## Easy to build projects for everyone MAY 80 $50 p$ ancolioilc



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Choosing the products to advertise each month can be quite a task at AMBIT since we tend to introduce at least one new line per week. So it is nearly impossible to say all we would like in this space - other than to bring you as far up to date as possible with current events. The major medium for finding out about what we have to offer is our unique catalogue system, and we ask that you invest in a copy of parts 1,2 \& 3 since many questions we are asked can be readily answered by reference to these.
Each part costs 60 p, or $£ 1.60$ for all three current editions.
We are also launching a new and greatly elongated version of our PRICE LIST which now includes a large number of quantity listings, and many items not previously listed. The new style price list is a quick reference short form to our general catalogues - available FOC with a large (A4) SAE please.
As a result of the soaring price of oil - and the subsequent huge increases in the cost of wax for Mr Tom Jackson's famous moustache, the Post Office have increased their charges (Feb. 4th). Accordingly, our standard cover charge has been increased to 35 p per order (CWO).

## TOMPODEITF

## DIGITAL FREQUENCY READOUTS / SYNTHESISER SYSTEMS

Ambit has the biggest range of digital frequency readout systems for various applications in Broadcast and Communications. Prices range from f 18.50 for a complete AM/FM broadcast frequency display (kit of DFM2). Most are detailed in the latest catalogue.
TUNING SYNTHESIZERS are atso heavily featured, and we offer our first complete system covering MW/LW/ SW2 and FM based on Hitachi parts. The unit is retrofittable to yoltage tuned radio systems - and will shortly be incorp orated in a complete tuner project. Cost for the synthesiser will be circa $£ 40$ A versatile communications system based on the new Mullard 2 ic system is nearing completion, together with 16 station CMOS memory and optical shaft encoder system with fast tune facility. Synthesiser circa $\mathrm{E70}$, memory f 50.

## Latest semiconductor news:

CM OS. TTL and LPSN TTL are in stock lask for our OSTS price leaflet). Some of the very popular types ara still "difficult" but we have things like 4011s, 4017s at the time of writing.
BADIO 1Cs _--interesting developments here, we now have the Hitachi HA11225 and the HA12412 ultra high specification members of the CA3089E family. The PLESSEY SL 1600 range now includes the SL6600 high performance PLL NBFM IF and detector.

| CA3089E | 2.11 | HA1197 | 1.61 | SD6000 | 4.31 | SL 1610 | 1.84 | SL 1626 | 2.80 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| CA3189E | 2.53 | CA3123E | 1.61 | TDA4420 | 2.59 | SL 1611 | 1.84 | SL1630 | 186 | $\begin{array}{llllllllll}\text { CA3189E } & 2.53 & \text { CA3123E } & 1.61 & \text { TDA4420 } & 2.59 & \text { SLL } 1611 & 1.84 & \text { SL. } 1630 & 1.86 \\ \text { HA1137W } & 1.95 & \text { TDA1072 } & 3.09 & \text { MC1330P } & 1.38 & \text { SL1612 } & 1.84 & \text { SL1640 } & 2.17\end{array}$ $\begin{array}{llllllllll}\text { HA11225 } & 2.47 & \text { TBA651 } & 2.53 & \text { MC1350P } & 1.38 & \text { SL1613 } & 2.17 & \text { SL } 1641 & 2.17 \\ \text { HA12412 } & 2.81 & \text { TDA1090 } & 3.51 & \text { KB4412 } & 2.24 & \text { SL } 1620 & 2.50 & \text { SL6600 } & 4.31\end{array}$ $\begin{array}{lllllllllll}\text { HA12412 } & 2.81 & \text { TDA1090 } & 3.51 & \text { KB4412 } & 2.24 & \text { SL } 1620 & 2.50 & \text { SL6600 } & 4.31 \\ \text { KB4420 } & 1.95 & \text { TDA1220 } & 1.61 & \text { KE4413 } & 2.24 & \text { SL } 1623 & 2.80 & \text { SL6640 } & 3.16\end{array}$ $\begin{array}{llllllllll}\text { TBA120S } & 1.15 & \text { TDA1083 } & 2.24 & \text { KB4417 } & 2.53 & \text { SL } 1624 & 3.77 & \text { SL6690 } & 3.68 \\ \text { KBA408 } & 0.80 & \text { TDA1062 } & 2.24 & \text { MC3357P } & 3.16 & \text { SL } 1625 & 250 & \text { MC1496 } & 1.44\end{array}$

TRANSISTORS: New lower prices, wider range, large stocks. Also the woeld's lowest noise gudio devices (2SC2546E and 2SA1084E) firsi from AMBIT of course. Power MOSFETs \& all sorts of other devices. Our 3SKS1 MOSFET replaces the $408 \times$ X and 40673 farnilies. $\begin{array}{lllllllll}\text { BC237-89 } & 0.092 & 2 S C 1775 & 0.207 & 2 S A 1034 E & 0.368 & \text { BF } 256 & 0.437 & \text { BFY } 90 \\ \text { BC307-8-9 } & 0.093\end{array}$ $\begin{array}{lllllllll}\text { BC307-8-9 } & 0.092 & \text { 2SA872A } & 0.207 & \text { 2SC2547E } & 0.391 & \text { 2SK55 } & 0.368 & \text { BF224 }\end{array} 0.0 .253$ $\begin{array}{llllllll}\text { BC413-5 } & 0.115 & \text { 2SD666A } & 0.345 & \text { 2SA1085E } & 0.391 & \text { 2SK168 } & 0.402 \\ \text { BF274 } & 0.207 \\ \text { BD414.6 } & 0.12 \overline{2 S} & \text { 2SB646A } & 0.345 & 2 \text { SK133 } & 6.32 & 3 S K 51 & 0.62\end{array}$ BC5 $\begin{array}{lllllllll}8 C 546-556 & 0.138 & 2 S D 760 & 0.52 & 2 S J 48 & 6.32 & 3 S K 60 & 0.667 & \text { VN66AF } \\ 1.092 \\ \text { BC550.560 } & 0.138 & 2 S B 720 & 0.52 & \text { 2SK135 } & 729 & \text { BF960 } & 1.426 & 2 \text { N4427 }\end{array} 0.977$

RADIO CONTROL: A special section for all RC fans. New and exciting stuff. KS4445/K84446 : complete 4 channel RX/TX dig-prop IC pair RF\&control in one 4.75 pr MSE $3362 /$ MSL9363 : logic section of a four channet dig,prop link, with switch opt. 3.75 pr NE5044: Signatics varsatile 7 channel ancoder, suitable for mixing ett, $£ 2.14$ ea NE544 Signetics famous sarvo driver IC $£ 2.07$ MC3357P as used in RCME design $£ 3.16$ ea
AMBIT RCRX4 - RCME FM sybitem compstible AMBIT RCRX4 - RCME FM system compatible, complete RX kit with box/connector XTAES: FM pairs $£ 3.74$ (no front end with 27 MHz coramic filter $£ 16.10$ (kit XTAES: FM pairs £3.74 (no splits) TX is fund. // op frequency, RX 3rd OT- 455 kHz
AM pairs $£ 3.57$ (no splits. Both 3rd OT types, again RX IF at 455 kHz

The new MK III FM tuner sitting under the Dorcheste multiband AM/FM tuner

Revisions to the Munk 11 include a centre zern tuning indicator meter and sitent preset pretet
switahing

New 944378-2, the last word in stereo decoders with the KB4437/4438.

## MODULE NEWS

We are at last able to quote for quantities of our modules, following a program of standardization and revision to speed manufacture and test. The following types are the results of the standardization program:
UM1181 5 varicap MOSFET input VHF band 2 turethead $E 12.00 \mathrm{mc}$
911225 A High Performance FM IF system, with switched BW 911225 B Single BW filters, single tuned detecior
91072 A DC tuned and single pole switched MW LW tune
91072 B As type : $A^{2}$ but with either SW1 or SW2 band
910242 A As type A but with either SW1 or SW2 band
92242 A Combined LW/MW tuner, with FM IF.detector section
92242 A Combined LW/MW tuner, with FM IF detec
92242 B As 92242 A but with $5 \cdot 10 \mathrm{MHz}$ SW section
$E 2395 \mathrm{inc}$
E14.95 inc
$E 14.35$ inc
E14.43 ine
E 14.43 inc
E 15.90 inc
92242 B As 92242 A but with $5 \cdot 10 \mathrm{MHz}$ SW section $£ 34.00$
All are supplied housed in screened metal cases $97 \times 56 \times 24 \mathrm{~mm}$, with all connections along a single edge, suitable for verticle or horizontal mounting.
Previously advertized units are still availatie. although there may have been some price changes in the latest edition of the Price List (Date Feh. 80 ). A separath leaflet covering the new range of modules is available from April 80 . with an A4 SAE please.

NEW LINE: ALPS switches and rotary potentiometers. With a general catatogue that's over 3 inches thick, we cannot begin to offer a comprahenive list of what we can offer - but we ate already stocking the keyboard, switches, keyswitches, pushbutton switthes etc. In particula, the pushbutton switches really put all others in the shade (schadow? when it comes to quality and
price. A special new shortform is being prepared fand may be ready when you read this) All the potentiometers and switchos you could ever need from a singie source. Keypad switches coast as littlo as 15 p ea (1 off). with e range of two part caps for easy ledgending. You must see the shortform catalogue (30p) and our new pricalist for full details of this huge range of component


## AMBIT SHOP NOW OPEN

We are graduatly petting our caller sales area sorted out.
with displays of the products on offer and a browsers corner to sit and study data/catalogues. Call in next time
you are in the area - parking autide the dor you are in the area - parking outside the door COMPUTER CAPABILITIES
Ambit has been keeping a low profile on the subiect of the MPU and its applications Interestingly enough, the
first proiect we offer with MPU content does rather first proiect we offer with MPU content does rather more in the way of processing than simply plaving a daft parne, or looking like an enormous catculator. Ou
MPU facility and Memmeriaty and expertise is now for hire on
comis. $Z 88$. 6800 , 6809,2650 etc.

NEW LINE: OC/DC+AC converters for fluorescent displays. TOKO CPS series
12 V IN, -20 and $3 V \mathrm{AC}$ out at 65 mA . Thick film desion $£ 2.34$ ea Otv, prices $O A$


## GENERAL INFORMATION

 Ambir stocks the following ranges of components for ex stock volume defivery SIGNAL COILS, CERAMIC,MECHANICAL and CRYSTAL FILTERS. RADIO ICS for AM/ FM/SSB. TOROID CORES FOR RADIO and EMI FILTER CIRCUITS, INDICATING AND PANEL METERS, AUDIO ICS, RF TRANSISTORS, FETS MOSFETS, DIODES (PIN, VARICAR SCHOTTKY), PASSIVE DBMM (thke MOIOR etc), IC SOCKETS. SWITCHES, TUNERHEADS IF AMPS AM RADIO MODULES, TUC ITC.
MEAS, IF AMPS, AM RADIO

NEW LINE : DVM176 - the definitive ICM7106 LCD DVM module, 3', digt f22.37 ea.
CM161: LCD $12 / 24 \mathrm{hr}$ alarm clock/day/dare/backlight (eq.RS308.499) 7 mm digits E 11.44 each CM174: LCD 12 hr atarm dock/stopwatch/backlight with 30 mm height digit $\quad \mathrm{E} 14.32$ each


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CABLE \& PIPE LOCATOR. Mar. 80. K3-40 less coil former.
KITCHEN TIMER. Mär, $80, ~ £ 22.46$. DOORBELL REGISTER. Mar. 80 . $\mathrm{E} 3-39$. TOUCH SWITCH. Mar. 80. EA. 65 LIFIER.
STERE HEADPHONE AMPL Mar, 80. 514 .94.
5 RANGE CURRENT LIMITER. Mar. 80 E4I24.
MICRO MUSIC 8OX. Feb. 80 £13. 82. Grey Case
 80, 240:46. MORSE PRATICE OSCILLATOR. Feb. 80. E3.75. ${ }^{\text {INBD }}$ BURGLAR ALARM. DEC. 79. 54.95 .

LIGHTCALL, Dec, 79.58 .30,
BABY ALARM. Nov, 7988.20
BAEY ALARMA. Nov. 79 £8-20
OPTO ALARM. Nov. 78 E5. $\mathbf{7 7}$ inc.
ODtional ports. optional ports.
MW /LW RADIO TUNER. Nov. $79815-50$ less dial.
3. FUNCTION GENERATOR. Nov. 79 ONE ARMED BANDIT. Oct 79. £18-39. case extra $23-88$.
HIGH IMPEDANCE VOLTMETER. Oct. 79. Ef5:87.
LGHTS ON REMINDER. Oct. 79. 84 -85. CHASER LIGHTS. Sept. 79. E18.95. VARICAP M.W. RADIO. Sept. 78. SIMPLE TRANSISTOR TESTER. SERt. 79, \&6-28. TUNINE FORK AUg. 79 \&9.95. Sultable microphone \& plug $\approx 1.59$ extra. WARBLING TIMER. Aug. 79. E8-25 SV POWER SUPPLY Aug. 79.59 .94 inc . DWVANEE WHISTLER Aug. $79 \Sigma 3$ 19 TOUCH ON PHOT LIGHT Aug. 79
 DARKROOM TIMER, July 79.82 .47 .
WATER LEVEL INDICATOR. July 79. K4.60. ELECTRONIC CANARY.JUNe 79. 54: 99. LOW COST METAL LOCATOR. JUnE $78,55 \cdot 44$.
Handle \& coll former parts extra 25.55 . OUAD SIMULATOR QUAD SIM LATOR JUNe 19, $\mathbf{2 6 - 2 5}$. INTRUDER ALARM. May 1979, E96-71. ponents. WAVE CONVERTER. May 79. z15 36 inc, cases. THERMOSTAT. 'PHOTO' SOLUTIONS. May 79. © is 02. Less socket, tube TRANSISTOR TESTER. April 79. TOUCH BLEEPER. ADrH 79. £3.34. LATEST KITS: S.A.E. OR

ONE TRANSISTOR RADIO. Mar. 79 : Wlth Amplifier o Headset. Less case. E5:93. DELAY INDICATOR. Mar. 79. EA.21, SATILE POWER SUPPLY. Mar. 79. E8.99. MODULATOR. Feb. 79. $\mathbf{~} 1.56$ lass case and pins.
 ADJUSTABEE PSUU. Feb. 9 . Case (horizontal layout) 25.21 extra. CONTINUITY TESTER, Jan. 79. 15-02. FUZZ BOX. Dec. $78 .\{5.53$.
VEHICLE
IMMOBILISER. Inc. PCB.
 Case i ROd. AUDIO EFECTS OSCILLATOR. Nov
78. 23 asinc. board.
FUSE CHECKER. Oct 78. 81.97 .
C.MOS RADO.
C.MOS RADIO. Oct, 78. 59.33 . less handle a coll former.
GUITAR TONE BOOSTER. Sept. 78. E4.g9 inc. p.c.b.
SOUND TO LIGHT. Sept. 78 . Es.98.
FITER SLAVE FLASH. AUg. 78. 53.20 less SK1. SLAVE FLASH, Aug. 78. E3. 20 less SK1.
LOGIC PROBE. JHy 78. 22.53
IN SITU TRANSISTOR TESTER. JUNe 78, 25 -75.
 And difuser. POCKET TIMER. ADril 78 . $£ 2.93$. POCKET TIMER. ADHI 78, E2.98.
WEIRD SOUND EFFECTS GENERA. TOR. Mar. 78, E4.61. CHASER LIGHT DISPLAY. Feb, 78. E23.59 inc P.i.b. case extra $E 5 \cdot 21$.
AUDIO VISUAL METRONOME. Jan. 78 E4.93. RAPID DIODE CHECK. Jan. 78, $£ 2-34$. AUTOMATIC PHASE BOX. Dec. 77 . E9.55 inc. p.c.b. V . 77 . 814.36 ULTRASONIC REMOTE CONTROL. Nov.Dec. 77. E16-09. ${ }^{\text {TREASURE }}$ LOCATOR. Oct. 77, 810.31
 ELECTRONIC DICE. March 77. E4.83.
SOLL MOISTURE INDICATOR. JUne 77. E4.07 Inc. probe.

CA. E6 BATMERY STATEINDICATOR. Sept. 78. غ1.79 less case Inc. PCB.
RF. SIGNAL GENERATOR. Sept RATOR. Sept. 78. ADD-ON CAP
AT. $25-98$, A.F. SIGNAL GENERATOR. Aug. 78 less dial. E12 89 , MEADPHONE ENHANCER. Jan. 79. PASS
PASSIVE MIXER. Oct. 78. E3.72. MIC AMP. Dec. $78 . \pm 2$ 82-80.
AUDIBLE FLASHER, Dec. 78, $51-21$.

## LOW cosT METAL LOGATOR

COMPLETE KIT with HANDLE, COIL
FORMER, SCREWS etc. EEETRONIC
 3 BAND S.W. RADID Simple T.R.F. Design, Covering most cast Bands, Five controlst- Bandset, Aandspread, Reaction, Wavechange and change Switch. Use with Headphones or a Crystal earpiece Kit contains all the
components required, including the P.C.

Board a Case. Instructions are included Board \& Case. Instructions are included
with this kit.
KIT: EI . 97 . Headphones extra $£ 3.28$.

## MARCH 80 KITS

CABLE \& PIPE LOCATOR, $£ 3-40$ less STEREO HEADPHONE AMPLIFIER. E14.9\$
DOOREELL REGISTER, E3-39 5 RANGE CURRENT LIMITER. Es-24. KITCHEN TJMER. E12.46. UNIBOARD TOUCH SWITCH. E8-65.

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## 1980 ELECTRONICS CATALOGUE

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No minimum order-all products are stock class delivery of first class EUROBREAD BOARD. $£ 6: 20$. LOW COST LONG NOSE PLIERS. E1.97.
LOW COST CUTTERS, \&1-98.
SWING STORAGE DRAWERS ES-98. MULTIMETER TYPE 3100,000 o.p.v. with transistor tester ranges 239 -95. WIRELESS INTERCOM 2 STATION SIREM
SIREN. 12 V £5-95.
P.C.B. ASSEMBLY JIG. ह11.98.
P.C.B. ETCHING KIT. E4.98.
A.M.F.M. AIRCRAFT BAND PORABLE RADIO E10-95
WIRE STRIPPERS \& CUTTERS E2-21. ULTRASONIC TRANSDUCERS. $\mathbf{2 5} 50$

## Pal

P.A. MICROPHONE colled lead \& STEREO MICROPHONE PAIR £ $10-95$. MULTIMETER TYPE 1. 1,000 0.p.v. with probes, $2^{\prime \prime} \times 3 \frac{1}{n}_{\prime \prime \prime} \times 1^{\prime \prime} . E 6.98$.
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F.M. INDOOR AERIAL, 57p.

TELESCOPIC AERIAL. $120 \mathrm{c} . \mathrm{m} .22-38$. TELEPHONE FICK-UP COIL. 72p. CRYSTAL MICROPHONE INSERT. 58,
SPEAMERS MINIATURE. 8 ohm 87 p . 64 ohm stp. 80 ohm 81.25 .
PILLOW SPEAKER, 8 ohm stp.
$6^{\prime \prime}$ ROUND SPEAKER. $80 \mathrm{hm}, 5 \mathrm{~W}, ~ 22-28$. CABINET SPEAKER. 8 Dhm, $5 W$. $5^{\prime \prime}$ apeaker. Cabinet $10^{\prime \prime} \times 7^{\prime \prime} \times 4^{\prime \prime}, ~ ธ 6.75$. REEENTRANT HORN SPEAKER. 8 ohm
EARPIECES. Crystal 48p. Magnetic 18 p . STETHOSCOPE ATTACHMENT. Fits our earpieces 69p.
BUZZER. 6 V 32p. 12 V 25p.
MONO HEADPHONES. 2 K Padded. MONO HEADPHONES. 2K Padded. Superir. S MEADPHONE STEREO HEADPHONES. 8 ohm.
Padded, $£ 4.35$. Padded, £4.35.
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stamps.

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 S.A.E. Automatically Included with catalogues and orders. Keeps you up to date with Magenta and includes extra circuit ideas.INTERCOM. 2 Station. Desk. E7. 48 MICROPHONE DYMAMIC. 600 ohm Cassette type, E1-38.
DENTISTS MIRROR. Adjustable. 62 . 44 . JEWELLERS EYEGLASS. Ef-08p TRIPLE MAGNIFIER. E1-63.
HAND MAGNIFIER. $3^{\text {se }}$ Lens, $\mathbf{\text { H3. 43. }}$ SPECTACLE MAGNIFIER. CHps on to spectaci
ILLUMINATED MAGNIFIERS. $11^{\prime \prime}$ lens POCKET TOOL SET
POCKET TOOL SET. 20 piece, EA-02. SCREWDRIVER SET. SIX plece. e2-98. 9. MAX PUNCH. in $^{\prime \prime}$ 23.98. $\frac{1}{2}$ " 23 -06. DRILL 12V. Hand
DRILL 22 V .
Stand E ©-8s.
CAPACITANCE SUBSTITUTION BOX. Nine values, $100 \mathrm{pF}-0.22 \mathrm{uF}$, E2. 21. QUICKTEST. Malns connector. $\mathbf{£ 7}$-38. PLUG IN POWER SUPPLY. $6,7.5-9 \mathrm{~V}$ d.c. 300 mA . $84-65$.

