EDGWARE ROAD LONDON W2. Telephone: 01-723 8432. ( 5 minutes walk from Edgware Road Tube Station) Now open 6 days a week \(9-6\). Prices include VAT

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21A HIGH STREET, ACTON W3 6NG
Note: Goods despatched to U.K. postal addresses only All items subject to availability. Prices correct at 30/7/83 and subject to change without notice. Please allow 14 working days from receipt of order for despatch. RTVC Limited reserve the right to up

\section*{SPEAKER \\ KIT \\ 2 WAY 10 WATT}
\(8^{\prime \prime}\) bass/mid range and \(3 / \mathrm{h}^{\prime}\) weeter. Complete with screw wire, crossover components far pre-mitred D \(Y\) form Finish - chipboard covered wood simulate size \(141_{2}{ }^{\prime \prime} x\) \(3 \mu^{\prime \prime} \times 4^{\prime \prime}\). PAIR for ONLY
£ 12.50 plus \(£ 1.75\) p\&p.


\section*{125W HIGH POWER AMP MODULES \\ £10.50 £14.25}

\section*{\(+£ 1.15 p \& p+£ 1.15 p \& p\)}

The power amp kit is a module for high power applications disco units, guitar amplifiers, public address systems and even high power domestic systems. The unit is protected against short circuiting of the load and is safe in an open ir cuit condition. A large safety margin exists by use of gener ously rated components, result, a high powered rugged unit The PC board is back printed, etched and ready to dril ease of construction and the aluminium chassis is preformed and ready to use.
Supplied with all parts, circuit diagrams and instructions.
ACCESSORY: Stereo/mono mains power supply kit with transformer. \(£ 10.50\) plus \(£ 2.00\) p\&p.


SPECIFICATIONS
Max. output power (RMS): 125W.
Operating voltage (DC): 50-80 max
Loads: 4 - 16 ohms.
Frequency response measured @ 100 watts: \(25 \mathrm{~Hz}-20 \mathrm{KHz}\). Sensitivity for 100 watts: \(400 \mathrm{mV} @ 47 \mathrm{~K}\). Typical T.H.D. @ 50 watts, 4 ohms: \(0.1 \%\).
Dimensions: \(205 \times 90\) and \(190 \times 36 \mathrm{~mm}\).

\section*{StEREO}

CASSETTE DECK

Stereo cassette tape deck transport with electronics Manufacturer's surplus brand new and operational sold without warranty.

\section*{£11.95}
plus \(£ 2.50\) p\&p
Just requires mains transformer and input/output ockets and a volum

Supplied with full connection

details

\section*{AUDAX 40W FERRO-FLUID HI-FI TWEETER}

X/over on 5 kHz
\(22 \mathrm{kHz} .60 \mathrm{~mm} \quad € 550\)
square. 8 ohm . +60 p \& p

Telephone or mail orders by ACCESS welcome.


| N "O" and " \(A\) " level physics it is often necessary to demonstrate the use of a simple pendulum for determining the acceleration due to gravity. Normal methods, using a retort stand and clamp, cotton suspension, metal "bob", and a stopwatch for timing, are both tedious and time consuming.

This project will enable the physics teacher to set-up and easily demonstrate the principle of a simple pendulum, whilst removing a large proportion of the time consuming element. Using the apparatus it is possible to show that the period of the pendulum is directly proportional to the square root of its length and that the period remains constant, irrespective of the amplitude of the swing, providing amplitude does not exceed 10 degrees of arc.

It may also be shown that by increasing the number of timed swings increases the accuracy of the result and that a moving body will come to rest due to air resistance unless an external force acts upon it.

\section*{UNIT DESCRIPTION}

A mild steel bob and cotton suspension pendulum are kept swinging at a constant amplitude by an electromagnet sited below the pendulum. Every time the bob passes over the apparatus the electromagnet gives a kick to the pendulum inputting a small amount of energy. The amount of energy is precisely controlled by the circuitry and a constant amplitude of swing results.


A BBC Microcomputer is linked to the pendulum by a simple interface. A pulse from the interface is fed into the user port of the computer every swing of the pendulum. The computer counts the swings and works out the period. Then, using the formula:
\[
g=\frac{(2 \pi)^{2} 1}{t^{2}}
\]
the computer calculates a value for " g "

\section*{COMPONENTS}

Resistors
\begin{tabular}{ll} 
R1, 11, 12,13 & \(4.7 \mathrm{k} \Omega\) (4 off) \\
\(\mathrm{R} 2,3,5\) & \(1.2 \mathrm{k} \Omega\) (3 off) \\
R4 & \(2.7 \mathrm{k} \Omega\) \\
R6 & \(1 \mathrm{k} \Omega\) \\
R7,8,14 & \(1.0 \mathrm{k} \Omega\) (3 off) \\
R9,10 & \(68 \mathrm{k} \Omega(2\) off) \\
R15 & \(330 \Omega\) \\
All \(\frac{1}{4} \mathrm{~W}\) carbon \(\pm 5 \%\) See
\end{tabular}

Capacitors
C1.2 \(0.01 \mu \mathrm{~F}\) polyester (2 off)
C3 \(1000 \mu \mathrm{~F} 40 \mathrm{~V}\) elect. axial leads page 629
C4 \(\quad 0.047 \mu \mathrm{~F}\) polyester

Semiconductors
D1
D2
D2
TR1,2,4,6,7 2 N3904 silicon
TR3
TR5
IC1

\section*{Miscellaneous}

S1,2
VR1

VR2
PCC1

SK 1,2

SK3
SK4,5
s.p.s.t. toggle
(2 off)
\(100 \mathrm{k} \Omega \mathrm{lin}\).
contrak potentiometer \(1 \mathrm{M} \Omega\) lin. controi potentiometer ORP1 2 light dependent resistor 4 mm banana socket red (2 off)
4 mm banana socket white 4 mm banana socket black (2 off)

Printed circuit board: singlesided size, \(170 \times 95 \mathrm{~mm}, E E P C B\) Service Order code 8310-03; case, \(193 \times 158 \times 58\) (rear) \(\times\) 34 mm (front), (RS type 508-188); 14 -pin di.I. holder; 30 mm of \(12: 8 \mathrm{~mm}\) internal diameter mild steel tubing; \(7 / 0.2 \mathrm{~mm}\) wire; 36 s.w.g. enamelled copper wire.

\section*{Approx. cost \\ Guidance only}
£19.00
the acceleration due to gravity.
Where \(1=\) length (metres) of the pendulum and \(t=\) the periodic time in seconds.

\section*{CIRCUIT DESCRIPTION}

The potentiometer VR1 and the lignt dependent resistor (PCC1) form a potential divider at the base of TR1. As the pendulum bob passes over the l.d.r. the voltage at the base of TR1 increases.

Transistors TR1-TR3 form a Schmitt trigger which provides a good squarewave pulse for the bistable and VR1 is used to compensate the changing light conditions found in the laboratory. The circuit diagram is shown in Fig. 1.

One half of the 4001 , a quad 2 -input NOR gate (ICla and IC1b) forms a basic set/reset bistable latch. Resistors R7-R10 and C1, C2 form the necessary a.c. coupling to enable triggering by the pulses received from the Schmitt trigger.

The other two gates in the 4001 (IC.1c and IC1d) package are used in a monostable circuit which is triggered by the bistable. The monostable will stay on for a set amount of time determined by the values of VR2 and C4. VR2 is mounted on the front panel of the unit and enables the pulse width to be controlled.

The output from the monostable drives a Darlington pair amplifier (TR4 and TR5). The Darlington pair energises the electromagnet (Ll) and gives the bob a

Fig. 1. The complete circuit diagram for the Electronic Pendulum.

sufficient "kick" to maintain a constant amplitude. Switch SI is incorporated so that the electromagnet can be disconnected to allow free swinging of the pendulum.

The input to the interface circuit (TR6) is connected to the output of the Schmitt trigger, thus TR6 is switched on every time the pendulum bob passes over the l.d.r. The 5 -volt supply for TR6 (SK2) is derived from the computer power supply available at the user port.

For connection to the user port consult the BBC user guide. Note that if a separate power supply is used for TR6 it must not exceed 5 V as damage to the computer may result. The output from TR6 (SK3) is connected to \(\mathrm{PB} \emptyset\) of the user port.

\section*{CONSTRUCTION}

The electronic circuitry is built into an RS Components console case measuring \(193 \times 158 \times 58\) (rear) \(\times 34 \mathrm{~mm}\) (front), although any case of approximately this size should be suitable. VR1, VR2, the light emitting diode (D2) and the electromagnet assembly are mounted on top of the case. The l.d.r. is mounted inside the core of the electromagnet as shown in Fig. 2 and all the connections to the console are made via 4 mm banana sockets mounted at the rear of the case.

\section*{ELECTROMAGNET}

The core of the magnet is made from 12.8 mm (internal diameter) mild steel tubing which will accept the l.d.r. as a comfortable fit. The coil of the electromagnet uses 800 turns of 36 s.w.g. enamelled copper wire.

The components are mounted on a printed circuit board (shown in Fig. 3) measuring \(170 \times 95 \mathrm{~mm}\). This is large for such a simple circuit and could be reduced in size considerably. One advantage of such a large board is that various parts of the circuit can be easily identified and fault finding is made simple.

The board is available from the \(E E\) Printed Circuit Boärd Service, Order code 8310-03.