SPRINGS-SMALL. 100 Asstd, Ef-08: CROC CLIP TEST LEAD SET. 10 leads with 20 cilps. 61-15.
DIMMER SWITCH. $240 \mathrm{~V}, 800 \mathrm{~W}$. $\mathbf{8} 4.48$. TRADITIONAL STYLE BELL. 3-8V. Jomm chrome gong. ei.60. UNDEROOM
70 mm . ER 2.48. F.M. TUN.
C.c. 29 -4s. PANEL KEY, HIgh speed, E4. 28. PANEL METERS. $60 \times 45 \mathrm{~mm}$. Modern style. $500 \mathrm{~A}, 100 \mathrm{uA}, 1 \mathrm{~mA}, 1 \mathrm{~A}, 25 \mathrm{~V}$ d.c.
NIGHT LIGHT, Plug type. E1-08.
CONNECTING WIRE PACK. $5 \times 5 y d$. coils, 5sp.
YERO SPOT FACE CUTTER, E1- 21 . VERO PIN INSERTION TOOL. $0-T^{\prime \prime}$ PESIOTOR COL
RESISTOR COLOUR CODE CALCU.
LATOR. 2iP.

## 

An easy to follow book suitable for all ages, ideal for beginners. No Soldering. Uses an 'S Dec' breadboard. Gives clear instructions with lots of pictures. 16 projects-including 3 radios, siren, metronome, organ, intercom, timer, etc. Helps you learn about electronic components and how circuits work. Component pack includes an S Dec and the components for the projects. ALSO AVAILABLE Adventures With Electronics . $\mathbf{2 1} \mathbf{7 5}$. Component Pack $£ 16 \cdot 72$ less battery.

## TEACH IN 80

E.E. SERIES-ALL COMPONENTS IN STOCK NOW FOR FAST DELIVERY. All top quality components as specified by Everyday Electronics. Our kit comes complete with FREE COMPONENT IDENTIFICATION SHEET. Follow this educational series and learn about electronics-Start today LIST A \& B components £22.95 also available LIST C (parts $7-12$ ) $£ 2 \cdot 45$. All orders sent by FIRST CLASS POST. Our kit contains all these parts:LIST A \& B: METER, BREADBOARD, TRANSFORMER, LEDS, POTENTIOMETERS, SWITCHES, SPEAKER, PLUGS, SOCKETS, BATTERY CLIPS, WIRE, CABLE, FUSES, FUSE-HOLDERS, KNOBS, RESISTORS, PHOTOCELL, DIODES, CAPACITORS.
CASE WOODWORK KIT $\mathbf{2 5} \cdot 98$ extra. Complete kit for tutor deck woodwork, contains all the softwood, hardboard, ramin. panel pins, adhesive, screws, feet, strap-handle, and fixings. Cut to size and ready to assemble.
IDEAL SOLDERING EQUIPMENT FOR THE TEACH IN AND ELECTRONICS

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SPARE BITS. Small. Standard, Large. 65p each.
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QUARTZ LCD
11 Function Slim Chronograph


6 digit, 11 functions, Hours, mins, secs., day, date, day of week, 1/100th, $1 / 10$ th, secs.,
10X secs., mins.
Split and lap modes.
Back-light, auto calendar.
Only 8 mm thick.
Stainless steel bracelet and back.

Adjustable bracelet. | Also avaliable: |
| :--- |
| SOLAR CHRONOGRAPH |
| M9 |

SAME DAY DESPATCH.
M3 Price includes POST \& PACKING

## QUARTZ LCD

Ladies Day Watch
Hours, mins., secs., day,
date, back light, auto. calendar.


Fully adjustable bracelet.
Only $25 \times 20 \mathrm{~mm}$ and 6 mm thick.

Price only
Silver or Gold.
e7.95
M15 SAME DAY DESPATCH. P.\&P. included
HANIMEX Electronic LED Alarm Clock
Features and Specification:
Hour, minute display. Large LED display with p.m. and alarm on indicator. 24 Hours alarm with on/off control. Display flashing for power loss indication. Repeatable 9 -minute snooze. Display bright/dim modes control. Size: $5.15^{\prime \prime}$ $\times 3.93^{\prime \prime} \times 2.36^{\prime \prime}(131 \mathrm{~mm} \times 11 \mathrm{~mm} \times 60 \mathrm{~mm})$.
Weight: $1.43 \mathrm{lbs}(0.65 \mathrm{~kg})$.

M13


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ALARM
with Snooze Alarm


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6 functions plus Alarm. Conference signal,
5 minute snooze alarm, Conference signal sounds 4 secs. before main alarm to give advance warning and an option to cancel. Snooze sounds 5 mins. after main alarm and is always preceeded by the conference signal. SAME DAY DESPATCH Price on M4 Price includes POST \& PACKING

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Ladies Cocktail Watch
Beautifully designed with a very thin bracelet.


Hours., mins, secs., day, date,
 backlight and autocalendar. Bracelet fully adjustable to suit slim wrists. State Gold or Silver finish. Only $25 \times 20 \times 6 \mathrm{~mm}$.
M18 SAME DAY DESPATCH.
QUARTZ LCD 5 Function

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 M1 SAME DAY DESPATCH. $£ 6.95$

Price only £14.95
P.\&P, included

QUARTZ LCD
ALARM CHRONOGRAPH
with $12 / 24$ display


Alam
Hours, mins, secs, day of week. Month, date, day of week, alarm,
hour, mins., a.m./p.m. 24 or 12 hour display mode. Alarm test. Chronograph, lap time, stop watch $1 / 10$ secs. M16

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## TEACH IN 80

We are again supplying all parts required for this major series which started last October. The price for all the Tutor Deck parts is £19-50. Also supplied without breadboard for $£ 13 \cdot 50$. The price for the additional components required for Parts $1-6$ is $£ 2-00$ and Parts 7-12 is £3-00. All prices include VAT and Postage, Reprints of parts 30p per month


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pins 3 \& 5 length 1.5 m
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116 Car aerial extension sc
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117 AC mains connecing!

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1335 pin DiN plug to 2 Phono plugs
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## Projects... Theory...

## and Popular Features ...

The integrated circuit ("microchip") is nowadays a commonplace component widely used in all manner of projects, from the simplest to the quite complex designs.

Indeed we have become dependent on this device to a very marked degree, for without it many projects would just not be possible. This month, as it happens, four out of the six projects presented use i.c.s.

The circuit diagram of any project where one or several i.c.s are used gives a much simplified picture of the true or actual state of affairs circuitwise. Within the outline symbol representing the ic. exists another world of microcircuitry, made up of elements that are the microscopic counterpart of those discrete resistors, diodes and transistors we are also dependent upon.

But is it important what lies within the ic.? Maybe not to the average constructor who sticks to published designs. The i.c.s can remain simply "black boxes" performing their allotted functions. The inside "works" need not concern the user.
Most significantly of all, thanks to the widespread use of i.c.s we have become accustomed to low component counts for many projects. So when some moderately ambitious circuit is presented today in an entirely discrete form we are likely to be staggered by the number of individual components involved.

The Dual line Game (an analogue computing system) uses 67 resistors, 49 transistors and 139 diodes. Truly profligate, one might exclaim. But we needs must refocus to the discrete scene, and regain true perspective in terms of actual basic circuit elements.

Of course using a large number of discrete components makes for more physical assembly work. But to the hobbyist that should not be any punishment. As a matter of fact the method of assembly adopted for the Dual Line Game is interesting and instructive in itself; it illustrates a compact "book form" of construction where stacked circuit boards are interconnected by ribbon wiring-a method often used in advanced professional equipment.
The hobby of electronics construetion undoubtedly owes much of its advancement and popular appeal to microelectronics. Yet there is still plenty of fun and instruction to be derived from the use of discrete semiconductors, which bring one into more intimate contact with the basic circuitry. And suitable devices are generally obtainable at attractive prices.


Our June issue will be published on Friday, May I6. See page 337 for details.

## Readers' Enquiries

We cannot undertake to answer readers' letters requesting modifications, designs or information on commercial equipment or subjects not published by us. All letters requiring a personal reply should be accompanied by a stamped self-addressed envelope.

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## Component Supplies

Readers should note that we do not supply electronic components for building the projects featured in EVERYDAY ELECTRONICS, but these requirements can be met by our advertisers.
All reasonable precautions are taken to ensure that the advice and data given to readers are reliable. We cannot however guarantee it, and we cannot accept legal responsibility for it. Prices quoted are those current as we go to press.

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free with this issue E.E. CONSTRUCTORS RULE (Mounted on front cover)

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Certain back issues* of EVERYDAY ELECTRONICS are available worldwide price 70 p inclusive of postage and packing per copy. Enquiries with remittance should be sent to Post Sales Department, IPC Magazines Ltd., Lavington House, 25 Lavington Street, London SE1 OPF. In the event of non-availability remittances will be returned.

* Nor available: October 1978 to May 1979.


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performance. Insufficient voltage reduces sensitivity, while excessive voltage tends to cause instability.

The positive supply is at pin 1 , via R3, and obtained from a potential divider formed by R2 to the battery negative line, and VR1 with R5 to the positive line. VR1 allows adjustment to cover component tolerances, to avoid any difficulty in this direction.

Audio output is via R4 and C9 to the volume control VR2.

## AUDIO SECTION

The amplification section consists of TR1 and IC2, the latter requiring very few external components.

Transistor TRI is connected as a common emitter amplifier with feedback biasing via R6. This acts as a preamplifier for IC2 giving additional boost to the output of IC1.

# 4.STATION 

THIS receiver allows switch-selection of Radio 1, 2, 3 or 4 , without any need for tuning adjustments when in use. For a good performance without too much circuit complication, two i.c.s and a transistor are employed.

With the aerial and trimmer values shown, approximate possible coverage at each switch position is: 1 and 3 , $1800850 \mathrm{kHz} ; 2500-950 \mathrm{kHz}$; and $4150-$ 230 kHz (longwave). This includes the alternative programme frequencies where applicable.
Note that in the London area, Radio 4 is best received using the 720 kHz frequency (medium wave), switch position 2. Other frequencies include the following: Radio 1, 1053 kHz , 1089 kHz ; Radio 2, 693 kHz , 909 kHz ; Radio $3,1215 \mathrm{kHz}$; Radio $4,720 \mathrm{kHz}$, 200 kHz ; Radio Wales 882 kHz ; LBC, 1152 kHz ; Radio London 1458 kHz ; Luxembourg 1439 kHz ; Capital 1548 kHz .

## DETECTOR SECTION

The circuit diagram is shown in Fig. 1. The 2 -pole 4 way switch Sla allows the choice of four pre-set frequencies. With SI at 1 trimmer Cl is in circuit, while position 2 introduces C 2 instead, and 3 connects C3. Positions 1,2 and 3 of S1b short the longwave aerial winding L2, for medium wave reception by L1. Position 4 gives longwave reception, with trimmer C4.

To avoid the additional tuning and extra components of the superhet, the ZN414 10 -transistor t.r.f. integrated circuit ICl is used. The r.f. resulting from the resonant tuned circuit composed of C1, 2, 3 or 4 with L1 and L. 2 is fed to the input (pin 2) of IC1 which provides audio frequency output at pin 1, so that it is only necessary to follow this with an audio amplification stage.

The ZN414 i.c. requires a supply of about 1.3 volts and small changes in voltage have a significant effect on its

IC2 has a gain of 50 , set internally and an output power rating of over 1 watt into a 4 ohm loudspeaker. In this design an 8 ohm loudspeaker is used to reduce the peak current drawn and conserve battery. This still provides adequate power output for normal listening.
In areas of strong signal TRI could be omitted. Anyone wishing to reduce the number of components and experiment with this can omit R4, R6, R7, R8, TR1, C10, C11 and C12, taking pin 2 of IC2 to the wiper tag of VR2.

## HOW IT WORKS



Radio frequency signals from the transmitter induce voltages in the coils on the ferrite rod aerial. Four r.f. tuned circuits are formed with the aerial which can each be inputted to the r.f. amplifier/detector i.c. The output from this is the required audio signal which receives boost from the audio amplifier to be heard in the loudspeaker.


Fig. 1. The full circuit diagram for the Pretuned 4-Station Radio. The coils with ferrite rod core is a standard ready made component type FR5.


## CHASSIS

The chassis, which finally forms the top of the case, consists of a $7 \times 4$ in "universal chassis" flanged member. Drill or punch holes for VR2 and S1. Also drill the fixing holes to match those near the corners of the board, to enable the board to be fixed to the chassis. These bolts are countersunk headed, and at least 15 mm long. Spacers or extra nuts hold the board clear of the metal panel, and the tag

MC (Fig. 2) must provide a sound connection from negative line to panel.

Connections between board, S1 and S2 can now be made according to Fig. 2.

## SELECTOR SWITCH

A 2-pole 4 way rotary switch may be supplied as 6 -way, with a stop which can be moved before the switch is

## CIRCUIT BOARD

The circuit board employed is $0 \cdot 1$ inch matrix perforated s.r.b.p. measuring $15 \times 37$ holes.

Both sides of this are shown in Fig. 2. First drill the board for the two metal brackets, and the bolt to secure tab MC. Both brackets are made from aluminium to the dimensions shown and can be fixed in place.

When fitting the board mounted components as shown, their wire ends will be long enough for many connections. Elsewhere, use 20 s.w.g. or similar tinned copper wire. With ICl, note that pins 1 and 2 are well separated, with the earth lead 3 between. The preset VRI is placed so that it can be adjusted from the rear of the completed radio with back panel removed.

IC2 is fitted in a holder, and must of course be the right way round. An indentation exists between pins 1 and 14 to aid polarity identification. Note that there are eight pins of ICI that connect to the negative line.

Connect suitable lengths of flying leads to reach the chassis mounted components and the loudspeaker. Multistrand wire is recommended.


The completed prototype with back and side panels removed.

## PRETUNED 4-STATION RADIO



A close-up view of the trimmer capacitor bank showing fixing brackets used in the prototype.

## 

## Resistors

| Resistors | R1 $100 \mathrm{k} \Omega$ |
| :--- | :--- |
| R2 $680 \Omega$ | R5 |
| R3 $37 \mathrm{k} \Omega$ |  |
| R4 $470 \Omega$ | R6 $820 \mathrm{k} \Omega$ |
| R4 $4.7 \mathrm{k} \Omega$ | R7 $10 \mathrm{k} \Omega$ |
| All +W carbon $+5 \%$ | R8 $33 \mathrm{k} \Omega$ |

Potentiometers
VR1 $1 \mathrm{k} \Omega$ vertical miniature pre-set
VR2/S2 $10 \mathrm{k} \Omega$ log. pot with ganged switch
Capacitors


## C8 10 nF ceramic <br> C9 $\quad 0.1 \mu \mathrm{~F}$ ceramic or polyester <br> C10 $0.47 \mu \mathrm{~F}$ ceramic or polyester <br> C11 10 nF ceramic <br> C12 $4 \cdot 7 \mu \mathrm{~F} 10 \mathrm{~V}$ elect. <br> C13 $470 \mu \mathrm{~F} 10 \mathrm{~V}$ elect. <br> C14 $1000 \mu \mathrm{~F} 10 \mathrm{~V}$ elect.

Semiconductors
TR1 2N706 npn sllicon
IC1 ZN414 t.r.f. radio i.c.
iC2 LM380 audio amplifier i.c.
Miscellaneous
S1 2-pole 4-way rotary switch
LS1 8 ohm moving coil loudspeaker


L1/L2 ferrite rod aerial type MW/LW FR5 (Denco)
0.1 inch perforated s.r.b.p. size $37 \times 15$ holes; s.r.b.p. board size $30 \times 40 \mathrm{~mm}$ (for aerial mount); 14 pin d.i.1. socket; control knobs (2 off): PP9 battery clips; materials for case; aluminium for brackets; 6 BA nuts, bolts and washers; "universal chassis" flanged member size $7 \times 4$ inch ( 2 off).
mounted. Fixing the switch holds the stop in place.
Those with an experimental inclination can use the two spare ways of a 6 -way switch for l.w. and m.w. manual taning, with a 200 pF variable capacitor.

## FERRITE AERIAL

Bolt a strip of perforated board $30 \times 40 \mathrm{~mm}$ (approx) to the central bracket. Secure the rod to this, as shown in Fig. 2, with adhesive and cotton passed through holes in the board. L1 (medium wave) and L2 (long wave) windings are then slid on. The small coupling winding provided on L1 is not needed, so this should be removed.

The beginning of L1 is denoted by A. Put sleeving on this lead and solder
it to Sla. The end of LI and beginning of L2 is denoted by $B$, and these are joined at the strip supporting the rod, and a lead run from this point to Slb.

The end of L2, C, runs to one side of all trimmers and then Slb. The windings of L1 and L2 must be in the same direction. This can be seen. Alternatively, if tuning of 200 kHz is impossible, take $L .2$ off the rod, reverse it and replace it.

## TRIMMER BANK

The four trimmers are supported by a strip spanning two brackets, see photo. Stick a strip of paper marked 1, 3, 2, 4 close to the trimmers on the chassis to identify adjustment from behind. Wire the trimmers to S1, connect LSI and a battery.

Initially position L1 and L2 each about 20 mm from the ends of the


The fully wired chassis assembly with aerial secured to board.
rod, and check that with S1 in position 1, Radio 1 can be received with adjustment of C1. Similarly, that positions 2, 3 and 4 allow Radio 2, Radio 3 and Radio 4 to be tuned by C2, C3 and C4 respectively. If not, move LI slightly, or L2 for Radio 4. When in order, cement L1 and L2 to the rod.

Other radio stations may be pretuned in place of any of the above mentioned; see frequencies and recommended position discussed earlier.

## CASE

The case can be made with negligible cost and difficulty, but with good appearance. In the prototype, another $7 \times 4$ in "universal chassis" flanged member formed the bottom panel. Each side is wood, cut 115 mm high by 130 mm . Four 6BA countersunk bolts fit each side to the chassis members. The bottom is flush, but the top panel is set down 5 mm or so. The front is another piece of 4 mm plywood size $178 \times 115 \mathrm{~mm}$, with an aperture cut for the speaker, which is held with countersunk bolts, or self-tapping screws.

After checking that the pieces will fit, temporarily remove the sides. A piece of fabric about 215 mm wide and 305 mm long is then fixed with a suitable quick-drying adhesive along the front of the bottom member. When this is dry, stretch the fabric upwards over the front, across the top panel, and cement it at the back edge of this panel. It is necessary to cut holes for the controls.

When the adhesive is dry, apply more adhesive to the end flanges at one side, and along the edge of the front. Pull the fabric taut, fold it over, and screw on the side. Repeat for the other side. The back panel is the same size as the front panel and held with self-tapping screws.

## FINAL ADJUSTMENTS

Glue on a facia and label as seen in the photographs. Carefully check the four trimmers. Also set VR1 so that about 1.2 V to 1.3 V is present across C6 (using a high resistance voltmeter) or so that sensitivity is good without instability.

For best reception in some areas it will be found necessary to rotate the receiver, for best volume of the wanted transmission, and minimum pick-up of other signals.

In many areas VR2 should not need to be advanced very far and the receiver will in general be operated at considerably lower volume than the maximum which could be obtained.
Final adjustment to the trimmers, for the transmissions which are required in any particular region, are preferably made with a completely insulated tool, with all the assembly completed.


$A^{4}$
L drivers must at some time have left their lights on when parking. This is particularly likely to happen when driving in daytime with lights on, during rain or foggy conditions, and also when setting off in darkness and arriving at one's destination in daylight.

The circuit to be described will alert one to this condition. It has the advantage that no setting or cancelling switches are required; in some cases, no additional warning light may be needed. Instead, an existing panel lamp may be used.

The indicator lamp (which may be replaced by a buzzer if required)
flashes when the engine is switched off with the lights left on. Turning the lights off cancels the alarm. If required, the lights can then be turned on again without any alarm indication. This allows for parking with lights in awkward or hazardous circumstances. If the alarm is activated and the engine re-started, the alarm is cancelled.

## CIRCUIT DESCRIPTION

The circuit diagram is shown in Fig. 1. The circuit has three basic inputs, +12 V , ignition and lights. The 12 volt supply (car battery) is
connected at all times. The current drain of the circuit is negligible and the car could be left for weeks without the battery becoming flattened. A car clock consumes far more.
The +12 V supply is filtered by R1 and C3 which removes any noise on the line. The filtered supply is taken to R5, R7 and pin 14 of IC1, a quad dual-input NOR gate.

The ignition line is also filtered, this time by R2 and C1. When the ignition is turned on, TRI is turned on via the flow of current through its base network R3 and R4. The collector of TRI is therefore pulled low, taking the input to pin 6 with it.

Fig. 1. The complete circuit diagram of the Lights Warning System. LP1 is a 12 V bulb up to 2 watts and is only required if no in-built panel light is available.


## COMPONENTS



The prototype component board. This differs from that shown below where D2, D3 and R14 have been incorporated on the board.

Resistors

| R1 | $470 \Omega \frac{1}{2} \mathrm{~W}$ | R8 | $100 \mathrm{k} \Omega$ |
| :---: | :---: | :---: | :---: |
| R2 | $1 \mathrm{k} \Omega$ | R9 | 100k |
| R3 | $10 \mathrm{k} \Omega$ | R10 | $1.5 \mathrm{M} \Omega$ |
| R4 | $2 \cdot 2 \mathrm{k} \Omega$ | R11 | 10kS |
| R5 | $100 \mathrm{k} \Omega$ | R12 | $2 \cdot 2 \mathrm{k} \Omega$ |
| R6 | $10 \mathrm{M} \Omega$ | R13 | $10 \mathrm{k} \Omega$ |
| R7 | $1 \mathrm{M} \Omega$ | R14 | $47 \Omega \frac{1}{2}$ |
| All $\div$ W carbon $\pm 5 \%$ unless stated otherwise |  |  |  |
|  |  |  |  |
| Capacitors |  |  |  |
| C1 $0.1 \mu$ |  |  |  |
| C2 $0.47 \mu \mathrm{~F} 16 \mathrm{~V}$ t |  |  |  |
| C3 | $100 \mu \mathrm{~F} 16$ | elect. |  |
| C4 | $0.33 \mu \mathrm{~F}$ p | yester |  |

## Semiconductors

> D1 OA200 or similar silicon

D2, 3 1N4001 or similar 1 A rectifier (2 off)
TR1, 2 BC108 npn silicon (2 off)
TR3 BC307B pnp silicon
IC1 CD4011 CMOS Quad $2-$ input NAND gates

## Miscellaneous

0.1 inch matrix stripboard size 25 strips $\times 18$ holes; Veropins ( 7 off); 14 pin d.i.1. i.c. socket; 7 -way 2-amp screw terminal block; case.

REAR
SCREFN

(O) = VEROPIN



Fig. 2. The layout of the components on the topside of the board and the breaks to be made in the copper strips. Numbered leads from the terminal block refer to Veropins on board. Only one indicator lamp should be used, either LP1 or existing panel lamp for some other function (e.g. rear screen demister indicator).

## HOW IT WORKS



The logic control circuitry normally provides an output to inhibit the operation of the oscillator (it does not oscillate). However, when a condition arises where the lights are on and the ignition is turned off, the logic output changes and allows the oscillator to operate. This activates a warning lamp (or buzzer) inside the car to flash (or sound). The logic circuitry has a built-in parking facility which is enabled by turning the lights off and then on after the alarm has sounded.

A bistable latch is formed with ICla and IClb which can be set and reset by placing a low on pins 1 and 6 respectively. Once the latch is set or reset it remains in that condition indefinitely until the corresponding input is taken low to change it.

Pin 1 is normally held slightly below positive rail voltage (logic 1) by the network R6 to R8. We shall now assume that the car is being driven with the lights on. The ignition and lights lines are both high (logic 1) and TR1 collector is low. The output from pin 3 is low.

## OPERATION STAGES

The driver arrives at his destination and turns off the ignition. The ignition line goes low, and TR1 turns off; pin 6 goes high. The low excursion of the ignition line takes pin 1 low, via C3. Capacitor C3 then charges via R7 so that pin 1 is high again. The low on pin 1 has, however, set the latch, so that the output from pin 3 is high. This output is taken to pin 8 where it enables the relaxation oscillator consisting of IClc and ICld to operate.

The output from pin 11 pulses alternately high and low, at a rate determined by the time constant ( $\mathrm{C} 4 \times \mathrm{R10}$ ). In this case a rate of approximately 1 Hz is obtained.

The output from pin 11 is taken via R11 to TR2, and turns TR2 on and off. Likewise, the output from TR2 turns TR3 on and off, via the network R12 and R13. Transistor TR3 is a $p n p$ transistor fed from the lights rail. Its output is taken to a suitable indicator lamp LP1, for example 12 V 100 mA .

In the author's prototype, the lamp used was the fitted rear screen demister indicator. This is fed via D2, R14 and D3, so that the light is on dimly when the demister is in use, but when the alarm turns on, the lamp flashes brightly.

## PARKING FACILITY

It will be noticed that DI is also connected to the hights line. When
the driver turns the lights off, the cathode of D1 goes low, and the input to pin 6 is pulled low. This resets the latch, and cancels the alarm. The lights can then be turned on again, but since the latch has been reset, the alarm condition is no longer given. The output at pin 3 will now be low, and so will pin 11. Thus TR2 and TR3 are turned off.

If the driver, instead of switching his lights off, starts his engine after an alarm has been given, TR1 turns on again and pulls pin 6 low, so resetting the latch and cancelling the alarm.

It will be observed that pin 6 is permanently low while driving. This ensures that no transient pulses from the car can set the latch and give a spurious alarm. This could be a most disturbing situation in the middle of a motorway, for example.
This system has been in use from the winter of 1977 onwards without malfunction and has proved very useful.

Those readers wishing to use a separate alarm lamp or buzzer may dispense with D2, D3 and R14. The lamp is then connected between chassis and the collector of TR3.


## CIRCUIT BOARD

All the components (except the indicator lamp or buzzer) are mounted on a small piece of 0.1 inch matrix stripboard size 25 strips by 18 holes. The component layout on the topside of the board and the breaks to be made on the underside are shown in Fig. 2. This differs slightly from the author's prototype as shown in the
photograph where D2, D3 and R14 were mounted around the alarm lamp.

The prototype was wrapped in foam sponge and this held the board in position in the case. If a bolt fixing is preferred, the board should be extended to accommodate the fixing holes. The screw-terminal block can be fitted to the outside of the case which should be equipped with lugs.

## ASSEMBLY

Begin by making the breaks on the underside and fitting the Veropins and i.c. socket. Continue with the link wires, resistors, capacitors and semiconductors in this order. Solder sufficient lengths of flying leads and secure these to the terminal block. Finally insert ICl and the unit is ready for assembly in its case and installation in the vehicle.

Note that if a buzzer is used in place of a lamp, a diode, must be connected across its coil to prevent the back e.m.f. generated when activated, from damaging TR3. The cathode of the diode should connect to TR3 collector with its anode connected to chassis, i.e. the diode is normally reversed biased.

## INSTALLATION

It is recommended that the unit be installed inside the car in an area not prone to excessive temperature changes (e.g. heater flow duct) or moisture build-up. The glove compartment is a likely position.

All the connections from the terminal block should be made with flexible (multistrand) wiring to reduce possibility of vibration fractures. Connection to the required points in the electrical system are best made with push-fit spade connectors or the "piggy-back" type. Self-adhesive foam pads are suitable for fixing the case to a horizontal surface and do provide some form of shock absorption.

"I was going to construct a Time Machine, but there's no future in it."


## By Dave Barrington

For The Workbench
We would think that with the "shrinkage" in size of most electronic components over the years, a light duty vice is now considered a must for most workshops. Two such products worthy ofcon sideration for addition to the workshop are the OK VV-1 vice and the Absonglen Minibench.
Both professional and amateur will find OK's new VV-1 light duty vice ideal for handling small components and assemblies.
The 33 mm jaws are adjusted by a knob which takes the place of the usual "tommy" bar. Normally the vice would be fixed to worksurfaces by a lever operated suction mechanism but where permanent installation is required it can be screwed down.
The $V V-1$ vice is available from $O K$ Machine \& Tool (UK) Ltd., Dept EE, Dutton Lane, Eastleigh, Hants., SO5 4AA, price $£ 3 \cdot 58$ inclusive of packaging.
Although we have not had the pleasure of trying the Minibench we like the idea of gripping small boards in its jaws and being able to "flip" the workplece over for component mounting and soldering. Another good feature is the facility for flexible arms terminating in crocodile clips to hold components and possibly to act as heat sinks.

Further information can be obtained from Absonglen Ltd., Dept EE, The Forge, Staplow Cottage, Staplow, Ledbury, Herefordshire, HR8 1NP.


VV-1 vice from OK Machine \& Tool

## Magnetised

An item which you would not normally associate with the amateur workshop, but once used soon becomes an invaluable "tool" is the magnetiser/demagnetiser.
Ideal for demagnetising those screwdrivers and drills which have a habit of becoming magnetised, without reason and picking up any metal debris, including nuts and bolts, often found on the workbench and depositing them in the workpiece before you can take any action. Then there's the opposite case where you wish you had a magnet to reclaim an item which had accidently dropped into the "works" and caused untold time wasting and curses trying to devise other methods of retrieval.
A suitable magnetiser/demagnetiser, at a fairly low budget price of $£ 9 \cdot 95$, is available from Magnadyne Products and may prove to be a worthwhile investment for those "butterfinger" types like myself.
Claimed to be capable of producing a bar magnet up to 9 mm diameter and any length, the magnetiser/demagnetiser is available from Magnadyne Products, Dept EE, 6 Lulworth Road, Welling, Kent, DA16 3LQ.


## Extractor Plier

Another unusual item, available from Toolrange, which may appeal to the constructor who has tried to remove i.c.s from circuit boards is the DIP Extractor Plier, price $£ 6 \cdot 50$ plus VAT.
It is claimed that this new hand tool will remove 8 to 40 pin dual-in-line devices from sockets and desoldered circuit boards in seconds. The teeth of the pliers grip devices as close as 0.25 in in densely packed boards. The pliers are insulated for live board work and the metal teeth also act as a heat sink and shorting bar.

## PLEASE NOTE

All prices quoted in this issue are Pre-Budget

## Soldering Irons

A new range of soldering irons from S \& R Brewster Ltd., may be worth further investigation.

These irons are the Ceco Vari-Stat range, for which they have acquired the manufacturing and sales rights, used extensively by the trade.

Amongst the range is the S.R.B. "Mighty Midget" which is an excellent 18 watt iron with interchangeable bits suited for our type of work. The more robust of their range, primarily designed for industrial use, are temperature controlled and ratec from 50 W to 500 W . For more details reajers should contact $S \& R$ Brewster Ltd, Dept EE, 86-88 Union Street, Plymouth, PL1 3HG.
Another company we must not overlook when we are talking about soldering iron requirements is Light Soldering Developments Ltd.
They cover such an extensive range that you can practically hand pick one to meet your own personal needs, including a "carry-it-anywhere" cordless type with built-in recharger.
For more details and leaflets readers should write to Light Soldering Developments Ltd., Dept EE, 97/99 Gloucester Road, Croydon, CRO 2DN.

## CONSTRUCTIONALPROJECTS

Avallability of components for this month's constructional articles should not cause too much dificulty and should only be a matter of looking through advertisers catalogues.

## Dual Line Game

In view of the large quantities of transistors and diodes called for in the Dual Line Game it would seem to be a good idea to approach advertisers regarding a special "bulk purchase" price for these items. A discount price may also be available for the 3.5 mm jack plugs and sockets. The rest of the components for this project should be ready available items.
Most advertisers will probably supply the BC108 in the metal TO-18 can. This also applies to the Lights Warning.

## Battery Voltage Monitor

The multiturn preset potentiometer VR1 listed in the Battery Voltage Monitor project is available from Watford Electronics and Maplin Electronic Supplies. However, there is no reason why a standard miniature preset potentiometer should not be used with a slight loss in fine adjustment.

## Lights Warning System

The CD4011 used in the lights Warning System is a standard CMOS i.c. and is usually generally available from most advertisers. Transistor type BC307B is listed by Watford Electronics. The suffix $B$ denotes the gain of the device.

## Pretuned 4-Station Radio

The ferrite coil needed in the Pretuned 4-Station Radio is the Denco type MWI LWFR5 and is stocked by Watford, Maplin and Ambit International. Suitable trimmer capacitors for C1 to C4 are available from Home Radio as types VC29LC and VC29L.
We cannot foresee any buying problems for the Uniboard-Audio Tone Generator and the Auto Fade projects.


Following last month's description of the basic principles of transistors, we will now go on to look at one of the most important uses of transistors-as simple two-state devices, with a clearly defined ON state and OFF state.

In isolation such circuits can be used to drive simple devices such as indicator lamps or relays but together with other similar circuits they can form part of a logic circuit such as may be found in a digital computer.

## ANALOGUE AND DIGITAL

Over the past twenty years or so electronics has divided into two main fields: analogue and digital. They have almost become two separate disciplines since the techniques used in each appear to be completely different.

Analogue electronics is concerned with the generation and manipulation of continuously changing signals. The term "analogue" is used since the electrical signals are often an analogue of some physical quantity. For instance, an amplifier has voltages inside it which are analogies of the sounds which we want to hearvoltage is used to represent air pressure.

In digital circuits only two states are normally to be found for any particular section. These two
states are used to represent the two logical alternatives true and false. By coding information in terms of complicated patterns of true and false, complex and manipulations can be carried out on the data at incredible speeds.

Analogue electronics is the older field, probably since continuously variable signals occur more often in nature than discrete signals. A wave in water or sound travelling through the air can be simulated using electronic voltage to represent height of the wave or sound pressure. Fig. 8.1 shows an analogue and a digital signal compared.

Fig. 8.1. The analogue signal shown at (a) varies continuously between its limits whilst the digital signal at (b) only takes two distinct values, the change from one to the other being very rapid.

Fig. 8.2. A switch which is part of a system must be able to switch other similar circuits and must be capable of being switched by a circuit similar to itself.

Digital circuitry implies a more abstract conception of real situations but the power of this approach can be judged by the proliferation of digital machines in more and more aspects of modern life.

The purpose of this month's part is not to look at "logic" in the abstract but to look at how transistors can be used as twostate devices and how they can be combined to do useful tasks.

## THE TWO STATES

The basis of the digital approach is that the transistor is used simply as an on/off switch, the two states representing "true" and "false."

It does not matter which state we use to represent true provided all the circuits are designed with the same convention. Also it does not matter what the absolute voltages are which are used to indicate true and false; provided there is no possible way in which they can be confused we are quite safe.

The two states which we are trying to achieve must therefore be stable, repeatable and independent of variations in the transistors themselves.

Of course, there must be a time when the transistor is in neither the ON state or the OFF state, that is when it is changing from one state to the other, but we must make sure that the time the transistor spends in this intermediate state is as small as possible and that it cannot get stuck in this undesirable position.



Fig. 8.3. A simple transistor switch using the common emitter configuration. The base current must be greater than that just needed to saturate the transistor.

Let us look at how the circuitry around a transistor can be designed so as to achieve the conditions we require.

If the switch that we are designing is to be used along with other similar switches in a system, then it must both be capable of being switched by another stage and of switching other similar stages.

Fig. 8.2 shows how the switch might fit into a system. Note that we do not allow outputs of different switches to be connected together as we could get into a situation where one switch is trying to switch on and the other offwhich one wins will be a matter dependent on the circuitry rather than the inputs to the two switches.

However more than one input can be connected together since a switch will often be required to drive many other switches.

## THE OFF STATE

If we define the transistor as being off when no current flows in its collector then we can achieve this quite simply with a common emitter circuit by either leaving the base open circuit or by making sure that the voltage on the base is not high enough to forward bias the base-emitter junction.

Now, since we have said that the switch must be able to be driven by another similar switch, the option of open-circuiting the base is not possible. We must therefore make sure that the voltage on the output of our switch is well below the voltage needed to forward bias base-emitter, that is about 0.7 V for a silicon transistor.

As we saw last month, the current through the collector will never be exactly zero as some
leakage current will flow even when the base is not forward biased. However, this current is so small that it can be ignored in this context.

When the transistor is in this state we say that it is "cut-off".

## THE ON STATE

Looking at the circuit of Fig. 8.3 , if the base current is gradually increased from zero the current in the collector resistor $R_{0}$ will gradually increase and the voltage at the collector will fall steadily because of the increasing voltage across the resistor.

Now the voltage at the collector cannot fall to zero but reaches a minimum of about 0.2 V for a silicon transistor. In order that this voltage is reached no matter what the gain of the transistor, we can use a safety factor of ten by making the base current ten times that which is needed to just saturate the transistor, assuming minimum gain.
For a BCl08 the minimum gain is 110 so providing the base resistor is ten times the value of the collector resistor we can be certain that even at worse case the output voltage will be around 0.2 V when the input is taken to the positive supply.


Fig. 8.4. The transistor switch (Fig. 8.3) as it looks when in a system. Note that only one output feeds it whilst it feeds many inputs.

Increasing the base current still further produces a lower voltage at the collector until a point is reached when increasing the base current any further produces virtually no change in the collector voltage.

What has happened is that the base current has produced a current in the collector which produces a voltage across the collector resistor which is almost equal to the supply voltage. Obviously this resistor cannot drop more than the supply voltage so the extra base current has very little effect.

We call this state saturation for obvious reasons. To define saturation more rigorously we say that saturation will be reached when the base current exceeds the maximum collector current (the power supply voltage divided by the collector resistor) divided by the $h_{\mathrm{FE}}$ of the transistor.

$$
\text { Saturation: } I_{\mathrm{b}}>\frac{V_{\mathrm{s}}}{R_{\mathrm{c}}}+h_{\mathrm{FE}}
$$

## RESISTOR VALUES

To put in some resistor values then let us use a 1 kilohm collec tor resistor and a 10 kilohm base resistor. Does the circuit thus produced fulfil our needs? To find this out we have to slot the circuit we have produced back into our original system of Fig. 8.2 (see Fig. 8.4).


Fig. 8.5. The loading on the switch in Fig. 8.4 can be redrawn thus. The load appears as a number of $10 \mathrm{k} \Omega$ resistors in parallel all to 0.7 V .


Fig. 8.6. When a pulse is applied to the input of our switch the output does not change instantly but with the delays as shown here.

Assume the first stage $(A)$ is on so that the output voltage is 0.2 V . The base emitter junction of the second stage ( $B$ ) cannot be forward biased as the voltage to do this is 0.7 V . Thus stage $B$ will be off (no current in its collector resistor). Now the collector resistor of stage $B$ has to supply the base current to all the other stages to which it is connected.

Each stage has a 10 kilohm input resistor, all of which have 0.7 V on their lower ends if their respective transistors are conducting. We can thus draw the loading on stage $B$ output as in Fig. 8.5.
Note that if ten stages are connected to output $B$ then the voltage drop across its collector resistor will be 4.15 volts and the current through each of the base resistors will be only half of that stipulated in the first design.

If we now say that this current is our absolute minimum then we are merely stating that one of these circuits can drive no more than ten similar circuits.
This figure is called the fan-out of the circuit since it defines how many other circuits we can fan out our output to.

Apart from this restriction our circuit seems to meet the criteria of two distinct states, independent of the transistor variations. But how long does it take to switch from one state to the other?

## SWITCHING SPEED

To predict how long a switching circuit takes to go from one state to the other requires a much deeper investigation of the workings of a transistor that it is possible to present in this series.

Fig. 8.7. If it is desired to drive a light emitting diode from one of our switches then the circuit shown can be used.

However, we can look at the actual behaviour of the circuit and make some remarks about the speed of switching circuits in general.

Assuming we are driving the switch with a voltage that changes from high to low and back again in zero time, what will the output of the circuit under consideration look like? Fig. 8.6 shows the input and output waveforms that we are likely to see.

We see that the output does not follow the input immediately but both at "switch on" and "switch off" it goes through two stages.
There is a delay when nothing appears to be happening followed by a gradual change from one state to the other. The delay before the transistor switches on is due to the time taken for the base voltage to build up to a value sufficient to forward bias the baseemitter junction. After this the collector current increases steadily to its final value. We call these two times the turn-on delay and the rise time.
Now, we are putting much more base current into the transistor than is needed to produce the collector current that we have. When the input voltage is taken to zero this extra current is "remembered" in the form of a charge within the transistor which tends to keep the transistor on for a short time. This is called the turn-off delay. After this the collector current falls steadily to the off state. This is called the "fall time".
You may have got the impression that we have a very slow circuit but it must be realised that the times we are talking about are measured in millionths of a second
(microseconds or $\mu \mathrm{s}$ ) or less so that even this very crude switch is extremely fast by everyday standards.

## HIGH CURRENT LOADS

Quite often the requirement will arise to be able to drive some sort of output device such as a lamp or a relay using a simple switch such as the one we have described. We used light emitting diodes in an earlier section of this series to indicate the flow or absence of current, so how would one of these be driven by a transistor?
First ascertain the current needed through the load; for an l.e.d. this would be about 10 mA . Now the l.e.d. will drop about 1.6 V when it is conducting. The transistor if fully on will drop about $0 \cdot 2 \mathrm{~V}$ so that we must use a series resistor to drop the remaining voltage which will be the power supply minus $(1 \cdot 6+0 \cdot 2) \mathrm{V}$. For a 9V supply the resistor must drop 7.2 V at 10 mA . This works out as 720 ohms, the nearest preferred value being 680 ohms.
Assuming the transistor has a minimum gain of 100 then a base current of 0.1 mA will just saturate the transistor. Allow a safety factor of ten, giving a required base current of 1 mA . If we are driving this switch from a similar one then we must be rather careful in selecting our base resistor since our base current has to flow through the l.e.d. of the previous stage.