\section*{PENDULUM}

The bob is made of mild steel (soft iron would be better) and measures 30 mm in length by 20 mm in diameter. A piece of black card is attached to the bottom of the bob and is approximately 35 mm in diameter. This disc of card is essential for the operation of the system since it causes the electromagnet to be energised just before the bob reaches the electromagnet as shown in Fig. 4. This gives the necessary kick to the pendulum to keep it swinging.

The support for the pendulum is not critical and is left to personal choice.

The unit is easy to use with the program listed. Set the bob swinging and adjust VR1 until D2 is reliably pulsed on every time the bob passes over PCC1. Adjust VR2 until the amplitude of the pendulum is less than 10 degrees of arc.

This will require patience as the bob will react slowly to pulse width changes. Once the amplitude is correct VR2 will need very little adjustment unless the length of the suspension is changed dramatically. One setting will work for suspension lengths between 0.5 and 1 m .

\section*{OPERATION}

Run the program and enter the length of the pendulum, and remember, the length of the pendulum is measured from the centre of the bob to the pivot. When the computer is counting the swings, the elapsed time will be displayed in the centre of the visual display unit. Fairly accurate results are obtained with ten swings although accuracy will increase by increasing the number of swings to be timed.

Although this project is used with a BBC Microcomputer it could be used with virtually any microcomputer with a user port. The program listing given is very simple and could be easily adapted to other machines.

\section*{ELECTRONIC PENDULUM SOFTWARE}
```

                                    2 REM Electronic Pendulum
                                    4 REM October 1983
                                    6 REM <c> Everyday Electronics
    10CLS
20 MODE 5
3OPRINT TAB(1.10);'ELE CT R ON N C"
4OPRINT TAB(3,15):"P E N DULUM
45 LET A$=GET$
60 MODE 7
70 CLS
75 PRINT:PRINT
80 PRINT" PROGRAM TO DETERMINE g. ACCELERATION
9 0 ~ P R I N T : P R I N T ~
100 INPUT "INPUT THE ĽENGTH OF THE PENDULUM IN
110 PRINT:INPUT "HOW MANY SWINGS DO YOU REQUIRF",K
1.20 PRINT:PRINT:PRINT`PRESS ANY KEY TO CONTINUE
130 ?\& FE62=0
140 A=-1
150 AS=GET\$
160 CLS
170 IF ?\&FE60=254 THEN 180 ELSE }17
180TMME =0
190 T=TIME
200 PRINT TAB(14,12);(TIME-T)/100
210X=?\&FE60
220IF X=254THEN240
230 GOTO 200
240A=A+1
250 IF A = K*2 THEN 280
260IF?\&FE60 =255THEN230
270GOTO250
280T 1 =(TIME-T)/100
290P=T1/(A/2)
300G=((2*PI) 2)* U/P
305 LET G=INT(G*100)/100
310 PRINT"LENGTH OF PENDULUM=";L "m."
312 PRINT''NUMBER OF SWINGS = ";K
314 PRINT:PRINT"'q.ACCELERATION DUE TO GRAVITY="';G:" m/s}\mp@subsup{}{}{\wedge}
320 PRINT:PRINT:PRINT"DO YOU WISH TO RUN AGAIN <Y OR N>
330 IF GET\$ = 'Y" THEN GOTO 70
340 END

```

100 INPUT "INPUT THE LENGTH OF THE PENDULUM IN
110 PRINT:INPUT "HOW MANY SWINGS DO YOU REQUIRE", K
20 PRINT:PRINT:PRINT'PRESS ANY KEY TO CONTINUE
30 ? \& FE62 =
\(140 A=-1\)
\(150 \mathrm{AS}=\mathrm{GET} \$\)
160 CLS
\(180 \mathrm{TM} \mathrm{ME}=0\)
\(190 \mathrm{~T}=\mathrm{TIME}\)
200 PRINT TAB (14, 12);(TIME-T)/100
\(220 \mathrm{IF} \mathrm{X}=254\) THEN 240
230 GOTO 200
240A \(=A+1\)
250 IF \(A=K * 2\) THEN 280
260 IF \(3 \&\) FE60 \(=255\) THEN 230
OTO250
-T//100
\(290 \mathrm{P}=\mathrm{T} 1 /(\mathrm{A} / 2)\)
\(300 \mathrm{G}=\left((2 * \mathrm{PI})^{2}\right)^{\circ} \cup \mathrm{P}^{2}\)
305 LET G \(=1 \mathrm{NT}(\mathrm{G} * 100) / 100\)
312 PRINT
MBER OF SWINGS \(=\) ' \(\cdot K\)

330 IF GET \(\$=\) " \(Y\) " THEN GOTO 70
340 END

DUE TO GRAVITY

Fig. 4. Black card disc shown mounted on the mild steel bob.