A base resistor of 8.3 kilohm is needed if connected to the power supply, so dropping this to $6.8-$ kilohm should give some allowance for the voltage drop across the l.e.d. and its resistor.

But note: if we try to feed two similar stages then the current through the output l.e.d. will be 1.5 mA , enough to illuminate it! Therefore this circuit has a much lower fan-out than the transistor switch described previously. In fact it has a fan-out of only 1. Fig. 8.7 shows the circuit.

## A TWO-STAGE CIRCUIT

As an alternative to a l.e.d. we might want to drive a relay which needs say 100 mA . Out first problem is that 100 mA is the maximum current for a BC108 transistor so right away we must look for a transistor with a greater current capability than this.

The next problem is that using our "ten times" criterion means that we need 10 mA to switch on the driving transistor and if we want to drive it from the output of one of our simple transistor switches, this value of current is just not available.

The answer to the problem is that we will have to use a twostage circuit of some kind: the first one to "buffer" the output of the driving switch from the high current demand of the relay driver.

A simple circuit is shown in Fig. 8.8. The first transistor (TR1) needs only 1 mA to switch it on. When it is on there is only 0.2 V on the base of the driver (TR2) so the latter is off. When the first transistor is off then 10 mA base current flows into the driver through the collector resistor of the first transistor which now be-


Fig. 8.8. A relay with its higher current requirement and back e.m.f., requires a slightly more complicated circuit. This circuit also produces an inversion of the logic input.
comes the base resistor of the second.

## DIODE PROTECTION

Note the diode across the coil of the relay. This is to protect the transistor from the large back e.m.f. which is produced when the
transistor switches off.
The current which flows when the transistor is on cannot disappear instantaneously as the inductance of the relay coil acts as a store of current. If the diode is not there the voltage at the collector would rise to many tens of volts and could destroy the transistor.

The diode clamps the voltage at the collector to the supply voltage plus one diode forward voltage drop $(0.7 \mathrm{~V})$. This does mean that the energy in the relay takes longer to be dissipated but we are not interested in this time for our particular application.

One further point to note about this circuit is that it inverts the sense of the input. When the input is high (near the positive supply) the relay is off and when the input is low (near the 0 V supply) the relay is on.

## EXPERIMENT 8.1: LED DRIVER

Components needed: $680 \Omega \frac{1}{4} \mathrm{~W}$ resistor (2 off), $10 \mathrm{k} \Omega \div W$ resistor (2 off), BC108 transistor (2 off).

In this simple experiment we use a switch to provide the input signals for a light emitting diode driver which then drives another similar stage. The circuit is shown in Fig. 8.11a and the layout on the Tutor Deck in Fig. 8.11b.

With the switch in one position, l.e.d. D1 is on and l.e.d. D2 is off. The situation is reversed when the position of the switch is changed, showing the inverting action.

Fig. 8.11a


## EXPERIMENT 8.2: SWITCH WITH DELAY

Components needed: $680 \Omega \frac{1}{4} \mathrm{~W}$ resistor, $100 \mathrm{k} \Omega \mathrm{W}$ resistor, $470 \mu \mathrm{~F} 16 \mathrm{~V}$ capacitor BC108 transistor.




Fig. 8.12(a) Circuit diagram of Experiment 8.2. (b) Layout of the components on the Tutor Deck for Experiment 8.2.

This experiment shows how the charging time of a capacitor can be used to produce a delayed switching circuit. The circuit is shown in Fig. 8.12a and the layout in Fig. 8.12b.

The switch is used to switch the l.e.d. on and off but the change in the state of the l.e.d. occurs same time after the change in the state of the switch. The
capacitor has to charge (or discharge) through the base resistor before the base becomes forward biased and the transistor switches on (or off).


In this experiment the simple AND gate using two diodes (see Fig. 4.11) is improved by adding a transistor switch. Note, however, that the diode AND gate is now a diode-transistor NAND gate because the transister switch produces an inversion of the logic levels.

The circuit is shown in Fig. 8.13a and the Tutor Deck layout in Fig. 8.13b.

Fig. 8.13b

Fig. 8.13a

EXPERIMENT 8.3: TRANSISTOR-DIODE NAND GATE

Components needed:
$680 \Omega \frac{1}{4} \mathrm{~W}$ resistor
$10 \mathrm{k} \Omega \frac{\mathrm{W}}{4}$ resistor,
IN4148 diodes (3 off),
BC108 transistor.

Fig. 8.13(a) Circuit diagram of the two input NAND gate as described in Experiment 8.3. (b) Layout of the components on the Tutor Deck for Experiment 8.3.


## LOGIC GATES

When we looked at diodes (Part 4) we saw how a simple logic "gate" could be formed which gave a "true" output only when both inputs were "true". The problem with the gate shown there (Fig. 4.8, Fig. 4.9) was that it could not easily be connected to other circuits of the same type.

Combining the diode gate with a transistor switch produces a gate circuit which can be connected to many other circuits.

Note the extra diode D3 that is needed in the base circuit of the transistor (Fig. 8.9). This is necessary because even with both diodes (D1, D2) fed with a low voltage $(0.2 \mathrm{~V})$ from previous stages, the voltage at the base of the transistor without the extra diode would be $0.2+0.7 \mathrm{~V}$. This is more than enough to forward bias the base-emitter junction and switch the transistor on when it should be off. The extra diode reduces the voltage at the base to 0.2 V so that it is well and truly off.

## DIGITAL INTEGRATED CIRCUITS

Because of the importance of digital techniques, a lot of work has gone into producing switching circuits which act at extremely high speeds yet consume as little power as is possible. The huge number of switches needed in a large digital computer led to the development of integrated circuits which are packages in which a large number of transistors and resistors and diodes are fabricated on a single piece of silicon (a "chip").

## PART 8 QUESTIONS

8.1. The collector to emitter voltage of a saturated transistor is:
a) $2 V$
b) 0.2 V
c) 5 V
d) 0.7 V
8.2. What current into the base is needed to saturate a transistor with hife $=50$ and collector current of 100 mA :
a) 100 mA
b) 1 mA
c) 2 mA
d) 10 mA
8.3. To increase the fan-out of our switch we could use a transistor with different characteristics. Which of the following is most useful, all other parameters being the same:
a) higher gain
b) higher power

The most dominant technology in digital integrated circuits over the last ten years has been Transistor Transistor Logic or TTL for short.
TTL circuits have now become extremely cheap because of the vast quantities in which they are made. They are very fast, switching speeds being measured in thousandths of millionths of a second (nano seconds or ns) rather than microseconds. They are reliable and can drive many other similar circuits. Nowadays improved versions of these circuits have been produced which are even faster and consume far less power.
A typical TTL circuit is shown in Fig. 8.10. Some special features worth noting are the input transistor TR1 with its multiple emitters which act just like the
c) higher collector current capability
d) higher speed
8.4. Switching time of a transistor switch depends on which of the following:
a) type of transistor
b) the circuit used
c) both of these
d) neither of these
8.5. What does TTL stand for:
a) Two transistor levels
b) Transistor timing logic
c) Transistor transistor logic

## PART 7 ANSWERS

7.1. a) 7.2 . c) 7.3 . b) 7.4 . a) 7.5 . b)
diodes in our simple gate, and the fact that the collector resistor of the output transistor TR3 has been replaced by another transistor TR4. This gives the circuit the capability of driving other circuits at higher speeds than would be possible with just a simple resistor.

Other technologies that are popular are CMOS which has the advantage of working with a wide range of supply voltages whilst consuming only miniscule amounts of power and PMOS which is used for cramming vast numbers of circuits onto a single piece of silicon since each transistor can be made very small.

In a later part of this series we will look at CMOS logic and go a little further into its capabilities.

Next month we will consider the transistor as a linear amplifier.


Fig. 8.9. The circuit shown is a logical NAND gate. Only when both inputs are high does the output go low.
Fig. 8.10.(right) The circuit of a TTL two-input NAND gate which performs exactly the same logical function as the circuit of Fig. 8.9 but a great deal more effectively.



Atelevision feature described some of the secret cat and mouse activities in World War II that are now called Electronic Warfare. And from time to time this intriguing term appears in the press, as for instance the announcement by Decca Radar of the order for $£ 2$ million work of "EW equipment" to be installed in new Danish warships. Little else was said and in fact intimate details are never published by the manufacturers or armed forces conconcerned.

This is generally in line with military necessity; but new aircraft, for instance, are often itemised minutely, particularly in American specialist magazines, so why this secrecy about Electronic Warfare equipment and what is it anyway?
A major chess game is in continuous operation between NATO and Iron Curtain electronic design engineers, each side trying to get a step ahead of the other. As made clear in the TV programme, there is nothing new about this, the first instance of the use of EW having been by the Royal Air Force in April 1942

As soon as a radar is switched on, either by an aircraft, ship or ground station, two things follow. First, a suitable receiver can
supply its operator with several of the radar parameters (from which can be obtained considerable information about their source) and second, after an internal delay for assessment, a jamming device can be set in operation which will confuse the radar's operator.

Bearing in mind that radar first came about for warlike purposes, it was natural that these two activities should follow its introduction almost immediately.

## EARLY JAMMING DEVICES

It was a jamming device that first appeared, RAF bombers using it against the German "Freya" early warning radars, and within months an improved version was in use along a 200 mile front. The next year US bombers used "Chaff", thousands of small strips of aluminium foil that were released to provide an effect somewhat similar to dense sea clutter, in order to obscure their own echoes.

Chaff is still in use (a Squadron Leader was recently on record describing an amusingly Heath-Robinson method he thought up for dispensing it from the Jaguar aircraft) but enormous strides have since been made in both monitor-
ing and jamming. Nor is that all anti-jamming measures on behalf of the radar have been developed and the scene soundly set for our international chess game.

Here it is necessary to introduce the accepted EW jargon. The passive side-monitoring-is ESM (Electronic Support Measures); the active side - jamming - is ECM (Electronic Counter Measures); the defence against jamming is ECCM (Electronic Counter Counter Measures); and there is also Elint (Electronic intelligence, of a future tactical rather than immediately useful nature).

The early jammers (ECM) were simply radio transmitters that radiated a continuous amplitude noise-like signal across the enemy radar's frequency band, already deduced by special aircraft with sensitive receivers (ESM). Generally speaking the radar would receive a strong signal that obliterated that of the inevitably weak echo-return.

In the case of Plan Position Indicator displays, the jamming signal served to illuminate a wedge-shaped sector whose angular width corresponded to the beamwidth of the radar aerial. Because the early radars had relatively wide beamwidths an air-
borne jammer could nullify a large portion of the display. The early aerials also had significant sidelobes and this further assisted the jammer by expanding the width of the "whited-out" section.

## COUNTER COUNTER MEASURES

With the Cold War, followed by conflict in Korea and Vietnam, EW continued to develop, particularly ECCM. The airborne jammer had weaknesses that the ground radar could exploit, one being limited power due to limited size and weight. Not so restricted, the ground radar designer had recourse to greater power, defying jamming, and larger aerials with reduced beamwidth and sidelobes.

Another ploy was the use of radar magnetrons which could be rapidly retuned to a different frequency, forcing the jammer either to carry a special operator with radio receiver for monitoring, or to resort to "barrage jamming" across an active band which reduced available jamming power.

Yet another technique was to have two or three different transmitters, so that perhaps three successive radar pulses would all go out on different frequencies. This necessitated a separate jammer for each frequency and the jammer was further embarrassed by the introduction of new types of kigh power valves which could be retuned more quickly and over a broader band. The advent of the digital computer provided new

techniques for processing radar signals that made it easier to detect weak target echoes among jamming.

It might be thought that ECM was now well on top of ECCM but developments in microwave valves were also helping the designers of jammers. For example, valves such as the backward-wave oscillator and voltage-tuned magnetron enabled the jammer to retune more rapidly or to sweep the jamming signal across the known band. The chess game was gathering momentum!

## TRAVELLING WAVE TUBE

The travelling wave tube (TWT) amplifier, a broadband linear device was the next ally of the jammer, offering an entirely different approach in that in company with the frequency memory loop it could be used to deceive a radar rather than overwhelm it.

This combination made it possible to receive and transmit back to a ground radar a copy of its own pulses, however varied these might be by ECCM techniques. They stored the received pulse in a memory device and then retransmitted it, so that the radar first received the weak echo return and then an identical but stronger one. The result was a series of spurious targets, each at a different range. This type of jammer was of particular use against radars used for fire control, in particular the many types of missiles which depend on radar for guidance.
automatic gain control to adjust to this stronger signal (because it assumes it to be the chosen target). The deception jammer then gradually introduces increasing delays before transmitting back. The radars range-gate circuiting begins to track the stronger jammer signal and gradually wanders off from the correct target range.

## BEAM SPLITTING

In order to measure target position within the beam more accurately, modern surveillance radars use a signal processing technique called beam-splitting. This boils down to measuring the point of strongest return because it corresponds to the centre of the beam. A deception-repeater can upset such a technique by using a procedure called inverse-gain. This effects the direct reversal of the normal method of retransmission of the received radar pulse using power proportional to the strength of the received signal.

When the leading edge of the radar beam first (weakly) strikes the jamming aircraft, this retransmits the pulse with higher power; and when struck by the strong signal from the centre of the radar beam, the jammer replies weakly. Thus the beam-splitting process is turned on its head.

## CONICAL SCAN

Some fire control radars, both ground and airborne, employ a conical-scan technique to track their targets. The aerial points in
the last-known direction of the target and its narrow beam is made to scan across a small arc around the aiming point to ensure that the beam will catch the target occasionally, despite the latter's movements.

This small volume scan is usually effected by a high speed, off-centred rotation of the aerial or its feed. The return will be strongest when the beam is directly on target, it will be weak to zero when the target is only partially or not illuminated at all. The radar continuously compares echo strength with aerial position to determine target position (and to redirect the aerial centre in the right direction).

A deception-jammer can disrupt this type of radar every effectively by retransmitting pulses that are modulated with spurious amplitudes at the conical-scan frequency, i.e. displaced in time-phase. This will cause the radar to sense that the target is, for instance, below and to the left of its aerial position when in fact it is above and to the right. The jammer can gradually introduce errors by increasing the spurious time-phase shift.

An alternative radar design which is not vulnerable to this jamming technique is to use a nonscanning transmit aerial and a separate receive only aerial the beam of which does the scanning. This means that the jammer cannot tell the frequency of the radar's receive aerial, forcing the former to sweep its modulation across a band.

The Electronic Warfare Engagement Simulator built by Plessey. Under development for four years, it can simulate, in real time, many different combat conditions including intensive use of ECM techniques by the supposed enemy. Its purpose is to aid development of new ECM techniques that can be used to protect aircraft flying against fire control radar systems.



Spy in the sky. A German Dornier "Kiebitz" tethered rotor platform goes aloft from its support truck. Unmanned, it takes electronic equipment up to $1,000 \mathrm{ft}$ greatly increasing its operation range.

Monopulse radar is another method of frustrating angledeception jammers. This employs the interferometer principle to measure target angle, using two or more separated receiving aerials for training and two others for elevation (the displaced aerials measuring the relative phase, or amplitude, of the echoes to determine angle).

## DOPPLER SYSTEMS

When continuous wave and pulse Doppler type radars were introduced to detect low flying aircraft in ground clutter (see Rapier Missile System, Everyday Eleotronics, April, 1977) special deception procedure had to be developed. The basic principle of these radars concerns the Doppler shift of echoes due to target velocity.

For the first mentioned (continuous wave) the method of deception is to retransmit the signal with a spurious Doppler shift, gradually increasing its magnitude to cause break-up of velocity tracking. Apart from introducing errors into the system's computation of lead angle, velocity gate wander can result in loss of target tracking when the jammer stops its retransmission.
Against the pulse Doppler type, a TWT is used, the jammer introducing a spurious Doppler shift when it retransmits the received pulses.

## AT SEA

Though the foregoing has concentrated on land and air, navies are equally dependent on Electronic Warfare. Ships are loath to use their radars for fear of giving
their presence away to an ESM equipped enemy. When they do, ECCM is employed to the full to confuse the ESM operator.

Chaff is fired from rockets or guns, ships sometimes putting up a modern version of the old smoke screen, behind which they manoeuvre to outwit the detected missiles.

Missiles do not always fly straight, some rise to a great height and plunge down on their target, though the mechanism to ensure accuracy to the necessary few yards beggars the imagination and is of course hideously expensive.
Many types of missiles can be launched from over the horizon, directed by near real-time targeting data provided by reconnaissance aircraft or air/ship-based Elint receivers and data linked to the launching platforms. These can be small fast patrol boats, as mentioned or submarines below the surface. It is also thought that the integration of data obtained by satellite is possible. Probably the best defence is a mixture of weapons (such as guns and anti-
missile missiles), the EW techniques touched on above, and suitable tactics.

## SUPERIORITY BATTLE

Further examples of move and counter move are endless, sometimes one side gaining the ascendancy, sometimes the other. It is obvious why secrecy is vital. Historically, the Soviet invasion of Czechoslovakia was a big plus to them, since their ECM technique was successful in blinding NATO's early warning devices on the frontier. Later the Americans can be said to have won the EW battle against the Russian supplied N. Vietnamese air defence system.
Lastly, in the Israeli-Egyptian "October" war, Electronic Warfare was a key element all round, both at sea and in the air. The Israelis used a pilotless decoy plane that on one occasion was attacked by a total of 32 missiles but returned safely and in a naval engagement the Eilat defeat was reversed by four Israeli fast patrol boats who hit eleven Egyptian ones, all concerned armed with missiles.


Double cake shaped aerial of Decca's EW equipment. Being designed to ff around a mast, it does not compete for the sought after masthead position.

It is a fair conclusion that all radiating electronics can be jammed and any jamming technique can be countered eventually. So an end to Electronic Warfare is not in sight, nor ever seems likely to be.
USING YOUR FRF5 GIFE


Pocket size instrument with pockial and metric scales four impertal $(0-153 \mathrm{~mm})$, plus infuess inches of particular uses gauges electronics conductors.
CAPACITOR/RESISTOR LEAB BENDER. The holes correspond to holes in 0.1 in matrix stripboard or perforated plains leads in general, componen lhan 0.1 should be bent not of the compoin. from the body

When working to published design, insert resistor or capacttor leads into holes spacing, giving the required onent layout as denoled on co diagrams. push component body close Push compone, straighten and to Rule sur leads. Finally frim paraliel th leads about 1 mm from
off both Rule and remove.
BA AND METRIC GAUGEInsert BA AND ME
screw or drill into wide end of wedge-shaped cutout and slide along as far as possible. Scainst calibration is read of screw/dril. leading edge or sol SIZES BA CHEARANCE HOLE Sill be a tight fit
Correct drill will ber correct appriate hole.
in appropr
S.W.G. GAUCE for Sheet metal and connecting wires in the most common sizes useful for 26 s.w.g. especill and wires. matching drills and wires.


THIS article describes a simple car voltmeter using l.e.d.s to give a continuous indication of battery voltage. The instrument can be calibrated over a range of about 11 to 15 volts and is easily constructed and installed in any car whether positive or negative earth.

## CIRCUIT

The circuit is shown in Fig. 1. The LM324 integrated circuit contains four operational amplifiers (op-amps). The inverting terminal of each opamp is connected via R13 to the junction of R14 and D5 (D5 is a Zener
diode, providing a fixed reference voltage of about 6 V ).

## REFERENCE VOLTAGE

Resistors R1, R3, R6, R9 and VR1 form a potential divider chain. The values of these resistors are chosen to give voltages of between 5 and 7 volts at the non-inverting terminals of the op-amp.
The actual values will vary with the supply voltage, that is the battery voltage. Each op-amp will amplify the difference in voltage between its inverting and non-inverting inputs several thousand times.

If the voltage at pin 10 of IC1 is higher than that at pin 9 (which is the reference voltage), then the output at pin 8 will be high ( 12 V approximately) and the l.e.d. Dl will be on.
As the supply voltage increases, more of the non-inverting inputs will be at a higher potential than the reference voltage, hence more l.e.d.s will be turned on. VRI is adjusted to give the range of voltage measurement required.
The diode D6 is used to prevent damage to ICl if the instrument is wrongly connected to the supply. It could be omitted but this is not advised.


The output from the car regulator is fed into the unit. As the output from the generator increases so the voltage going into the sensing switches rises. The l.e.d.s are switched on one by one by the voltage comparators within the voltage sensing switch circuitry. The number of illuminated l.e.d.s gives an indication of the size of the input voltage and hence the battery voltage.