METRES", L


\title{
COMPUTER AIDED EXPERIMENTS
}

\section*{USING THE RM380Z MICROCOMPUTER}

\author{
BY A. A. CHANERLEY B.Sc. M.Sc.
}


\section*{2. NAPHTHALENE COOLING CHARACTERISTIC}

THIS experiment uses the 590 kH temperature sensor with its associated amplifier (see page 646) to feed an analogue-to-digital converter. The latter is connected to the user port of a 380 Z microcomputer which allows the results to be stored and processed to produce a plot in real-time of the cooling curves of various substances.

In the following, naphthalene is the substance under investigation, and the ADC employed is the one featured in EE Sept '83.

Naphthalene is heated to 100 degrees C and the liquid naphthalene is then allowed to solidify and cool down to room temperature. The temperature transducer produces an amplified voltage output proportional to temperature in degrees C , and this analogue voltage is input directly to channel 0 of the ADC. Here the voltage is converted into an 8 -bit binary format (byte), which the microcomputer continuously monitors, and plots on the vDu the temperature as a function of time, that is, a cooling curve. In effect, the microcomputer is acting as a voltmeter, which, at the same time plots the voltage variations across it, just as in last month's laser experiments, where the varying voltage was produced as a result of variations in light levels.

\section*{EXPERIMENT SET-UP}

The experimental set up is shown in Fig. 2.1, and the resulting cooling-curve for naphthalene in Fig. 2.2.

The solid naphthalene is placed in a test tube which in turn is placed in a beaker of boiling water. After a few minutes the naphthalene melts and liquefies. Place the temperature sensor assembly in the 100 degree \(C\) liquid naphthalene and RUN the software on the RML 380 Z microcomputer. The top

Fig. 2.1. Experimental set up for determining the cooling characteristics of naphthalene. The A-to-D Converter for interfacing with the 3802 Machine was described last month.

right of the vDu should show a decimal number around 200. Our conversion equation is:
\[
V=k N
\]
\(k=(10 / 255)\), and at 100 degrees C the value of the output voltage (V) should be 8 volts from our amplifier, therefore
\[
N=(V / k)=(8 \times 255 / 10)=204
\]

Lift out the test tube from the boiling water and allow the naphthalene to cool. As the temperature drops, the microcomputer will start to plot the variation of temperature with time. The curve will initially exhibit a steep drop, due to a large rate of heat loss, but then the curve will be seen to level out to a plateau as the naphthalene begins to solidify, at about 80 degrees \(C\), and it will stay level until all the naphthalene is solid again, when the temperature will start to drop again and the curve will fall but much more gradually than before.

Best results were obtained with fresh samples of naphthalene. Candle wax was also used, but since this is a mixture of several waxes, each one solidified in turn, so the "plateau" had a definite gradient!

Cooling curves are extensively used in the steel industry as a means of determining the purity of molten metals.

The combination of Temperature Transducer/Amplifier/ADC/380Z can be put to less esoteric (but just as important) applications, as the measurement of one's own body temperature, with perhaps the microcomputer responding with a "go and see a doctor" diagnosis?!

Or, at a particular setting of temperature the microcomputer can activate an aux output on the ADC unit with an appropriate POKE statement. For example if AUXI is connected to an alarm circuit and if the temperature exceeds 50



Fig. 2.2. Photograph of the cooling curve for the experiment shown in Fig. 2.1.
degrees \(C\) then the chances are a fire is starting somewhere. On reading this, the computer then responds with a POKE 64511,128 , which makes pin 21 of the \(380 Z\) user port go high. This can be used to saturate a transistor, close a relay and turn on an alarm.

Similarly, last month's light sensor could, for example, be used as part of an intruder alarm system with the microcomputer monitoring the voltage output from the light transducer, which if low, could again signal the activation of an alarm.

\section*{THE A-TO-D CONVERTER}

The ADC referred to is based around the 7581 analogue-to-digital converter
chip, having eight separate analogue inputs (channels) each one capable of handling up to 10 V maximum input.

Constructors need not confine themselves to this ADC, but can use this temperature transducer/amplifier with any ADC as long as its maximum input voltage is known so that the gain of the amplifier can be tailored to meet that maximum voltage. The same applies to last month's light sensor, where the maximum voltage would need to be divided in order to meet input requirements for other ADCs.

\section*{SOFTWARE}

The software is almost the same as for Experiment 1 published last month on


The temperature sensor used in this experiment. Full details of this sensor and associated amplifier can be found on page 646.

\title{
EJERYDAY ELECTRONICS SOFTWARE SERUICE
}

The EE Software Service provides an easy and reliable means of program entry for our computer-based projects. All programs have been tested by us and consist of two good quality copies of the working program on cassette tape. Certain program listings are also available.