## COMPONENTS

Many components used in this design are not critical and there is a wide range of tolerances in the published values and types.

The variable resistor VRI is a multi-turn preset, although the more usual skeleton type could be used instead.

D6 is any diode which has a reverse voltage greater than 20 V and D5 is any Zener diode with a value between 5 and 6V.

Resistor RI may need to be a slightly different value to compensate for the actual Zener value used.

The l.e.d.s Dil to D4 may be of any type or colour. A different colour may also be used for the last l.e.d. D4 giving a warning of too high a battery voltage, indicating a faulty voltage regulator.

The brightness of the l.e.d.s may be altered by changing the values of R4, R7, R10 and R12.


Fig.1. Complete circuit diagram for the Battery Voltage Monitor.

## COMPONENTS

Resistors

| R1 | 8-2k | R11 $100 \mathrm{k} \Omega$ |
| :---: | :---: | :---: |
| R2 | $100 \mathrm{k} \Omega$ | R12 1 k / |
| R3 | $1 \mathrm{k} \Omega$ | R13 33k |
| R4 | $1 \mathrm{k} \Omega$ | R14 10ks |
|  | $100 \mathrm{k} \Omega$ |  |
|  | $1 \mathrm{k} \Omega$ |  |
|  | $1 \mathrm{k} \Omega$ | See |
| R8 | 100k $\Omega$ |  |
| R9 | $1 \mathrm{k} \Omega$ |  |
|  |  |  |
| All | $\frac{1}{4} \mathrm{~W}$ carb | $\pm 5 \%$ |

Semiconductors
IC1 LM324 quad op-amp i.c.
D1-D4 TIL209 I.e.d. (4 aff)
D5 BZY88C5V65.6 volt 400 mW Zener diode
D6 $\quad 1$ N4001 rectifier diode
Miscellaneous
VR1 20 ks 2 multi-turn preset potentiometer
TB1 2-way screw connector block
Stripboard, 0.1 inch matrix, 17 strips $\times 32$ holes; case, $100 \times 50$

25 mm (optional); connecting wire; screws for fixing unit to vehicle; d.i.1. socket for IC1.

 from the board, see text.



ASSEMBLY
All the components, including the 1.e.d.s, are mounted on a piece of 0.1 inch matrix stripboard, 17 strips by 32 holes. The layout is shown in Fig. 2.

The unit is housed in a small box, $50 \times 100 \times 25 \mathrm{~mm}$. Drill the lid of the box in such a way that the l.e.d.s protrude through the surface and fit tightly into these mounting holes.

The constructor may like to alter these mounting arrangements. One
possibility is to mount the l.e.d.s separately in the case. Another is to drill the dashboard and mount them where they can clearly be seen. fitting the stripboard somewhere behind the dashboard.

The input from the battery is made via a double screw connector block TBI glued on the outside of the case.

## CALIBRATION

Calibration is best done before iustallation. Connect a voltmeter across the input terminals and adjust VR1 so that the fourth l.e.d. Di turns on at about 15 V .

Check that all the l.e.d.s are extinguished at somewhere between 10.5 and 11 V . The value of resistor R1 may need to be altered to achieve this, either $6 \cdot 8$ or 10 kilohms may be necessary depending on the value of D5.

An 18 V supply to feed the current through a 1 kilohm variable resistor will suffice for a power supply to calibrate the instrument. The current consumed will be about 18 mA with all four l.e.d.s alight.

## INSTALLATION

When installing the unit in the car use an auxiliary lead from the ignition switch to provide power. In normal operation D3 and D4 should light when the ignition is switched on. D2 may also light when the engine is running.

If D4 lights up on its own, this indicates that the battery voltage is low and so the battery needs recharging or replacing. If all the l.e.d.s come on, the voltage regulator may be suspect (assuming the instrument has been properly calibrated).

## COUNTER INTELLIGENCE BY PAUL YOUNG

## First Exhibition

In December I visited the "Breadboard Exhibition" and splendid it was too. I do wish that they would rename it. Several people thought it was a Bakery Exhibition.
1 must say it bought back memories. It was held in the Horticultural Hall at Westminster and I remember years ago that one of the first of these exhibitions was held there. It was called the "Radio Hobbies Exhibition".
We had a stand there, and three incidents stick in my mind. My brother had pre-fabricated our display on hardboard. We arrived at the hall and started to erect it. Blunder number one; up rushes a Union Official and told us we had to employ a Union man to do it.
We had no alternative but to agree. As it turned out our chap was quite useful and didn't object to our helping him.
The next calamity was even more worrying. Thirty minutes before the show was officially opened, along comes
an Inspector. He takes one look at our hardboard and says, "That's not fire proof, I can't allow the show to open"
Fortunately in the end he compromised, andsaid that if we painted the backs of the display boards with fire proof paint he would lift the ban. I remember us slapping on black paint as hard as we could go, and the day was saved. Between you and me, the paint we put on would have made it burn better than ever, but as they say, what the eye doesn't see the heart doesn't grieve about.
Finally, when we were packing up to go, 1 carried out a beautiful display case made of quarter inch plate glass. It was raining and I stood it on the bonnet of our van. Slowly, it slid to the edge while I stood and watched hypnotised. then it plunged into the road and shattered into a thousand fragments.
My brother came along a few seconds later, while I was still gazing at the mess in sheer disbelief. His only cryptic comment was, "Ohl dear, what a pity, I would like to have seen itll'

## Component problems

Over two years ago, I forecast that the range of components might be reduced, and 1 pointed out that it was not likely to inconvenience the constructor. This has now happened. Toggle switches for example are reduced to two types, single-pole on/off and double-pole double-throw, but since this covers the whole of our needs it is of no great consequence.

Sometimes we run out of a particular value of quarter-watt resistor so we send a half-watt, This upsets no one and only worries the un-enlightened.
I had a case recently, of a customer returning a small transformer and complaining it was open circuit. I remembered that I had had a similar experience myself, but I refused to believe that both the primary and secondary windings were faulty, so I rubbed the connecting wires with "emery" and re-tested it. Itwas perfect.
These transformers are obviously dipped in a clear varnish which invisibly coats the connecting wires and prevents contact. I mention this in case you come across a similar problem.

## JICK PIUA \& FITIIT...

BY DOUG BAKER


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Price 50p

# Dual Liode Game <br> By A. Russell 

strategy of the machine consists of choosing the resistor across which the maximum voltage occurs. If two or more resistors show the same maximum voltage, one is selected at random.

The move is indicated to the human contestant by lighting two l.e.d.s which point to the row and column co-ordinates of the socket where the machine would like its move made. Of course this requires a measure of integrity on the part of the human contestant because the machine cannot tell when it is being cheated.

A simple machine may be made by using light bulbs, instead of resistors, in the network of Fig.1. The machine moves can then be determined by observing which bulb glows the brightest. It is often difficult to decide which of several bulbs is brightest and so a more sophisticated circuit was developed to automate this decision.

The difficulty here is how to detect the network resistor with the highest voltage across it, bearing in mind that there is no common point in the circuit to measure the voltage with respect to.

## RULES OF THE GAME

The game of Dual Line is played by two contestants on a board marked out with a lattice of interwoven red and blue lines. One contestant (for example blue) has a supply of blue plugs, the other (RED) has red plugs. Each player takes turns to insert a plug into an unoccupied socket on the board (there are sockets at each point on the board where red and blue lines cross).

The object of the game, for the RED player, is to create an unbroken red line from top to bottom of the board. When a red plug is inserted into a socket, the red line through that point is assumed to be completed, and the blue line broken. A winning line may zigzag about all over the board so long as it connects the two opposite sides of the board with an unbroken line.

Similarly the blue player is trying to complete a blue line stretching between the left and right hand sides of the board. Unlike the game of noughts and crosses the game can never end in a draw because one player can only prevent his opponent from winning by winning himself.

## RESISTOR ANALOGUE

The unit decides the moves for one of the players (in this case bluE). It operates on the following principle. A resistor network (Fig.1) corresponds
to the lines of play open to the blue player (lines of play for the RED player are shown on the same diagram as dotted lines). All resistors have the same value.

When blue inserts a plug into a jack socket, the resistor corresponding to that socket is shortcircuited. When red inserts a plug into a jack socket, the resistor corresponding to that socket is open-circuited. Thus when blue completes his line from left to right hand side of the boand and wins, the entire network is short-circuited.

Similarly when RED completes a line from top to bottom of the board the network is opencircuited. These two conditions are detected by the win detecting circuit (Fig.3) and cause the RED wins or blue wins l.e.d.s to light.

## MAKING A MOVE

When the circuit is called upon to decide its next move, a voltage source is applied across the network. The Game.


Fig.1. Resistor analogy network for the Dual Line


Fig．2．Block diagram showing all sections of the game．


In this circuit the emitter－base volt－ age of a silicon transistor（which is the same between transistors at the same temperature）is used as a refer－ ence．Each resistor in the modelling network has a transistor emitter－base junction across it．Those resistors where current could flow in either direction depending on the state of play have two transistors connected in opposite senses．

## BLOCK DIAGRAM

The sequence of events，which takes place when the machine decides its next move，is best visualised by refer－ ing to the block diagram（Fig．2）． When the bLue go button is pressed， row and column lock－out circuits （Fig．4）are reset which in turn send compute signals to the control circuit （Fig．3）．Compute signals cause a variable voltage generator in the con－ trol circuit to apply an increasing voltage to the modelling network （Fig．5）．

## POTENTIAL DIFFERENCE

Voltage continues to rise until the potential difference across one resis－ tor is enough to switch on its asso－ ciated transistor．Sensing lines（ $E, F$ ， $G, H, I$ and $J, K, L, M, N$ ）from the row and column lock－out circuits （Fig．4）detect row and column co－ ordinates of the switched on transis－ tor，light the appropriate l．e．d．s and take away the compute signals which causes the output of the voltage generator to return to zero．

Lock－out circuits hold the row and column l．e．d．s on until bLUE go is pressed again．If either bLUE or RED has won，no voltages will be developed across any resistor in the modelling network．The output of the control circuit will rise until the win detect－ ing circuit comes into play and identi－ fies the winner．

## COMPONENTS 䈍事讙

Resistors

| R1 | $10 \mathrm{k} \Omega$ | R18 | $10 \mathrm{k} \Omega$ |
| :---: | :---: | :---: | :---: |
| R2 | $47 \mathrm{k} \Omega$ | R19 | $100 \mathrm{k} \Omega$ |
| R3 | $1 \mathrm{k} \Omega$ | R20 | $2 \cdot 2 \mathrm{k} \Omega$ |
| R4 | $220 \Omega$ | R21 | $2 \cdot 2 \mathrm{k} \Omega$ |
| R5 | $10 \mathrm{k} \Omega$ | R22 | $100 \mathrm{k} \Omega$ |
| R6 | $2.7 \mathrm{k} \Omega$ | R23 | $2 \cdot 2 \mathrm{k} \Omega$ |
| R7 | $27 \mathrm{k} \Omega$ | R24 | $100 \mathrm{k} \Omega$ |
| R8 | $10 \mathrm{k} \Omega$ | R25 | $2 \cdot 2 \mathrm{k} \Omega$ |
| R9 | $10 \mathrm{k} \Omega$ | R26 | $10 \mathrm{k} \Omega$ |
| R10 | $100 \mathrm{k} \Omega$ | R27 | $100 \mathrm{k} \Omega$ |
| R11 | $2 \cdot 2 \mathrm{k} \Omega$ | R28 | $2 \cdot 2 \mathrm{k} \Omega$ |
| R12 | $10 \mathrm{k} \Omega$ | R29 | $10 \mathrm{k} \Omega$ |
| R13 | $100 \mathrm{k} \Omega$ | R30 | $100 \mathrm{k} \Omega$ |
| R14 | $2.2 \mathrm{k} \Omega$ | R31 | $2 \cdot 2 \mathrm{k} \Omega$ |
| R15 | $10 \mathrm{k} \Omega$ | R32 | $10 \mathrm{k} \Omega$ |
| R16 | $100 \mathrm{k} \Omega$ | R33 | $100 \mathrm{k} \Omega$ |
| R17 | $2 \cdot 2 \mathrm{k} \Omega$ | R34 | $2 \cdot 2 \mathrm{k} \Omega$ |

## Capacitors

C1 $50 \mu \mathrm{~F} 15 \mathrm{~V}$ elect．
C2 $50 \mu \mathrm{~F} 15 \mathrm{~V}$ elect．
Semiconductors


## Miscellaneous

SK1－SK25 3.5 mm jack sockets with one normally closed switch（ 25 off）
S1 single－pole on／off toggle
S2．miniature push－to－make single－pole switch
B1 9 V type PP3
Stripboard： 0.1 inch matrix， 36 strips $\times 51$ holes（ 3 off）； 6 metres of ten－way ribbon cable；PP3 battery connector；size 10 knitting needles，one set plastic one set metal，for pegs； 7 mm diameter wood dowelling and red and blue enamel paint for pegs；materials for cabinet；three 4BA nuts and bolts with spacers for mounting circuit boards．


Fig.3. Complete diagram of the control and win detecting circuit. This is built on board A .

## CONTROL CIRCUIT

The complete circuit of the control and win detecting circuit is shown in Fig. 3.

When the blue go button is pressed 9V is applied via diodes D1 and D2 to lines $A$ and $B$ of the row and column lock-out circuits. This resets the system and in turn applies 9 V back to the control and win detecting circuit via the diode OR gate D3 and D4.
Because of the C-R network (made up of R1 and CI), the voltage at the base of TR1 rises exponentially to-
wards 9 V . This rising waveform is buffered by the emitter follower stage formed by TRI and fed into the modelling network (Fig.5).

If nobody has won yet, the voltage across the modelling network continues to rise until the voltage across one of the resistors in the network is sufficient to cause the associated BC108 transistor (TR16 to TR49) to conduct. Current is then fed from this transistor, through two isolating diodes, onto two sensing lines from the lock-out circuits.

## WIN OR LOSE

If RED has won, the modelling network is open circuit and the emitter voltage of TR1 is free to rise toward 9 V . When this voltage reaches 5.8 V , D5 starts to conduct and feeds current into the base of TR3 which switches on and lights the RED WINS l.e.d. D7.

Similarly, if blue has won, the modelling network is short circuit and the current through TRI rises until about 3 mA is flowing at which time the voltage drop across R4 is about 0.7 V ; TR2 switches on and the BLUE wins l.e.d, D6 is lit.

## LOCK-OUT CIRCUITS

The modelling network is flanked by two lock-out circuits, one for the rows, one for the columns, and their function is to record and display the network element with the highest voltage across its terminals.

When the bLue go button is pressed, line $A$ or $B$ (depending on whether we are considering row or column lockout circuits) is taken to +9 V (see Fig.4). This energises TR9 (or TR15) which switches off TR4, 5, 6, 7 and 8 (or TR10, 11, 12, 13 and 14). At the same time compute signals are applied to the control and win detect circuit which in turn applies a rising voltage to the modelling network.

Assuming neither player has won, one of the elements in the modelling network switches on causing a volt-


Fig.4. Complete diagram of the lockout circuits. Note that two of these are required, one for the rows and one for the columns. Bracketed component references are for the column lockout circuit and unbracketed ones for the row lockout circuit. These are both built on board B.

## HOW IT WORKS



In this electronic game of Dual Line, the machine always uses blue pegs and the human contestant red pegs.

Assuming red moves first, the appropriate colour peg is taken and placed in the selected position. The bLUE GO button is then pressed. This causes the analogue computer within the unit to work out the best countermove and indicate its position by lighting up two l.e.d.s on the front panel.

The player then inserts a blue peg in the position given and the whole process is repeated again and again until there is a complete line of pegseither blue or red.

When this happens, the machine works out who has won the game and lights up the correct WIN I.e.d.
age to appear on its respective sensing line. When this happens, the associated transistor in the lock-out circuit switches on illuminating the l.e.d. in its collector line.

To stop other transistors being activated at the same time, a set of
inhibit lines are provided. These are initially all at 0 V . When one transistor turns on it immediately clamps its associated inhibit line to +9 V which turns off TR9 and prevents any of the other lock-out transistors from turning on.

## MODELLING NETWORK

The complete circuit of the modelling network is shown in Fig.5. You will notice that there are two types of circuit elements. However, both essentially consist of a resistor connected across a semiconductor junction, the ends of which are connected into the matrix. There are also two sensing outputs $X$ and $Y$ each connected to the sensing lines from the lock-out circuits.

When the applied voltage across the network is sufficient to turn the transistor on in a particular element, these outputs go high energising their respective sensing lines. This causes the lock-out circuits to register this condition and indicate the next move.


## CIRCUIT BOARDS

Most of the components in the Dual Line Game are mounted on three circuit boards. These are all identical pieces of 0.1 inch matrix stripboard 36 strips by 51 holes and full details of component mounting and drilling are given in Figs. 6, 7 and 8.

Construction should start with the control and win detecting circuit board (Fig.6). It will be noticed


Fig.5. Circuit diagram of the modelling network. Note that this is made up of identical network elements, the circuit diagrams of which are shown on the right. This is built up on board C .



The left hand photograph shows a view of the top of the lockout board. Note the use of ribbon cable for off-board connections. The other photo shows a view of the modelling network board. Note the circular shaped plastic encapsulated BC108 transistors and copper bus wires.



Fig.9. Wiring of the rear of the front panel.


Fig.10. Interior views of the cabinet. The diagram lett, shows the circuit board mounting arrangement and the one on the right shows the battery mounting clip.
immediately that only half the board area is used. However making all circuit boards the same size, means that mounting them in the cabinet is a lot simpler and this is what has happened here.

Multi-way ribbon cable is used for front panel connections both on this board and the others as this makes lead identification much easier and keeps the wiring reasonably neat.

To make it easier to keep track of interboard wiring, a simple system of board location identification has been adopted. The first letter refers to the board itself, and the rest of the letters and numbers refer to the strip/hole co-ordinates of a particular location in the usual way.

For example location CAA17 refers to location AA17 on board C (the modelling network board).

The lock-out circuit board (Fig.7) can be tackled next. Both row and column lock-out circuits are accommodated on the same piece of stripboard and apart from C2 are identical.

## NETWORK BOARD

Finally the network board can be assembled. A quick glance at Fig. 8 will confirm that this consists of a large number of separate small circuits each with two wires going to a socket on the front panel and one each to the appropriate row and column lock-out circuit.

Bus wires made up of thick gauge copper wire are mounted about 7 mm clear of the board from one side to the other to provide common returns for the lock-out circuit board input lines. This can be seen clearly in the layout diagram.

You will see that plastic encapsulated BC108 transistors have been used in the prototype. The more familiar TO-18 types can be used instead but take care that their cans do not short circuit to other parts of the circuit.

Component identification is made very simple by the fact that all transistors, all resistors and all diodes are of the same type and value. On the other hand, lead out wires from each section should be made easily identifiable so that they can be connected to the correct socket on the front panel. Care must also be taken to get the polarities of circuits round the edge of the board correct.

## CABINET AND FRONT PANEL

In the prototype the cabinet (dimensions $200 \times 150 \times 70 \mathrm{~mm}$ ) was made up from $75 \times 15 \mathrm{~mm}\left(3 \times \mathrm{I}_{2}\right.$ inch $)$ softwood with a hardboard bottom.

The front panel is made up from a piece of 3 mm coloured Perspex. Great care is needed in drilling this material and a hand powered drill is best. Red and blue lines are made
with coloured drafting tape and captions with rub on lettering protected by clear sticky tape.

The interior layout can be seen both in Fig. 9 and in the accompanying photographs. It is sensible to mark out and drill all the holes in the Perspex panel before mounting any components and it is also a good idea to check each jack socket. Some may need their contacts adjusting and this is easier to do before they are attached to the front panel.

A battery holder can also be made from a suitably shaped piece of plastic (or metal) and glued into position in one corner of the box (see Fig.10).

The completed circuit boards are mounted one above the other on three 4 BA bolts which have been fixed with Araldite to the floor of the case. Spacers can be made up from old ball point pen cases.

Once interwiring between the boards is complete, they can be fixed in position using 4 BA nuts, and wiring to the front panel completed. The unit is now ready for testing.

## BLUE AND RED

To complete and test the Dual line Game a set of plugs (or pegs) is required, 12 for RED and 12 for blue.

The red pegs are made up from size 10 plastic knitting needles. A 25 mm section of needle is cut and Araldited to a 7 mm long section of 7 mm diameter wooden dowel. The pegs are finished off by giving the top a couple of coats of red enamel.

The blue pegs are made up in a similar way, the only difference being

that metal knitting needles should be used, sanded down to ensure a good contact in the sockets.

Testing and setting-up the equipment can now proceed.

## TESTING

When the circuit is first switched on it should draw about 10 mA from a 9 V battery and two l.e.d.s should be lit, one row and one column indicator. Pressing blue go will cause the l.e.d.s to flash briefly, during which time current consumption will rise to about 25 mA . A row and column l.e.d. will again be lit, probably different.