All prices include VAT, postage and packing. Remittances should be sent to Everyday Electronics Software Service, Editorial Offices, King's Reach Tower, Stamford Street, London SE1 9LS. Cheques should be crossed and made payable to IPC Magazines Ltd.
\begin{tabular}{|l|c|c|c|c|}
\hline PROJECT TITLE & \begin{tabular}{c} 
CASSETTE \\
CODE
\end{tabular} & \begin{tabular}{c} 
CASSETTE \\
COST
\end{tabular} & \begin{tabular}{c} 
LISTING \\
CODE
\end{tabular} & \begin{tabular}{c} 
LISTING \\
COST
\end{tabular} \\
\hline \begin{tabular}{l} 
ZX81 SPEED COMPUTING SYSTEM \\
(Feb 83)
\end{tabular} & T001 & £2.95 & L001 & £2.95 \\
\begin{tabular}{l} 
REAL-TIME CLOCK (Apple) (May 83) \\
REAL-TIME CLOCK (BBC) (May 83) \\
EPROM PROGRAMMER (TRS-80 \& \\
GENIE) (June 83)*
\end{tabular} & T002 & T2.95 & L002 & \(£ 2.95\) \\
\begin{tabular}{l} 
STORAGE SCOPE INTERFACE (BBC) \\
(Aug 83)
\end{tabular} & T004 & \(£ 3.95\) & N/A & - \\
\hline
\end{tabular}

\footnotetext{
*'Includes Command List with examples.
}


\section*{SIGNAL INJECTOR}

ASMALL pocket-sized signal injector, designed as a penshaped signal probe, has just been marketed in the UK by Alcon Instruments Ltd.


The Chinaglia Usijet incorporates a blocking oscillator as the main signal generator, giving a basic 500 kHz signal which is modulated at 1 kHz for identification and demodulation check purposes. Because of the waveform used the equipment produces harmonics detectable up to 500 MHz very useful for audio, radio and television servicing applications.

In use the probe case is merely connected via a fly-lead to the earth line of the item under test and the probe tip touched to whatever point the signal is required at. The probe can be used in "live" test conditions and it can also cope with circuit voltages up to 500 V d.c.
Power consumption is 25 mA from an internal 1.5 V battery to give a 20 V peak-to-peak output at the probe tip. The price of the Chinaglia Usijet Signal Injector, complete with earthing lead and instructions, is \(£ 11.55\) inc. VAT.

> Alcon Instruments Ltd.,
> Dept EE, 19 Mulberry Walk,
> London, SW3 6DZ.

\section*{PORTABLE SOUND AND VISION}

The latest in sound and vision is a portable stereo cassette recorder, m.w./f.m. radio incorporating a black and white TV from Heron Electronics, and being marketed under the Ingersoll brand.
The Ingersoll XK 500 features a soft touch cassette deck with auto stop and soft eject, slide volume control, two built-in condenser microphones and a telescopic antenna. The TV has the usual brightness, contrast and vertical hold controls and a \(4 \frac{1}{2}\)

\section*{PSEUDO-STEREO}

THE first i.c. to combine stereo/pseudo-stereo/spatial stereo sound circuits is now being marketed by Mullard. Primarily for use in top-of-the-range TV and portable stereo radios, it can also be used in low-cost mono radios, where it makes possible the reproduction of the stereo effect without the need for expensive stereo decoder circuits.

The possibilities for stereo sound reproduction in the latest batch of TV sets which are coming on the market are, at present, being under-utilised because there are so few stereo TV broadcasts. The TDA3810 allows mono music to be converted into realistic-sounding pseudo stereo.

Basically, the pseudo-stereo circuit in the TDA3810 splits the incoming mono signal into two. One channel is presented straight to the output, while a delay is introduced into all frequencies of the second channel from 300 Hz to 2 kHz . The value of this delay depends on frequency, and is for example \(500 \mu \mathrm{~s}\) at 800 Hz ; this gives the illusion of stereo sound.


The frequencies of the second channel below 300 Hz and above 2 kHz are taken straight to the output to "fill-out" the sound between the speakers. The effect is, however, subjective, and lowpass filtering has been kept off the chip to allow manufacturers to customise the circuit.

The TDA 3810 i.c. gives automatic switching between stereo and mono, with manual switching between stereo/spatialstereo sound and mono/pseudostereo sound. It has outputs to directly drive l.e.d.s to indicate "spatial stereo" or "pseudostereo". The circuit operates from a supply voltage of 4.5 V to 16 V and has a supply current of typically 7 mA .

Mullard Ltd.,
Dept EE, Mullard House,
Torrington Place,
London, WCIE 7HD.

\section*{EXPRESS HANDLES}

THE latest edition of ImhofBedco Express Service catalogue contains a selection of enclosure handles.

Over 30 different handle styles are featured, many of them in a variety of sizes and fixings. Materials available include brass, chrome, steel, aluminium and plastics.

Styles range from simple bar handles to locking sets which come complete with mounting kit and keys.

Copies of the Imhof-Bedco Express Catalogue and addresses of nearest stockists are available from:
Imhof-Bedco Standard Products Ltd., Dept EE, Ashley Works, Ashley Road, Uxbridge, Middlesex UB8 2SQ.
inch screen. Loudspeakers are included at each end of the unit.