The 2.2 kilohm resistor on the modelling network board, board C, in circuit LG (R63) should now be increased in value (to about $3 \cdot 3$ kilohms) until every time blue go is pressed D22 and D63 come on (no jack plugs should be in the board for this adjustment). This ensures that when the circuit is given the first go in a game it will choose the centre position which is the best first move.

Each position on the board can then be tested by forcing the machine to choose every position in turn. This is done by making a line of red pegs from top to bottom of the board, broken only at the point you wish the machine to choose.

Obviously, if the circuit is working correctly, it must choose to place its plug in the gap this being the only
move which can prevent immediate defeat. Any fault shown by this test must be traced back from the front panel to modelling network board to lock-out circuit board until the error is found.

After passing this test the machine is ready to challenge you to a game. If you give it the first move you can expect to lose the game on most occasions. (Only one line of play has been found that can defeat it when it has first move.) When you take first move it has less chance of winning, but will if you make a bad mistake.

The advantage of the first player can be reduced if it is agreed that the first move may not be in the centre position. Under these rules the machine cannot calculate its first move, but it plays rather well if it is give position $M G$ as its first move.

## COMPONENTS

This circuit uses quite a large quantity of rather a small range of common components. For this reason it pays to shop around to find the best value. The advertisements in this magazine are a good place to start.
The semiconductors used in this circuit are not critical, any small signal silicon npn transistor will do instead of BCl 108 and any silicon pnp instead of BC557. The TLL220 and 1N4148 diodes may be replaced by any reasonable equivalent. It is not recommended that manufacturers "fall outs" be used in this circuit.

This article has described an analogue electrical resistance network which calculates moves for a simple game. There are similar resistance analogues for some other games. The processing power of these and more complex networks containing inductors, capacitors and other elements are overlooked in these days of digital computers.


## By ADRIAN HOPE

## Protective Words

Specialists in any area tend to talk their own particular language. This is partly because it is all too easy for an "expert" in electronics, hi fi, video or computers to forget that others are less expert in that fleld. But it is also a defence mechanism; a protective moat of words to ward off critics.
It isn't just electronics specialists who talk in jargon. Professionals, such as doctors, solicitors and accountants are past masters at the art of staying "one up" by using long words understandable only to other professionals in the same field. (An animal vet once told me that a cat which licked off its fur had "chronic lick. titis'"). Some shop assistants, especially in the electronics area, play exactly the same game, assuming that no one will want to risk losing face by admitting that they don't understand a technical term.
Don't fall for it. If you are being sold something, whether it's advice from a professional, education from an instructor or equipment from a shop, you are entitled to understand what is being said. If you don't understand a technical term when it's thrown at you, don't try and disquise the fact. Instead own up and ask for an explanation.

As often as not you'll find that those who use the most exotic technical terms, and gabble them the fastest, don't really understand what they mean. It is they who will be embarrassed, not you, if you ask for an explanation. If you don't believe me, try it just once when you are next shot a plece of jargon by an expert

## Headphone Stereo

This train of thought was triggered by the realisation that it still isn't sate to write, even in the hi fi press, about "binaural" or "dummy head" stereo without also adding a brief explanation of what this technique involves. The idea of binaural stereo reproduction, which provides a very wide spread of sound from a two channel recording heard over headphones, dates back to the early part of the century and long before anyone succeeded in reproducing a stereo spread of sound from a pair of loudspeakers.
When you think about it, binaural recording is the obvious way to reproduce a
realistic all-round image of sound. In real life we hear sound from all around, and identify the direction of individual sounds, by listening with our two ears. The brain is able to decode subtle directional clues in the sound arriving at each ear because the size and shape of a human head introduces phase, time and level differences into similar sounds arriving at both ears.
In theory, if you put a microphone in each ear of a human stereo head and record sound picked up at each ear on a two channel stereo recording system you will capture an accurate replica of the original sound field. You can achieve a similar effect by building a dummy head of the same shape, size and mass as a human head.
For reproduction you simply feed sound from the left recorded channel into the left side of a pair of stereo headphones, and feed the sound from the right channel into the right headphone. This gives each ear the artificial equivalent of what the ears of the recording head heard. The brain then decodes the phase, time and level difference to produce a mental illusion of the original sound field.
You'll notice that I used the phrase "in theory". A binaural (two-eared) or dummy head stereo recording system can work remarkably well but it isn't quite as straight forward in practice as it sounds on paper.
A microphone, which is an electromechanical transducer and not flesh and bone, obviously can't accurately imitate the audio characteristic of the human inner ear, especially when it's fixed outside in the ear lobes. Likewise a headphone unit, which is another electromechanical transducer and is clamped over the outside of the ear lobe can't completely accurately mirror the sound of nature.
In practice the main audible shortcoming of binaural sound reproduction is what has become known as "front back ambiguity". A sound recorded from the centre in front of the recording head may sound on replay as if it is coming from behind, and vice versa.

This is why many binaural recordings cheat a little by giving the listener an artificial fix on the sound field. For instance someone on the recording conveniently announces that they are "standing in front". It also has to be said that some binaural recordings are technically
much better than others in this respect and recordings that produce a good illusion of the sound field for one listener may not sound so realistic to another.

## Bad Image

Why hasn't binaural sound recording ever been a commercial success? In short it's anti-social and inconvenient to listen on headphones. So engineers soon lost interest in headphone stereo after they had devised a system of producing a stereo spread of sound from a pair of loudspeakers in a room.
Unfortunately a binaural stereo recording will not produce good stereo from loudspeakers. Moreover, a conventional stereo recording intended for loudspeaker reproduction (which in effect means virtually every commercially available recordings) will not produce a true surround of sound when heard on headphones. In fact an ordinary stereo recording usually sounds as if it is coming from inside the listener's head when heard on headphones.

When a binaural recording is played over loudspeakers the stereo image is muzzy and diffuse unless the listener sits in just the right position so that only sound from the left speaker reaches the left ear and only sound from the right speaker reaches the right ear. In fact this neatly sums up the basic difference between the two types of stereo.

With binaural stereo the sound for each ear is at all times kept separate: for loudspeaker stereo the sound from the left and right channels mixes and blends in the room after it leaves the loudspeakers. This mix or blend is heard slightly differently by each ear of the listener and the slight differences produce an illusion of natural sound.

## Binaural Sounds

For anyone whose appetite has been whetted by this brief introduction to binaural sound, the BBC occasionally broadcasts a binaural programme and there are several binaural records on the market. But because many record shopkeepers haven't a clue what binaural is all about (another good example of disguised specialist ignorance) they will often file binaural under "quadraphonic" and recommend that the records need to be played on "special" reproduction equipment. In fact all you need is a stereo gramophone system and a pair of stereo headphones.

So where to buy binaural records? A good starting point is Quadramail, the Huntingdon based mail-order company which specialises in specialist discs and usually has a sales stand at major hi fi exhibitions. Also, part-time audio journalist Mick Skeet has produced a range of binaural EP discs which are available at £1.50 each (including postage) from Whitetower Music, 2, Roche Gardens, Bletchley, Milton Keynes MK3 6HR.

Some discs are far more worthwhile than others but if you like rock music l'd suggest AMC 705 (which features a semipro rock band called "Sinner") and if your musical tastes are more conservative l'd suggest AMC 704 (which on one side features some delightfully original organ and trumpet music recorded in a church).


## PCBS

## VEROBOARD

$\begin{array}{lll}\text { Size in. } & 0.1 \mathrm{in} \text {. } 0.15 \mathrm{in} \text {. Vero } \\ 25 \times 1 & 16 \mathrm{p} & \text { Cutter } 110 \mathrm{p} .\end{array}$ $2.5 \times 3.75$
$2.5 \times 5$
$3.75 \times 5$ $\begin{array}{ll}52 p & 45 p \\ 60 p & 55 p\end{array}$
$3.75 \times 17$ 70 p
250 p
210 Pin insertion

SSpins/100 45p 45p
Fibreglass board: 80p each

## RESISTORS

Carbon film resistors. High Stability
$E 12$ series. 4.7 ohms to 10 M . Any mix: $\begin{array}{llll} & \text { each } & 100+ & 1000+ \\ 0.25 \mathrm{~W} & 1 \mathrm{c} & 0.9 \mathrm{p} & 0.85 \mathrm{p} \\ 0.5 \mathrm{~W} & 20 & 1.5 \mathrm{p} & 1.3 \mathrm{p}\end{array}$
Special development packs consisting of 10 of each value from 4,7 ohms to 1 Meg . METAL FILM RESISTORS very high stability, low noise rated at $\%$ W
$1 \%$. Avallable from 51 ohms to 330 k in $1 \%$. Avalable from 51 ohms to 330 k in
E24 series. Any mix:
$100+\quad 1000+$ $0.25 \mathrm{~W} \quad 4 \mathrm{p} \quad 3.7 \mathrm{p}$
3.5p

## POTENTIOMETERS

Preset vertical or horizontal 100 ohms-
1 M
Rotary $5 \mathrm{~K}-2 \mathrm{M} 2 \mathrm{Log}$ or Lin single. $\quad 30 \mathrm{p}$
Rotar Rotary $5 \mathrm{~K}-2 \mathrm{M} 2$ Log or Lin double 90 p Slide 60 mm travel $5 \mathrm{k}-500 \mathrm{k}$ Log or Lin, single
or Lin, single 60 p caps in red, blue, green, grey, yellow and black. Rotary controls 16 p each. Slide type 12 p each


| TRANSISTORS |  |  |  | $\begin{aligned} & \text { TIP32C } \\ & \text { TIP2955 } \\ & \text { TIP3055 } \end{aligned}$ | 60p |
| :---: | :---: | :---: | :---: | :---: | :---: |
| AC127 | 22p | ${ }_{80}^{8 C 548}$ | 110 |  | 20 |
| ${ }^{\text {ACl128 }}$ | 22 p | BCY72 | 15 p | $21 \times 300$ | 4 p |
| ${ }_{\text {A A }}$ AC176 1 | 22p | BD131 | 40 P | $27 \times 500$ | 150 |
| AD162 | 400 | 80132 | 33 |  | 25p |
| BC107 | 12 P |  |  |  |  |
|  |  | BFY50 | 23p | 2 N 3702 |  |
|  | 120 | BFY | 23 | 2 N |  |
| ${ }^{8 C}$ | 120 |  |  |  |  |
| ${ }^{8 C 109}$ |  |  |  |  |  |
|  | 9 p | MPSA | 168 | 2 N 3 |  |
| 8C177 | 16p | MPAb6 | 500 |  |  |
| BC178 | ${ }^{16 \mathrm{p}}$ | Tip30C | $48 p$ | 2 N 5459 |  |
|  | ${ }^{10 \mathrm{p}}$ | TIP31C | 50p | 2N5777 |  |
| BC184 | ${ }^{100}$ | DIODES |  |  |  |
| BC184L | 100 | 1N91 | 4 p | 1 N 40 |  |
| BC212 | 100 | Na14 | ${ }^{3 p}$ |  |  |
| 212L | 10 p |  | 5 p | B2Y88 |  |
|  | 10p |  | E1.50 p | 100. |  |

## CAPACITORS

High quality foil POLYSTYRENE 22 pf to 100 pf .
TANTALUM BEAD
$0.1,0.15,0.22,0.33,0.47,0.68$
1 \& 2.2uF @ 35 V
63 V working $5 \%$
$4.7,6.8,10 \mathrm{uF}$ © $25 \mathrm{~V} \cdots 10 \mathrm{p}$ each
$22 @ 16 \mathrm{~V}, 47 @ 6 \mathrm{~V}, 100 @ 3 \mathrm{~V}$. 22 p each
$0.001,0.01,0.022,0.033,0.047$ 4 e each $0.068,0.1$ STER
POLYESTER
$0.01,0.015,0.022,0.033,0.047,0.068,0.1 .6 p$ ea
$0.15,0.22$. 8p each $0.33,0.47$
$0.68 \quad$.
0.68
1.0 uF

CERAMIC
Plate type 50 V . Available in E12 series from
0.047 uF . 2 peach

MINIATURE TRIMMERS
Miniature film type, in $1.4 \mathrm{pF}-5 \mathrm{pF}, 2 \mathrm{pF}-22 \mathrm{pF}$,
$2 \mathrm{DF}-22 \mathrm{pF}, 2 \mathrm{DF}-10 \mathrm{pF}$
5
$2 \mathrm{pF}-22 \mathrm{pF}, 2 \mathrm{pF}-10 \mathrm{pF}, 5.5 \mathrm{pF}-65 \mathrm{pF}, 22 \mathrm{p}$ each
RADIAL LEAD ELECTROLYTICS

| 63 V | 0.47 | 1.0 | 2.2 | 4.7 | 10 | $6 p$ each |
| ---: | ---: | ---: | ---: | ---: | ---: | ---: |
|  |  | 22 | 33 | 47 | $8 p$ each |  |
| 100 |  |  |  |  | $16 p$ each |  |

STEVENEON Electronic Components

|  |  | 4023 | 20p | 4054 | 120p | 4502 | 120p |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 4024 | 50 p | 4060 | 120 p | 4507 | 60p |
|  |  | 4025 | 20p | 4063 | 120 p | 4508 | 330p |
|  |  | 4026 | 160p | 4066 | 60 p | 4510 | 80 p |
|  |  | 4027 | $45 p$ | 4068 | 20p | 4511 | 90p |
| 4000 | 20p | 4028 | 85p | 4069 | 20 p | 4512 | 80 p |
| 4001 | 20p | 4029 | 85 | 4070 | 20p | 4516 | 80 p |
| 4002 | 20p | 4031 | 220p | 4071 | 20p | 4518 | 80p |
| 4006 | 90p | 4033 | 150p | 4072 | 20p | 4520 | 80p |
| 4007 | 20p | 4036 | 350 p | 4073 | 20p | 4527 | 90 p |
| 4008 | $95 p$ | 4039 | 300p | 4075 | 20p | 4528 | 90p |
| 4011 | 30p | 4040 | 110 p | 4076 | 90p | 4529 | 150p |
| 4012 | 20p | 4041 | 85p | 4077 | 20p | 4531 | 150p |
| 4013 | $35 p$ | 4042 | 80p | 4078 | 20p | 4532 | 130 p |
| 4014 | 80p | 4043 | 95p | 4081 | 20p | 4538 | 160p |
| 4015 | 80p | 4046 | 110 p | 4082 | 20p | 4543 | 110p |
| 4016 | 30 p | 4048 | 60p | 4086 | $75 p$ | 4566 | 170p |
| 4017 | 65p | 4049 | $45 p$ | 4093 | 60p | 4558 | 120 p |
| 4018 | 90p | 4050 | 45p | 4095 | 110 p | 4559 | 420 p |
| 4020 | 100p | 4051 | 70p | 4098 | 120p | 4581 | 330 p |
| 4022 | 100p | 4053 | 80p | 4501 | 20p | 4585 | 110 p |
|  |  | 741547 | 80p | 74LS126 | 45p | 74LS175 | 90p |
|  |  | 741548 | 90p | 74LS132 | $80 p$ | 74LS190 | 90p |
| 74.500 | 16p | 74LS54 | 22p | 74LS136 | 50p | 74LS191 | 90p |
| 74LS01 | 22p | 74LS73 | 35p | 74 LS 138 | 75p | 74LS192 | 90p |
| 74LSO2 | 16 p | 741574 | $35 p$ | 74LS139 | 75p | 74LS193 | 90p |
| $74 \mathrm{LS03}$ | 22p | 74LS75 | 40p | 74LS151 | 60p | $74 L 5195$ | 90p |
| 74LS04 | 16p | 74LS76 | 40p | 74LS155 | 650 | 74LS196 | 90p |
| 74LSp8 | 22p | 74 LS78 | 45p | 74LS156 | $80 p$ | 74 LS197 | 85p |
| 74 LS 10 | 22p | 74LS83 | 68p | 74.5157 | 70p | 74LS221 | 100 p |
| 74LS13 | 380 | 74LS85 | $85 p$ | 74LS158 | 65p | 741.5251 | 70p |
| 74LS14 | $65 p$ | 74L586 | 40p | 74LS160 | 75p | 74LS266 | 35p |
| 74LS20 | 22p | 74.590 | 40p | 74LS161 | 68p | 74LS290 | 80p |
| 74 LS 21 | 22p | 74LS93 | 55p | 74 LS162 | 80p | 74LS365 | 55p |
| 74. S27 | 28p | 741595 | $65 p$ | 74LS163 | 80p | 74LS366 | 550 |
| 74L530 | 22p | 74 LS 107 | 45p | 74 LS164 | 80p | 7415367 | $55 p$ |
| 741532 | 30p | 74 LS 114 | 40p | 74LS 165 | 80p | 74 LS368 | 55p |
| 74LS37 | 40p | 74LS123 | 80 p | $74 \mathrm{LS173}$ | 135p | $74 \mathrm{LS386}$ | 50p |
| 74LS42 | $60 p$ | 74 LS 125 | 45p | $74 \mathrm{LS174}$ | 95p | $74 \mathrm{LS670}$ | 200p |
| 11 |  | 7442 | 40p | 7493 | 30p | 74157 | 40p |
|  |  | 7445 | 50p | 7496 | 38p | 74164 | 55p |
| 7400 | 12p | 7447 | 50p | 74121 | 29p | 74165 | 55p |
| 7402 | 12p | 7448 | 45p | 74123 | 40p | 74174 | 55p |
| 7404 | 14p. | 7473 | 23p | 74125 | 38p | 74177 | 50p |
| 7408 | 16 p | 7474 | 23p | 74126 | $36 p$ | 74190 | $50 p$ |
| 7410 | 14 p | 7475 | 26p | 74132 | 46p | 74191 | 50p |
| 7413 | 24p | 7476 | 25p | 74141 | 48p | 74192 | 50p |
| 7414 | 39p | 7485 | 55p | 74145 | 48p | 74193 | 50p |
| 7420 | 14 p | 7486 | 18p | 74148 | 90p | 74194 | 50p |
| 7427 | 22p | 7490 | 30p | 74150 | $55 p$ | 74196 | 50p |
| 7432 | 18p | 7492 | 30 p | 74154 | 68 p | 74197 | 50p |



DIN PLUGS AND SOCKETS

|  | plug | chassis <br> socket | line <br> socket |
| :--- | ---: | ---: | ---: |
| 2 pin | $8 p$ | $8 p$ | $12 p$ |
| 3 pin | $12 p$ | $10 p$ | $12 p$ |
| 5 pin $180^{\circ}$ | $12 p$ | $11 p$ | $17 p$ |
| 5 pin $240^{\circ}$ | $14 p$ | $11 p$ | $22 p$ |

$5 \operatorname{pin} 240^{\circ} \quad 14 \mathrm{p} \quad 11 \mathrm{p}$ 22p
JACK PLUGS AND SOCKETS

|  | unscreened | screened | socket |
| :--- | :---: | :---: | :---: |
| 2.5 mm | $10 p$ | $15 p$ | $8 p$ |
| 3.5 mm | $10 p$ | $16 p$ | $9 p$ |
| Standard | $16 p$ | $30 p$ | $19 p$ |
| Stereo | $23 p$ | $39 p$ | $22 p$ |

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$\square$ BARCIAVCARD
8 ACGESS WELGOME
Please add 50p carriage on orders under E15. Official orders welcome
Mail orders to: STEVENSON (Dept EE)
"Contrariwise," continued Tweedle-dee, "if it was so, it might be; if it were so, it would be; but as it isn't it aint. That's logic!"

Try asking the man in the street what he knows about the silicon chip and you may end up feeling like Alice. Indeed the very mention of the "mighty-micro" will probably cause his eyes to glaze over and conjure up visions of Star Trek and a certain Italian car commercial.

## In the Home

In fact the microprocessor is more commonly used than we think although you wouldn't really know unless you looked closely. The Daily Mail Ideal Home Exhibition held last month at Earls Court gave ample opportunity to seek out examples of the "new technology" applied around the home-and we do mean seek out, despite what you may have been led to believe by pre-show publicity.
It is clearly evident from the exhibition that domestic appliance manufacturers seem notoriously slow in adopting new ideas. It's true that the odd one or two washing machines, cookers and sewing machines use microprocessor control, but we are still a long way off complete electronic control-strange, when it is cheaper and more reliable than equivalent electromechanical systems. Perhaps it's not a strong selling point.

Quartz watches are now being produced in China for the growing world market. They use a Japanese electronic assembly but the cases are made locally.

[^1]

Singer Futura 2001 touch sensitive machine
Creda Carefree Electronic Cooker

The one exception to this was Plessey Ltd. Their whole stand was devoted to silicon chip technology and many devices one would expect in the Ideal Home of the '80s were on display. These included a new infra-red remote controlled slide projector, a heat sensitive

## Fly Electric

A new concept in aviation is the all-electric plane proposed by research engineers in Lockheed. It would still use aviation fuel for propulsion but would substitute very efficient electric generators and motors for the present hydraulic systems.