The approximate retail price of the XK500 is \(£ 150\) including VAT, and details of local
stockists can be obtained from:
Heron Electronics Lid.,
Dept EE, Heron House, 19 Marylebone Road, London, NWI SJL.


GOOD CONNECTION

Aminiature, low voltage 2 -pin connector set, originally designed for connecting batteries to Police Radios is now being marketed by A. F. Bulgin \& Co. P.L.C.

The flex connector P750/2M has 2 metres of black 2 -core cable terminated at one end with a moulded "female" plug. The
chassis mounting inlet P755 is fitted with solder tags and fits a panel hole of approximately \(9 \times\) 5 mm . The connector is rated at 3 A 100 V maximum.
Full details and stockists from:
A. F. Bulgin \& Co. PLL.C.
Dept EE, Bypass Road,
Barking, Essex IG1I OAZ.


\section*{POWER SUPPLIES}

Arange of low-cost d.c. power supplies suitable for powering logic, microprocessor and linear circuits are being produced by Electronart Design.

All units are housed in a 13 A plug style case and feature thermal fuse protection and l.e.d. output indicators. The stabilised output is by way of 4 mm sockets.

Models are available with single or twin outputs of \(5,12,15\) or 18 V d.c. Other "extras" such as crowbar over-voltage protection, Power-on Reset and Power Down signals for microprocessor and memory applications are available.

Further information and prices are available from:

Electronart Design,
Dept EE, 78 Kimberley Avenue,
Iford, Essex IG2 7AS.

\section*{HOME CONTROL SYSTEM}

The new Aquarius home computer system from Mattel Electronics, has been specifically designed to have practical applications in the house. "Research has indicated that consumers are concerned that if they buy a computer it would be out-dated within three months of purchase," said Michael Lunch, Managing Director of Mattel UK.

In addition to the recent launch of the Aquarius home computer (a Special Report will be published shortly), they also demonstrated the prototype of a unique new home control system which is to be launched on the UK market very shortly.

The Home Control System (HCS) simply plugs into the Aquarius via one lead and into the mains via another. Using the computer you can program the system to control lights, electrical appliances, central heating, airconditioning and any other electrically operated mains powered device.

\section*{The System}

The system uses the mains wiring in your house, so no extra
cabling or installation is required. Each selected light or appliance can be operated by the Home Control System by simply plugging it into a special "coding pod" which in turn plugs into standard 3-pin sockets. Equally, other modules can be attached to wall switches or thermostats, rather like adding a dimmer switch.

The HCS sends coded signals down the mains which are then received by the special modules which in turn activate the electrical appliances. Up to 256 separate items can be controlled and programmed in advance for a series of up to 32 individual on/off events in seven-day cycles.

Programming the system is made easy by the use of graphics indicating the type of room in the house and positions of appliances in the room. Only three keys on the Aquarius keyboard have to be used to carry out programmingfrom identification of light/ appliances to setting the on/off timings and entering the final program.

\section*{How It Works}

The system works by giving each light or appliance module a code number and then sending signals automatically, via the
mains, at the time you have programmed. The module then activates the appliance to turn on or off, in addition, the system is capable of dimming any light to a preset intensity.

For example, it is simple to set a porch light to come on at \(8.30 \mathrm{p} . \mathrm{m}\). and off again at 12.30 . This operation could be set either on a regular daily basis, separately timed for each day or in "security" mode. Security means that the computer will select random times to switch on or off at half an hour either side
of the time input-a useful feature should you wish to deter burglars whilst you are on holiday.

The system could also be used to perform a "close down" function in the home or in a small office to make sure that items such as television, electric heaters, lights, and other appliances are not left on overnight.

Mattel Electronics UK Lid, Dept EE, Mattel House, North End Road, Wembley,
Middlesex HA9 0AB.


\section*{CIRCUTI EXCHANGE}

\section*{OUTSIDE LIGHT SWITCH}

THIS circuit is for an Outside Light Switch, which can be activated by a car headlight beam. The relay is rated at 12 V . The circuit is activated by S1 (push-to-make) and switched off by S2 (push-to-break).

Robin J. Clements, Sheffield.


\section*{REACTION METER}
/hen the Reaction Meter is switched on each l.e.d. lights up in turn, the frequency at which each l.e.d. lights up can be varied by altering VR1. When S2 is depressed the l.e.d. which was on when S2 was depressed remains on until S2 is released. The idea of the reaction meter is to depress S2 when D6 the only green l.e.d. is lit so that D6 is the only l.e.d. alight. Altering VR1 makes it easier or harder.

IC1 acts as a voltage controlled oscillator. The frequency of the pulses depends on VR1 which acts as a potential divider. As IC2 receives each pulse the output is moved along one so the I.e.d.s light in turn. When S2 is depressed the "diode inhibit" (pin 13) goes high every time a pulse is produced, so IC2 stops counting and the l.e.d., which was lit when S2 is depressed, remains lit when S2 is released. Pin 13 is kept low by R3.
D. J. Sparkes, Green Lane, Coventry


\section*{ELECTRONIC TIMER}

THIS is an Electronic Timer with a timing range of one to eleven minutes. The lower half of IC 1 is used as a standard monostable triggered by S 1 . This output is inverted by IC2 and fed to the reset pin (11) of the astable constructed from the other half of IC1.

The output from this astable is modulated by a further astable in IC2 and this produces a "whooping" note in the loudspeaker LS1.

Paul Dellar,
Harpenden,
Herts.


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We are sorry to announce the cancellation of this year's Electronic Hobbies Fair, planned for 27-30 October.

In spite of a significant success last year, the continuing recession is hitting the electronics hobby industry pretty hard. This has meant that many companies feel that this year they cannot sensibly allocate the resources of time, money and manpower involved in participation in exhibitions.
We feel that any exhibition sponsored by Everyday Electronics must offer the visitor a full range of
components, equipment, projects and demonstrations from a wide selection of companies across the industry. As we cannot be absolutely sure of doing just this, we have decided with regret that we must disappoint our readers now rather than in October.

Everyday Electronics would like to thank those companies who had already given their support to the Electronic Hobbies Fair this year. With our apologies for the disruption of their plans we couple our hopes for a future event in a more buoyant business climate.

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12 volt MOTOR BY SMITHS wound and they become more powerful as load increases. Size Price \(£ 3.45\)
Ditto, but permanented \(£ 4.25\).


EXTRA POWERFUL 12v MOTOR
Made to work battery lawnmower, this probably develops up to compressor etc. etc. \(£ 7.95+£ 1.50\) post.
(This is easily reversible with our reversing switch - Price \(£ 1.15\) ): MAINS MOTORS
We have very large stocks of motors from 2 watts to \(\frac{1 / \mathrm{hp} \text {. Most at }}{}\) 67 R.P.M. Motor: \(1 / 10\) no. reversible mains operated split phase motor with gear box - shaft fitted with chain sprocket \(£ 11.50\) plus 100 R.P.M. Motor: \(1 / 6 \mathrm{hp}\). Mains driven reversible motor with
gear box, \(1 /\) " shaft from gear box - Very powerful \(£ 16.50\) plus \(£ 3\) gear box, \(1 / 2\) " shaft from gear box - Very powerful \(£ 16.50\) plus \(£ 3\)
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and 3500 kHz with wiring diagram. \(\mathrm{f} 1,15\).

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This ready assembled unit is the ideal tuner for a music
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personal stersor personal stereo radio give you superb reception.
Other uses are as a "get you to sleep radio"; you could even take it with you to use in the lounge when the rest of the
family want to view programmes in which family want to view programmes in which vou are
interested. You can listen to some music instead.
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medium wave band \(525-1650 \mathrm{KHz}, \mathrm{FM}\) band 87 108 MHz , mono, stereo \& AFC switchable, uning meter to give You spot on stereo tuning, optionat LED wave band
indicator, fully assembled and fully aligned. Full wiring indicator, filly assembled and tully aligned. Full wiring phones and details of suitable FM aerial ample ferrite rod
pher aerial is incluted for medium and
made up on very compact board.