This, says Lockheed, would use less overall power and reduce fire risk because an all-electric system would be lighter and hydraulic fluid is prone to ignition.
Reduction in fuel consumption, first costs and maintenance costs over the lifetime of a medium-sized transport aircraft could be as much as $£ 50$ million. Electronic solid state control would be an essential feature in any new design.
hair dryer and a special set-up designed to illustrate silicon chip manufacturing techniques.

So perhaps next year we can look forward to a true Ideal Home of the future.

## Exhibition

"Hands on Experience" is the title of a series of new exhibitions currently being planned by Eurofairs to appear throughout the country, commencing in Bristol on 27 to 29 June.

The exhibitions aim to introduce the latest home computers, electronic games and remote control equipment/models to potential buyers. For further details write to Eurofairs, 9 Park Place, Clifton, Bristol BS8 1 JP .

## Inta-Stella Encounters

Scientists working on high energy physics are sending each other scientific data through the European OTS communications satellite.

The new scientific data network is called Stella and it enables participating laboratories to exchange data with high accuracy at a rate of a million bits of information a second. This is roughly the same speed as a scientific computer and will obviate the long delays of sending magnetic tapes by post.
Stella will link the Rutherford Laboratory in the UK with other laboratories in Germany, France, Italy, Austria and Ireland.

## CB FINES

In 1979 there were 78 prosecutions for illegal use of CB Radio with another 56
cases pending. This compares with only three prosecutions in 1978 and four in 1977.

The Irish Republic, reversing an earlier decision, is now to legalise $C B$ but the waveband to be used is still in question.

## COMPSTAT'80

The fourth Symposium on Computational Statistics, COMPSTAT ' 80 , is being held at University of Edinburgh from 18 to 22 August.

Papers will include reviews of current methodology as well as descriptions of recent research on new techniques and software. An exhibition featuring the main statistical pachages, computational aids and microprocessors is also planned, including "hands on" demonstrations.

Since Compstat ' 80 coincides with the first week of the Edinburgh Festival accommodation is obviously going to be limited and anyone interested in attending is advised to register early.

Further information and registration forms can be obtained from Compstat 80 , Program Library Unit (Dept $E E$, University of Edinburgh, 18 Buccleuch Place, Edinburgh EH8 9LN.

## On-Line Computers

A computer-controlled voice and data radio link is to be introduced on the St. Pan-cras-Bedford line of British Rail. It is being supplied under a $£ 750,000$ contract by Ultra Electronics and the German company AEG-Telefunken.

London Transport has invested in two extra Honeywell computers for control of times, routes and frequencies of trains to Cockfosters.

## Carfax

Some 450 cars fitted with radios adapted for the BBC's Carfax motoring information service should be on the road by the end of June. The service is only in its trials stages but if proved successful later this year the go-ahead will be given for car radio manufacturers to produce equipment.

Motorists can stay tuned to Carfax or listen to other programmes with Carfax automatically overriding the programme whenever there is a motoring announcement. Additional cost to the motorist for a Carfax radio is expected to be in the region of $£ 5$ to $£ 20$.

## Credit Card Petrol

A pilot experimental scheme in the Norwich area allows motorists to pay for their petrol at the pump by direct debit using their Bar clay cards. The service is available at a number of Esso garages.
Automatic funds transfer is the basis of the cashless society which reduces "mugging" and the risk of robbery where there are large concentrations of cash.

## Video <br> Pirating

Illegal recording which has cost the audio industry millions of pounds is now worrying the video com-
panies who are now researching encoding systems which will make it difficult or impossible to copy a video recording successfully on a home recorder.

## Cheap Video

Jamanese fescarchens are feported to be active in developing techninues to emakle kiritell vileo sigmals is ke recorded on a stansiard anaio alassethe alinough the tape may need to be of zpecial type.
 operate in a similar manner. When the control input is taken high the switch is turned on and a low resistance appears between input and output, this being about 300 ohms.

In the normal rest state the control inputs to gates IC4a and IC4b are high, thus the switches are on and the music is able to pass straight through to the output.

The remaining two gates, IC4c and IC4d have their respective inputs connected together to which the speech signal is applied. Their outputs, pins 9 and 10 are connected to the LEFT and RIGHT channels respectively. Thus when these gates are turned on speech is applied to both channels at the same time. In the normal rest state the control inputs are low.

## AMPLIFIER

An LM382 dual amplifier is used in two stages as a microphone amplifier with a gain of about 20 dB . The first stage consisting of IC2a amplifies the low level signal from the microphone to a sufficient level so that it can pass through the remaining stages with little attenuation. A proportion of this signal is tapped off by the TRIGGER LEVEL control, VR1 and applied to the second stage, IC2b.

The second stage provides a high gain limiting action, amplifying the signal to a more than sufficient level for applying to the d.c. rectifying circuit. The rectifying circuit comprising R4, D1, D2 and C8 provides a d.c. switching level which is used to change the states of IC3a and IC3b.
Normally pin 3 is high, and pin 4 low. We thus have the required conditions applied to the switches to enable them to assume the states mentioned earlier.

When speech is applied through the various stages, the d.c. level causes the high condition at pin 3 of IC3a to go low, thus turning off the music, and the low condition at pin of IC3b to go high thus turning on the speech enabling the speech to be connected through to the output.
When there is no further speech, there is a delay of about half a second before the circuit reverts back to its original state and the music is once more able to pass through.

Power for the unit is provided by two batteries giving 18 V . A 15 V regulator i.c., IC1 is then used to provide a stable 15 V supply to the circuit.


Music is fed into an electronic switch inside the unit. When the microphone is not in use, this music is directed straight through to the amplifier, but when an announcement is made the electrical signals from the microphone are amplified and passed to the electronic switch.
A control signal is also produced from the voice input in the switch control circuit which causes the switch to change over and direct the speech signals to the amplifier.

At the end of the announcement, the switch changes back to its original state and the music will again be heard.


## CONSTRUCTION

The entire circuit is built on a piece of 0.1 inch matrix stripboard 26 strips x 49 holes. The layout for this board is shown in Fig. 2. There is nothing critical about the layout and it could if desired be mounted on a printed circuit board.

Fig. 1. Complete circuit diagram of the Auto Fade Unit.



Fig. 2. Complete wiring details for the Auto Fade Unit. Be extra careful when wiring the links, as a mistake will be dificult to trace later on. Note the four links on underside.


Rear view of the unit showing the phono socket strip and mic. input jack socket.

Also shown in this diagram is the front and rear panel wiring details. The case used was a small Verobox type 65-2520J, with dimensions of $150 \times 80 \times 50 \mathrm{~mm}$, although any similar size case can be used. It does not have to be metal to provide screening for the circuit.

Remember to observe the usual precautions when handling the cmos i.c.s. Sockets were used in the prototype, but providing you are fairly swift with a soldering iron, they could be dispensed with.

Screened lead is used for the input and output sockets. It is not advisable to use ordinary connecting wire as there is a possibility of hum appearing. This is especially true of the input.

## TESTING

After assembly is complete, it is a good idea to check over for errors, particularly because of the large number of link wires. It is easy to miss one or two. If all seems well the batteries may be connected and the unit switched on. If possible check with a voltmeter that 15 V appears across the supply.

Connect a mono audio source to one channel of the muSic inputs and connect the corresponding output to an amplifier. If the circuit is functioning correctly the audio will pass through with little or no attenuation. Do the same with the other channel.

Now connect a microphone to the sPEECH input, start with the trigger level control fully anticlockwise and slowly rotate it clockwise until a point is reached where the circuit switches, and the speech appears across the output. You may find that after speech is removed the circuit tends to oscillate very slowly, not switching off cleanly. If this happens, it is probable that VR1 is


Layout of components on the completed circuit board for the Auto Fade. We strongly recommend the use of i.c. sockets as the i.c.s can easily be damaged during soldering.


## 

## Resistors

R1 $27 \mathrm{k} \Omega$
R2 $10 \mathrm{k} \Omega$
R3 $27 \mathrm{k} \Omega$
R4 $470 \mathrm{k} \Omega$
All $\ddagger$ W carbon $\pm 5 \%$

## Potentiometer

VR1 47k lin. carbon
Semiconductors
IC1 $\mu \mathrm{A} 781515 \mathrm{~V}$ voltage regulator
IC2 LM382 dual preamplifier
IC3 CD4011 quad 2-input NAND gate
IC4 CD4016 CMOS quad bilateral switch
D1, 21 N914 silicon diode (2 off)

## Miscellaneous

S1 miniature single-pole toggle
B1,29V PP3 battery (2 off)
SK1 standard jack socket
C1 $470 \mu \mathrm{~F} 18 \mathrm{~V}$ elect. C5 $0.1 \mu \mathrm{~F}$ polyester
C2 $0.1 \mu \mathrm{~F}$ polyester
C3 $0 \cdot 1 \mu \mathrm{~F}$ polyester $\mathrm{C} 74 \cdot 7 \mu \mathrm{~F} 18 \mathrm{~V}$ elect.
$\mathrm{C} 422 \mu \mathrm{~F} 18 \mathrm{~V}$ elect. $\mathrm{C} 84.7 \mu \mathrm{~F} 18 \mathrm{~V}$ elect.

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SK2-SK5 phono sockets (4 off)
Stripboard 0.1 inch matrix 26 strips $\times 49$ holes; Verobox type 65-2520 J $150 \times 80 \times 50 \mathrm{~mm}$ or similar size case; 14 pin i.c. sockets ( 3 off); small knob; miniature screened cable; stranded connecting wire.
set too high or that the value of R4 is too high. Rotate VR1 until the circuit switches off. This must be done slowly. Then recheck that, with speech applied, the circuit switches.

Fig. 3 . Example of how to connect the unit into an audio system. This is for
guidance only-there will probably be variations with the equipment used.
At this stage you may notice that the microphone signal is somewhat larger than the music signal. If this is the case then the value of R2 should be increased

The completed Auto Fade with top cover removed.
Note the careful positioning of the two batteries.
until the two signal levels are roughly the same.
If the tests are satisfactory the unit is then ready for use. Remember though that the level control is only adjusted for that one particular microphone, if others are to be used then the control needs to be readjusted each time.

## IN USE

The reader will probably have his own ideas on how to connect the unit into an audio system, one or two points however should be observed.

As the unit is in effect a changeover switch, both the microphone output and the music source output should be suitable for the same amplifier input. This means that if for example a magnetic cartridge is used, it will need some form of pre-amp and equalisation before being fed into the autofade unit.
The diagram of Fig. 3 shows one suggested arrangement.

The completed Auto Fade ready for use.


Both the Burglar Alarm Module and Opto Alarm projects earlier in this series have utilised a miniature audible warning device as the alarm tone generators. Being solid state, these units have the advantage of low current consumption (about 15 mA ) as well as small size when compared with electromechanical buzzers.

The project described here is a two-transistor circuit which produces an alarm tone via a miniature 8 ohm loudspeaker. It can be used in place of the audible warning device mentioned above.

The Audible Tone Generator is easily constructed from readily available parts and is ideal for beginners. This device incorporates a preset control enabling the frequency of the tone to be varied to a certain extent.
It can be used in any project which operates from a 9 volt rail; 12 volts must be considered the absolute maximum supply voltage for the unit. Being a direct replacement for the audible warning device described earlier, this unit also draws in the region of 15 to 20 mA .

## CIRCUIT DESCRIPTION

The circuit diagram of the Audible Tone Generator is shown in Fig. 1. Regular readers will recognise this as a unijunction transistor oscillator, TR1, followed by a single-transistor amplifier, TR2.

TRI has three terminals which, unlike a "normal" transistor, are designated emitter, base 1 and base 2. The transistor is comnected to form a relaxation oscillator operating at audio frequencies. When power is applied to the circuit, the capacitor CI will charge up through the resistors R1 and VR1. Eventually a point is reached when the transistor will switch over, and Cl will discharge itself into the emitter of TR1 and R4 to 0 V .

TR1 will then switch back over to its high impedance state, and so Cl starts to charge up again. The whole cycle repeats itself, a constant stream of pulses being present at base 1

By adjusting the timing components (R1, VR1 and C1), the frequency of the pulses can be altered. In fact by varying VR1 this particular unit was measured as operating between 100 and 350 Hertz.

The output pulses are coupled to TR2 by R3. This second transistor amplifies the audio tone generated by TR1 and associated components, and directly drives the loudspeaker LSI. The speaker used in the prototype was a 45 mm diameter type, although a larger size (e.g. 65mm) could be used.

Capacitor C2 decouples the power supply and enables the peak current requirements to be met.


AUDIO TONE GENERATOR


## CIRCUIT BOARD

The circuit can be built onto a standard-sized piece of $0 \cdot 1$ inch matrix stripboard measuring 10 strips x 24 holes, Fig. 2. This particular size is available from suppliers, but it can if necessary be cut out from a larger piece.

Commence by drilling two 6BA clearance holes in the positions B3 and $I 3$ as shown. These will enable the completed board to be fixed down with 6BA hardware and spacers.

There are six breaks to be made in the copper strips, in order to prevent the mounting hardware from shorting together adjacent copper strips. After doing this, the components may be soldered in.

The transistors being semiconductors of course are sensitive to heat, so it is wise to use a heatshunt on the transistor leads being soldered.

Transistor TR2 was fitted with a push-on TO-18 heatsink purely as a precaution, as the transistor temperature did rise slightly after prolonged use with a 12 V supply rail. The heatsink can be omitted if it is not available, but if one is used, it must be
fitted to the transistor before the device is soldered into position. Otherwise the possibility exists of the leads of the transistor being accidentally bent or deformed, because of the pressure required to push on the tight-fitting heatsink.
The only points to watch are correct orientation of the transistor leads, plus proper polarisation of the electrolytic capacitor C2. Four flying leads connect up the loudspeaker switch and the power supply; these can be made up from standard multipurpose hook-up wire (preferably stranded) and can either be soldered to terminal pins or straight to the circuit board.

## INSTALLATION

The completed unit can either be installed in a separate box (together with the loudspeaker) or it can be fitted into the same case (if there is any room) as the equipment from which it is to be powered.

The miniature loudspeaker is carefully glued to the inside of the case: a smear of Uhu or similar adhesive around the rim of the speaker should be enough to fix down the loudspeaker, but do not apply any glue to the actual cone of the speaker. A series of holes should be made in the case where the speaker is to be positioned, so that the sound of course can get out.
Testing the generator comprises applying 9 volts (from a battery) and setting VR1 until a tone of the desired frequency is obtained.

Next Month: Voltage Converter


Fig. 1. Circuit diagram for the Audio Tone Generator.


(above) The completed circuit board. The use of a heatsink for TR2 is optional.

## COMPONENTS

Resistors
R1 $22 k \Omega$
R2 $100 \Omega$
R3 $100 \Omega$ All $\frac{\frac{1}{4}}{} \mathrm{~W}$ carbon $\pm 5 \%$
R4 $68 \Omega$
Capacitors
C1 $0.15 \mu \mathrm{~F}$ polyester type C280 C2 $100 \mu \mathrm{~F} 10 \mathrm{~V}$ elect.
Semiconductors
TR1 TIS43 n-channel unijunction transistor
TR2 BC108 npn silicon

## Miscellaneous

VR1 $47 \mathrm{k} \Omega$ miniature horizontal preset
LS1 miniature 8 -ohm moving coil loudspeaker
S1 push-to make pushbutton switch
B1 9V type PP3 or in-built supply-see text
Stripboard 0.1 inch matrix size 10 strips $\times 24$ holes; push-on TO-18 heatsink (optional, see text); 6BA mounting hardware; case to suit; PP3 battery clip.

Approx. cosf
Guidance only
82.20

Fig. 2. Layout of components on the topside of the circuit board together with interwiring to loudspeaker, on/off switch and battery connector. Note that there are six breaks to be made in the copper strips on the underside.


The completed Audio Tone Generator. The method of housing the finished unit is left to individual choice.

By Pat Hawker, gzva

## Power for Radio

When most of us think of radio, telecommunications and broadcasting, we think in terms of electronics rather than electric power. Yet, in practice, as soon as one moves equipment away from the electric supply mains, the basic problem of providing electricity looms very large indeed.
A simple radio receiver can be run from dry batteries, although the cost, in terms of power delivered, is several hundred times as much as from supply mains.
Rechargeable cells, particularly the convenient nickel-cadmium (nicad) units, are the most popular and practical means of powering small amateur-radio transceivers, but this assumes moderate loads and regular access to the mains. What about equipment that needs to run, day-In day-out, often unattended, in remote loacations?
Many exotic, high-density power sources are regularly announced yet relatively few of these prove in the end to be as cost-effective as the traditional methods. Remember when in the sixties "fuel cells" were so confidently predicted as being about to sweep normal large capacity storage batteries of the market? Somehow Leclanche dry batteries and lead-acid car batteries seem to have survived
Some ingenious power sources were exploited during World War II for clandestine and infiltration radio links: stationary bicycle generators (an energetic "rider" can develop over 100 watts for a limited period of time); the more sophisticated form of this approach was a "beach chair" version that could be folded up and carried as a back-pack; wind-generators of course; thermo-couple chargers using the Seebeck effect of dissimilar junctions (in the 1930s this was used to permit radio recelvers to be "run from the gas mains ${ }^{\prime \prime}$ ); there was even a small steamdriven generator with a boller that could be suspended in a charcoal-burning brazler, coupled to a twin-cylinder engine which, with a steam pressure of 30 to 35 lb , could charge a 6 V battery at about 4 A .
More recently the solar generator has found increasing application, not only for space satellites, but also for many purposes in remote, sunny areas. Even the early form of zinc-air batteries (once used to operate door bells in many houses) has been very successfully revived and used in South Africa to run remote radio repeaters; the trick is to use the disposable but very large zinc-air battery to keep large nicad cell's (better able to supply intermittent loads) fully charged.
Many systems have been used for rural telecommunications in Australia. For example Telecom Australia has over 50 solar power systems in use and these are proving cost-effective for loads of up to about 200 watts.

This organisation has also used thermoelectric generators, mostly for powers of under 60 W . Originally these were heated from propane gas but suffered from high costs and poor reliablility; however a new range of improved units, delivering up to 120 W , with hermeticallysealed thermopiles and all-stainless-steel construction, are now being developed for locations where there are adequate supplies of propane, butane or natural gas.

## Wind Generators

Wind generators have been found to need careful design since their performance in low wind conditions is usually much more important than their ability to deliver considerable power in strong winds. Current work is on a 5 kW design that is expected to "cut-in" and deliver useful power whenever the wind exceeds a modest $10 \mathrm{~km} / \mathrm{hour}$; these are expected to support (with the aid of storage batteries) ten times the load of earlier 2 kW units.
The Americans are thinking of very large wind-generators to feed power into the electric grid; some of these may be in the form of tall, elegant windmills. Some people, however, are concerned lest the large revolving blades upset television reception over large areas, in much the same way as reflections from aircraft can do, or as happens in areas where there are large moving cranes producing a changing pattern of multipath ("ghosty") pictures.
Late last year, an experimental American medium-wave broadcasting station, WBNO at Bryan, Ohio, began using a large solar array of 33,600 photovoltaic cells that can generate 15 kW from sunlight; they are used in conjunction with four large unitized diesel tubular batteries.

The rising costs of electric power, even from the mains, is already worrying American broadcasters; a couple of years ago an engineer at one of the very high power u.h.f. stations told me that his station had given up late night TV as it was too expensive: not the cost of programming but the power bill!

## Lines on TV

The Science Museum in London is currently staging a special exhibition (open until September) called. "The Great Optical Illusion" and marking 50 years of television broadcasting. The anniversary is based a little tenuously on one of the many events surrounding the early Baird 30 -line mechanical system which helped pave the way for good TV pictures but which was never itself of real home-entertainment value.
The poor performance was due to poor synchronisation as well as the poor resolution of the pictures; though mechanical systems of televison subsequently
proved capable of providing very acceptable pictures on 180 lines, 240 lines and even (with the Scophony large screen system) on 405 lines. The limit of 30 lines was due to the restricted bandwidth available on medium waves; it was only when television moved up to v.h.f. that real progress became possible.
in the UK we all recognise the extremely valuable work done by the Marconi-EMI research team leading to the 405 -line system that opened (alongside the Baird $240-$ line sequential system) in November 1936 at Alexandra Palace. But I do feel that we tend, a little chauvinistically, to overlook the important work during the thirties in France and Germany.
A 180 -line service was officially opened in Berlin in 1935 and soon included the use of electronic cameras; at the same time many firms in Germany were working on 375 -lines and a public service opened in 1938 using 441 lines.
The Germans televised the 1936 Olympic Games using both mechanical and electronic cameras and, at the Berlin Radio Exhibition that year, receivers for 180 and 375 lines were exhibited by Fernseh, Telefunken, Loewe, Lorenz, Philips and Te-Ka-De (a firm that specialised in mechanical systems); the German Post Office (which had opened a video telephone link between Berlin and Leipzig using a co-axial cable and a 180 -line system) showed iconoscope-type cameras as well as an intermediate film system similar to that used by Baird at Alexandra Palace. With this technique the film was processed in just about a minute (the main drawback curiously enough was the poor sound of the film recording).