\author{
\(\qquad\) \\ Offered at a fraction of its cost: : only \(£ 6.00\)
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VIEWDATA EQUIPMENT
ORACLE VB 100 PCB T
systems. including the Prestel Unit which we are currenty selling other components. It has a TV aerial input and a TV UHF modifie (AZTEC UM 1233). We offer this board, new, unused and complet except for 6 of the 25 I.C 's at \(£ 5.75\). The plug in holders for the MINIKEY SERIES KL This is an American made membran keyboard with silver contacts as used on Prestel to dial into the
British Telecom phone system. It is realiy miniature, only 60 mm British Telecom phone system. It is really miniature, only \(60 \mathrm{~mm} \times\)
\(65 \mathrm{~mm} \times 5 \mathrm{~mm}\) thick. It has 16 press buttons, giving stándard \(0-9\) numbers and ABCD facilities. There are two other buttons \(0-9\) TELEPHONE LINE TERMINATION UNIT As used with Prestel but undoubredly suitabie for other applications. Impor Reed Relavs. All mounted on a pcb with 1.C. and other component VOLTAGE STABILISED POWER SUPPLY As use 20 watt mains transformer. Rectifiers and semi-conductors all mounted on P.C. 8 . size approximately \(42 /{ }^{\prime \prime} \times 2\) ". The stabilised D
output from this is \(-27 v-12 v-0+0+12 v+27 v\). Price \(\mathbf{f 4 . 6 0 \text { . }}\). INSTRUMENT CASE As used with the Prestel unit this comprises an ali chassis and a moulded front plastic cover secured
to the chassis by self-tapping screws. Overall size approx 12 " \(\times 10^{\prime \prime}\) o the chassis by self-tapping screws. Overall size approx \(12 \times 10\)
\(\times 21 / 2^{\prime \prime}\) deep. On the front is fitted the minikeyboard as described above and at though originally intended for Prestel, this case-should have other uses including
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 housing with plug in lead comectises the \(x\)-ray zube in a radiation proo housing with plug in lead connectors. The tube enclosed in the
housing ts a hospital size tube and unused and new. Price 699 . housing 15 a hospital size tube and unused and new. Price 669 .
EHT TRANSFORMER \& RECTIFIER UNIT We estimate that output voltage of this is probably 30 or 40 KV . Completely enciosed in an oit filled container, size \(13^{\prime \prime} \times 14^{\prime \prime} \times 15^{\prime \prime}\). There are four
rectifier sections, each using 20 EHT rectifiers connected in series these plug in for ease of replacement. The unit is powered by a
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reatised that many of the American made tools intended for realised that many of the American made tools intended for
if used on building sites and simitar damp conditions must be
isolated from the mains for safery reasons, isolated from the mains for safery reasons, as in many cases the
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British made transformers at ver
Primary Secondary Current atractive prices.
 \(\begin{array}{lccccc}240 \mathrm{v}: & 4.5-0-4.5 \mathrm{v} & 400 \mathrm{~m} / \mathrm{a} & 50 \mathrm{p} & 45 \mathrm{p} & 35 \mathrm{p} \\ 240-6 \mathrm{v} & 100 \mathrm{~m} / \mathrm{a} & 58 \mathrm{p} & 52 \mathrm{p} & 43 \mathrm{p}\end{array}\) (Postage \& Packing: 45p per transformer or \(\mathrm{f1} .60\) per 10 f £4.50 per 100 ) These very high quality British made two pin European adaptors are ideal for driving Radios, cassette recorders, TV games, calculators etc. The adaptors fit in the UK shaver socket.

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Price f14.90 each \(+\mathbf{E 2 . 0 0} \mathrm{PGP}\)
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\(160 \times 175 \times 295 \mathrm{~mm}\) \(160 \times 175 \times 295 \mathrm{~mm}\)
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Price f13.90 each + f1.50 P \& P.


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system cabinet kit \(\mathbf{E 7 . 0 0}\) each \(+\mathbf{E 2 . 0 0 ~ P}\) G P

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DECK MODULE}
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Price \(£ 26.70+£ 2.50\) postage and packing
Price \(£ 26.70+£ 2.50\) postage and packing.
Supplementary parts for 18 V O.C. power supply
transformer, bridge rectifier and smoothing capacitor) (transformer, bridge rectifier and smoothing capacitor)

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\(15^{\circ \prime} 100\) watt R.M.S. Impedance 8 ohms. 50 oz magnet. \(2^{\prime \prime}\) aluminium voice coil. Res. Freq. 20 Hz . Freq. Resp. to 2.5 KHz . Sens.
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95 dB . Price: \(\varepsilon 24.50\) each \(+\varepsilon 3.00\) P\&P \({ }^{\prime \prime \prime} 50\) watt R.M.S. Impedance 8 ohms. 20 oz. magnet. \(1 / \sqrt{2}\) " aluminium voice coil. Res.
Freg. 40 Hz . Freq Reso to 6 KHz . Freq. 40 Hz . Freq. Resp to 6 KHz . Sens. 92 dB , Black Cone. Price: \(£ 9.50\) each. Also available with black protective grille Price


12' 85 watt R.M.S. McKENZIE C12̇85GP (LEAD GUITAR, KEYBOARD, DISCO) \(2^{\prime \prime}\) aluminium voice coil, aluminium centre dome, 8 ohm imp., Res. Freq. 45 Hz ., Freq Resp. 106.5 KHz . Sens, 98 dB . Price: \(£ 23.00+£ 3\) carriage.
12" 85 watt R.M.S. McKENZIE C1285TC (P.A., DISCO) \(2^{\prime \prime}\) aluminium voice coil. Twin cone. 80 hm . imp. Res. Freq. 45 Hz . Freq. Resp. to 14 KHz . Price \(£ 23\) + \(£ 3\) carriage. 15" 150 watt R.M.S. McKENZIE C15 (BASS GUITAR, P.A.) \(3^{\prime \prime}\) aluminium voice coil Die cast chassis. 8 ohm imp., Res. Freq. 40 Hz ., Freq. Resp. to 4 KHz . Price: \(\mathbf{£ 4 7}+£ 4\) carriage.

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*Projects for Book 8 were in an advanced state at the time of writing, but contents may change prior to publication (due 13th August 1983).

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