Although the BBC television service closed down on September 1, 1939, the sound transmitter later being used to "bend" German navigational beams, the Germans kept their Berlin service going until November 1943 when the transmitter was destroyed by Allied bombing; they also ran a service for Wehrmacht occupation troops from the Eiffel Tower that actually stayed on the air until ten days before the Liberation of Paris in August 1944. Pictures from this service were monitored on the Kent coast and it has been claimed that useful information was obtained from this Intelligence operation.

## Our three-colour eyes

The truly remarkable colour-fidelity that can be achieved with modern television based on the transmission of three colour signals (red, green and blue) becomes more understandable now that it is firmly accepted that human colour vision similarly depends on fust three classes or retinal receptors and that these, also like colour TV, have overlapping spectral sensitivities.
Apparently these human colour sensitivities are centred on violet, green and yellow, though spectrally these are not far away from TV's red, green and blue.
If one could only find a way of stimulating the brain directly from video signals, it might be possible to transmit a colour picture without the need for a picture tube or having to keep our eyes open! But of course if we could do that, and similarly with sound, we would be well on our way to enabling the deaf to hear and the blind to see.

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[^2]

## STOCKING UP

THis month we suggest a range of capacitors for the new constructor to purchase as part of an initial stock of components.
Capacitors are not used to the same extent as resistors, but they do come in a variety of forms, shapes and sizes, to say nothing of the capacitance values.
Therefore the holding stock will be numerically much smaller than in the case of resistors. On the other hand the choice of actual types and values is more difficult. In the tables below we have set out our recommendations concerning type and value. This selection of capacitors should meet most general requirements.

## NON-POLARISED CAPACITORS

Capacitors are classified according to the material used to form the dielectric (the insulating surface between the two "plates" of the capacitor).
Table 1 lists the more frequently used values and the types of capacitor in which they are available.

Recommended minimum quantities are in italics. If resources permit, it would be wise to multiply all these quantities by a factor of two or even five. ve.

POLYESTER CAPACITORS: These are widely used and cover the middle range of capacitances: typically from $0.01 \mu \mathrm{~F}$ to $2.0 \mu \mathrm{~F}$. Tubular, tablet and block forms are common. Tolerance 10 and 20 per cent.

CERAMIC: These include the smallest capacitors and range from 1.8 pF to $(0 \cdot 1 \mu \mathrm{~F})$ typically. Close tolerance, low loss, and high stability are their important characteristics. They are small, thin disc- and plate-shape components.

POLYSTYRENE: These capacitors are available in the range 10 pF to $0 \cdot 1 \mu \mathrm{~F}$. Thus they cover very nearly the same range as ceramics. They have a better electrical performance than the ceramics, and are available with close tolerances such as $5,2^{1} 2$ and 1 per cent. However polystyrene capacitors are rather more expensive and are
only used where their special characteristics are essential.

Polystyrene capacitors are tubular in shape and rather more bulky than ceramics.

## ELECTROLYTIC CAPACITORS

Electrolytic or polarised capacitors are a special case. They provide quite a large value of capacitance, but must be connected right way round in circuit. They are made with specific working voltages, which must not be exceeded in practice. It is however permissible to use them at a lower voltage than the stated working voltage (WV).

Generally, always follow the recommendations given in Component Lists for projects and use the specified type of electrolytic.

Table 2 gives a selection of electrolytics that should cover many needs. A working voltage of 25 V is adequate for most battery operated projects. The higher rating of 63 V has been given for the three smallest capacitors, since this is the lowest WV in which they are available.

Electrolytics are available as "double-ended" types with the connecting leads emerging from either end; also as "single-ended" types with both leads at same end of component. Double-ended types are recommended for the present purpose, see illustraticn in Fig. 1.

Some typical capacitors drawn to scale and showing their characteristic physical forms. The electrolytic is a double-ended type with axial leads.


TABLE 2
ELECTROLYTIC CAPACITORS
3
2
3
5
2
2
1

| Value | $w y$ | Oty | Value | wy | Cly |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $1 \mu \mathrm{~F}$ | 63 | 1 | $47 \mu \mathrm{~F}$ | 25 | 1 |
| $2-2$ | 63 | 1 | 100 | 25 | 2 |
| 4.7 | 63 | 1 | 220 | 25 | 1 |
| 10 | 25 | 2 | 170 | 25 | 1 |
| 22 | 25 | 1 | 1,000 | 25 | 1 |

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[^3]
## The Extra Ordinar Experin Experi~ <br> \section*{}

 oİ Proíessor Eversir Evinest
## by Anthony John Bassett

$\mathrm{A}^{\mathrm{s}}$$A^{s}$ bOB arrived once again at the Laboratory to be greeted by the Prof., his mind was full of thoughts of the Simulated Space Environment which the Prof. had produced with the aid of a giant vacuum chamber equipped with a wide range of electronic equipment; computers and lasers, electron-guns and other beamproducing equipment for experiments on energy beams, and many other applications of vacuum electronics.
The Prof's experimental robots could enter the vacuum chamber through the air-lock and Bob could see through the large windows that they and the Prof. had been at work installing more equipment.

## FLYING SAUCERS

"Your Space Environment Simulator looks more like real space exploration equipment every time I see it, Prof! I dream that we might even take a real space-trip in it!"
"Then you'll be interested in my latest experimental device, Bob."
"Wow, Prof. are you going to turn the experimental gravity control down past zero, and actually take-off and fly in the vacuum-chamber?"
"No, Bob, last time I tried that it caused yet another flying saucer scare around here, and also a UFO defence alert. I had quite a difficult job dodging the modern radar trackers and the latest batch of highly sensitive UFO detectors which have been built by investigators of such phenomena.

## RADAR JAMMER

I had to rig-up an emergency radar jammer to produce pictures of hundreds of flying saucers all over the radar screens for a short time whilst I gave them the slip and got back here safely! It must have caused quite a flap!"
"How does a radar jammer work, Prof?"
"When a craft is being scanned by radar, receiving instruments on board can detect the radar beam. To jam the radar, the craft may then transmit a series of pulses of energy at the frequency of the radar beam. This results in a number of false images on the radar screen to confuse the radar observer.
"Just as with TV games, electronic circuits operating in synchronism with the scan frequencies of the TV set can produce a great variety of images on the TV screen. So, by detecting the scan frequencies of the radars and transmitting a variety of synchronous pulses, a whole range of interesting 'objects' can be made to appear on the radar screen to divert the operator whilst the craft escapes!
"The radar jamming beam or broadcasting can produce an almost endless series of replicas of the image of the craft to appear all over the radar screen. It is amazing what can be carried on a beam of energy and your interest in the study of various energy beams and their widely differing properties can lead you to some fantastic discoveries!

## ABOLITION OF SPACE

"Some investigators consider that force fields can cause matter to become organised into objects whose structure and function are direct results of the 'originating' force fields, which are in some cases considered to precede and 'mould' the objects, which may include living creatures!",
"But Prof., if that is so, won't it result in the abolition of space itself? Because magnetic, electrical and gravitation fields can extend for infinite distances in space, if they were really responsible for the formation of objects and living creatures, surely space would soon become so filled-up there would be no gaps left in it!"

## COMPLICATIONS

"If that were so, Bob, we could all retire relatively smartly at a speed of $c-\Delta c$ into our brand new individually pre-packed black holes, but fortunately things are more complicated than that. In most known instances, for example, force fields merely act in readily understood ways to guide the formation of objects, and the growth of crystals, for instance, is an example of this which is taught in many science classes. Here an object is formed under the influence of electrostatic field forces with no danger from black holes or other apocalyptic phenomena-the force fields merely acting to guide the atoms and molecules into places in the crystal lattice structures.

## REPLICATION BEAM

"With other objects less regular than crystals the picture is more complex and my latest experiment uses energy beams to influence the forcefields in space to induce the formation of replicas of objects, in regions where interfering energy-beams have produced congruent antinodal and nodal structures in the fields which exist in space."
"Prof., the ways in which forcefields and energy beams can influence one another, and in turn influence material objects are fascinating areas for advanced study-but how can antinodes and nodes be congruent with one another?"

## SCIENCE FAIR PROJECT

"An apparatus for demonstration of the existence of congruent nodes and antinodes would make a good practical construction project for your School Science Society, Bob, and could make a good, entertaining and educational School Fair exhibit. Think about it; such a projeot could demonstrate simultaneously a number of scientific principles and there is plenty of scope in it for the use of electronics!
"It would be quite easy, for instance, to arrange that the nodes and antinodes of a microwave low power field, for example, should be congruent with those of, say, an acoustic field of the same, or different wave length, and with the different forms of energy travelling in transverse directions or at various angles. Then, of course, there are quite a variety of mechanical shear or transverse waves, compression waves, surfacedisplacement waves and so on, to experiment with, which could make spectacular mystifying and educational displays where luminescent, reflective or refractive media can be used."
"I understand now, Prof! The nodes and antinodes of different fields of standing-waves or interfering waves,
whose energy may be travelling in differing directions or in the same direction, may quite readily be congruent with one another and produce interesting and sometimes complicated combined effects!'
"That's right, Bob. Now watch this!"

## TELE-PORT

The Prof. placed a small metal object upon a low pedestal which was surrounded by strange electronic sensors connected to an energy beam projector apparatus.

When he operated the apparatus, Bob saw that, just for a moment, the metal statue seemed to be surrounded by a phosphorescent halo of light, and also to glow with a light which seemed to come from within the very core of the solid metal.
But the statue seemed to come to no harm.
"This is because," explained the Prof., "what I am doing is merely making a record of the structure of the object-it does no more harm than taking an ordinary photograph. Now, the beam-projectors are set to focus and converge on that table. Watch, and also observe carefully what happens to the small heap of grey crystals on the pan of the nearby laboratory weighing balance."

The balance tilted sharply as the crystals turned from the grey of lead sulphide to the yellow of sulphur, and thus rapidly lost weight. Bob saw that one of the Prof's experimental robots had to remove quite a lot of weights from the opposite pan before the scales would balance again. At the same time a shimmering replica of the small lead statue formed on the table. The Prof, picked it up and handed it to Bob.
"It is identical to the original," Bob explained, "down to the last tiny scratch-and I'll bet it weighs exactly the same."
"Yes-the Robot has just confirmed that." The Prof. switched the machine into reverse, and with a shimmering
sparkle and a sudden loss of weight, the statue faded from Bob's hand, and at the same time the crystals on the scale returned rapidly to their customary weight and grey colour.
"Will this work with living creatures too, Prof?"
"Yes, Bob, try it!" Bob himself stood on the pedestal and as the Prof. once again operated the device, this time with the assistance of a couple of his robots, Bob seemed to find his whole body and consciousness filled for just a moment with a flash of shimmering light. Then a perfect replica appeared on the table a few metres away, together with another replica of the Prof's lead statue which was still on the pedestal of the Replication Machine.

As Bob picked up the statue and examined it once again, his replica on the table picked up the replica of the statue and examined that.

After a short time the Prof. switched the machine into reverse and once again the replicas disappeared. Bob stepped down from the pedestal.
"That was amazing!' he reported to the Prof.
"As I was examining one side of the statue, the replica was examining the other side, and so I could see both sides of the statue simultaneously. Also I could see you in close-up with my own eyes, and from a distance through those of the replica. I could see the whole lab from two standpoints, like being in two places at once! Really great! Want to try it again-but tell me, can this machine produce more than one-several replicas simultaneously."
"That is possible Bob, but the mental and psychological problems would be multiplied, and could easily reach dangerous proportions even with one replica of an unbalanced person. So my next Extraordinary Experiment will be aimed not at increasing the number of the replicas, but the distance to which they can be projected!"

To be continued

## The Adventures of Tanty Bead




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## ШОRKSHOP

 MATTERSBy Harry T. Kitchen

## Signal Sources

In an earlier article I covered the choice of a multimeter, and in the next article or so 1 want to cover signal sources. These can be very broadly defined as a signal that is injected into equipment in order to ascertain whether it is working within its parameters, or if it is not working, to ascertain where it has failed. Such sources cover a vast range: from the "tickle" induced into sensitive stages by an educated finger, to the more sophisticated sources. It is these that I want to discuss.

Let us start by drawing a division between the source that is very closely controlled for frequency and voltage accuracy, and that which is less closely controlled if at all. The former is defined as a generator, the latter as an oscillator.

## A.F. Sources

If you are an a.f. or hi fi type then you will almost certainly be interested in such sources as a means of checking your equipment, or if you are into design work, of checking your designs. What should such a source offer? Opinions differ, but my order of priorities, in descending order, would be:

1 Very low distortion.
2 High frequency accuracy
3 High voltage output accuracy
4 High resetting accuracy of frequency and voltage
5 Wide frequency range
6 Square wave output
Having detailed my order of priorities, let me now justify theml Most modern audio equipment has distortion figures that are extremely low, and if we are to measure these, or design amplifiers that are similarly rated, then our signal source must have an output that has distortion flgures that are very much lower: a factor of $\times 10$ is not unrealistic for work of the highest calibre.
Faced with equipment with a distortion figure of 0.01 per cent we must use a source that has, at the most, 0.001 per cent distortion. It is possible to reduce this distortion figure, but then our results will be less reliable.

## Frequency Response

For the plotting of frequency response curves, a high degree of frequency scale calibration is necessary, and this also helps when "spot" frequencies must be set. Greater accuracy is possible if the frequency varying element-almost always resistance or capacitance with modern designs-has some form of slow motion tuning incorporated. This has the additional advantage of permitting frequency sweeping to be effected.

Another important factor is freedom from frequency voltage bounce, an effect where the output voltage literally bounces up and down when the frequency is altered. It is very difficult to make any source completely free from bounce, but it should be closely controlled, and should settle down very rapidly.
The output voltage should be closely controlled so that it is possible to accurately set any required voltage. This can be effected by calibrated attenuators, sometimes complimented by an output meter that monitors the voltage applied to the attenuators. There should be at least two attenuators, one for the coarse adjustment of output voltage, the other for fine adjustment.
Both frequency and output voltage should be capable of being reset to any previously set figure, and should be adequately buffered so that neither is affected by any external loads. This is not always the case with simple designs where the output is extracted directly from the oscillator, and it is quite possible to alter both frequency and output voltage by an appreciable amount by varying the load applied to the output terminals. This can invalidate any results obtained; as such, simple designs must be used only for non-critical work.
Although it is accepted, by tradition, that the audible frequency range is from around 20 Hz to 20 kHz , the source must possess a much wider frequency range, ideally 10 Hz to 100 kHz . Such a wide frequency range is essential to check reproducing equipment for vagaries that, though they may well be outside the audible bandwidth, may yet affect the reproduction.
Subsonic instability may be inaudible, yet it can modulate the reproduced signal and cause audible distortion. Similarly, supersonic instability may be well above 20 kHz , yet it can degrade the signal and in extreme cases it is not unknown for multi-unit 'speakers to have the tweeters ruined through overheating of the voice coil due to an amplifier oscillating vigorously and applying an almost perfect square wave to the 'speaker. With a wide frequency source, the amplifier can be checked for misbehaviour at both ends of the frequency spectrum.

## Square Wave

If you own an oscilloscope you will find a square wave output invaluable. This is, in comparison to the sinewave output we have so far considered, a signal having perfectly vertical sides, and perfectly parallel tops and bottoms.
The square wave is usually derived from the sinewave oscillator and thus has the same frequency. The two outputs sometimes share the same outputterminal and attenuators, though separate attenuuators and output terminals increase
the versatility of the instrument since both waveforms can be used simultaneously if desired.

The great attraction of a square wave is the ease, and speed, with which it can tell the practised user the characteristics of the equipment through which it has passed. If the verticals slope and the originally sharp corners are rounded off, it indicates a poor high frequency capability.

Sloping tops and bottoms indicate a poor low frequency capability. The greater the sloping and rounding-off, the worse the h.f. or l.f. response.

## Supplies

Last, but not least, the power supply used for the source, and the case in which it is placed. Batteries are expensive, and if the equipment is much used, have a limited life. A mains supply thus seems to be attractive.

For certain work the mains power supply is no disadvantage, but at low levels the mains frequency of 50 Hz , and some of its harmonics can cause problems. For critical low level work I believe that there is no alternative to battery power, expensive though it may be.

For screening, a metal case is essential, but this should be well insulated on the outside, by the liberal use of paint or Vinyl covering, so that if two cases touch there is no unintentional earth path between them that may cause problems with earth-loops.


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> THE CASIO fx-7100 SCIENTIFIC CALCULATOR

Electronic pocket calculators have been with us for almost ten years now and electronic digital watches and clocks for a little less. It was only a matter of time before the two were combined. This time has come, and it is no surprise to find that it is from a Japanese company. Casio, the world's leading calculator manufacturer have just released some calculating timepieces, one of which is the $f x-7100$. Strangely though, this is only called a "scientific calculator".
The $f x-7100$ is truly a pocket (or handbag) sized instrument, measuring only $98 \times 65 \times 10 \mathrm{~mm}$ in its leatherlook wallet having a window over the display. It is similar in size to a season ticket/bankers card wallet and can thus be carried with comfort in the breast pocket for example.
The display is the more economical and popular liquid crystal type ( 5 mm high digits). Current consumption is therefore very low giving long life to the two silver oxide batteries fitted. Depending on the battery types obtained, continuous use over 9 to 12 months can be expected.
There are thirty-seven actual keys on the unit, most of them having dual functions and some having three. At first, this multilayer key arrangement caused problems, with wrong functions being keyed in, but in a short time one becomes accustomed to the key shift modes. Also the small size of the keys and the closeness in relation to the human digit takes a little getting used to.

## CLOCK

The clock has a 12 -hour format with a PM flag to indicate the same.

The readout is in hours, minutes and seconds and is accurate to within $\pm 3$ seconds per day. The clock is also equipped with an alarm.

The decimal keyboard is used for setting the alarm time which is entered in hours and minutes. This is much more convenient than the step setting on most digital watches. The new alarm time overwrites the existing alarm time. Once activated a piercing modulated tone sounds for 20 seconds. The alarm facility can be "switched" out. Alarm ready state is shown on the displays by a "sound" flag.

## STOPWATCH

Another useful facility is the stopwatch. This counts in $1 / 100$ ths of a second up to 0.01 secs short of 10 hours. This stopwatch once started can be stopped in either of two modes and these modes may be alternated throughout the timing sequence.

In one mode, the "watch" can be stopped and when restarted will continue to count from the displayed time. In the other mode, the counting sequence carries on counting internally displaying the halted time. When operated a second time, the display leaps to the counting time. Very useful for sports events where lap times can be recorded. The stopwatch would also be very useful in the electronic workshop where time delay circuitary is being tested and time constants observed.

## COUNTDOWN

The housewife is also catered for. In the kitchen, the device may be
used as a cooking timer when put in its countdown mode. Here the cooking time may be entered by means of the decimal keyboard, and when initiated will count down to sero at which time the alarm sounds, again for 20 seconds.

Countdown is from a maximum 9 hours 59 minutes or may be in minutes only from 99 minutes maximum. This is a repeat timer i.e. when it reaches zero, it reverts to the set countdown time. This facility should prove very useful once again for sport events such as when competitors set off at pre-arranged intervals, e.g. car rallying.

During both countdown and countup (stopwatch), the display may be returned to regular time display without affecting the counting.

## CALCULATOR

Calculations may be carried out without affecting any clock functions that may be running. A break from calculation back to clock function is possible, but any information you have accumulated and wish to retain must be transferred to memory. Other registers will be cancelled.

This tiny machine has enormous ability handling no less than 39 scientific functions as well as the usual basic calculations including constants for $+1-|x| \div / x^{y} / x^{1 y}$.

Positive and negative numbers may be entered up to 8 -digits in length (with or without decimal point) or in exponent form, 6 -digits with 2 -digit exponent up to $10^{\text {w3 }}$. The accuracy of the scientific functions in $\pm 1$ in the 8th digit.

We were pleased to see the inclusion of "brackets" on this machine which allows quite complex formulae to be "worked out" without recourse to the memory and inverse keys as with many other calculators. A fivedeep nest of brackets is available.

A statistical mode package is included which calculates all the common statistical functions.

As would be expected, a memory is included. This can be added to, subtracted from, recalled and overwritten. Memory contents may also be exchanged with display register. Whenever there is a number in memory this calls yet another flag, M , to appear on the display.

A useful feature included in the design is to exchange the contents of the display register with the working register. This enables the last but one entered number to be recalled.

These and the rest of the functions, including the versatile percentage key, are contained in a comprehensive 70 -page manual that comes with the calculator.

The recommended retail price of the $f x-7100$ is $£ 27 \cdot 95$, but can be obtained from Tempus (Dept. EE), Talk of the Town, 19-21 Fitzroy Street, Cambridge CBI 1EH, at a discount price of $£ 24.95$ including postage and packing.

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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
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