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The young, and others whose pockets are not especially capacious, find the use of alternative materials a necessity-and we know what necessity is the mother of, so the point hardly needs labouring. But it is no bad thing for everyone, no matter the size of their pocket, to cultivate the knack of converting for their special purpose some humble run of the mill article or to make use of discarded materials. As an example we can think of the boundless supply of plastic receptacles in infinite variety that have but a fleeting life once taken from the supermarket shelves. As our enterprising readers have demonstrated on innumeral occasions there are possibilities for extending the life of these plastic containers in our own special field.

Make and do for oneself is the underlying purpose of our Mini Module Series. The tobacco tin motif is not a leaning back to the good old days when do-it-yourself actually meant just that. But it symbolises the homespun as against modern orthodox construction. These useful projects have been used as vehicles to demonstrate simple but effective methods in assembling simple circuits. Cardboard, plywood, pins and nails, have all shared the honours, along with more orthodox items such as tag strips, terminal blocks and stripboard; and the tobacco tin after its inaugural appearance has been superceded by electricians' switch boxes and purpose built plastic boxes.

As readers will already have noticed another contributor has just joined our pages and Rummaging Around is intended to reinforce the make and do aspect, albeit in a lighthearted way. And why not, for it is fun.
In any activity, a certain degree of self-sufficiency must always be admired, and in electronics construction it provides a beneficial antidote to the all-too-easy-just-assemble-a-kit attitude which maybe we are in danger of slipping into.


Our July issue will be published on Friday, June 15. See page 343 for details.


## Readers' Enquiries

We cannot undertake to answer readers' letters requesting modifications, designs or information on commercial equipment or subjects not published by us. All letters requiring a personal reply should be accompanied by a stamped self-addressed envelope.
Telephone enquiries should be limited to those requiring only a brief reply. We cannot undertake to engage in discussions on the telephone, technical or otherwise.

## Component Supplies

Readers should note that we do not supply electronic components for building the projects featured in EVERYDAY ELECTRONICS, but these requirements can be met by our advertisers.

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THE TREMOLO is believed to be one of the first musical effects units-certainly for use with the electronic guitar-and was featured prominently in "pop" music of the fifties. It is used to a lesser extent in present day music but is still required for certain musical numbers.

In the past many methods have been used to obtain the tremolo


Tremolo is the name given to the amplitude modulation of the guitar signal by a very low frequency, typically 3 to 10 Hz . The result is to produce "packets" of musical signal, see Fig. 1. Varying the speed and depth of the modulation produces different effects.

The unit to be described here has two set depths-deep and shallow, but can easily be made


Fig. 1. The tremolo effect in graphical form. A sinewave "music signal" has been shown for clarity.

9 V battery. Current consumption is very low typically $1 \cdot 7$ to $2 \cdot 1$ milliamps, so the battery should have a long life.

The circuit is seen to be in two distinct sections: a current controlled amplifier and an oscillator.

## CURRENT AMPLIFIER

The amplifier uses a transconductance operational amplifier IC1. In many respects this is similar to "normal" op-amps and has the same pinning as the common 741. The gain of the amplifier however, is controlled by the current flowing into pin 5 , producing an output current into a fixed load, R7 in this circuit.

Resistors R1 and R2 effectively produce a split supply for IC1 with C2 performing decoupling.

The input impedance of the circuit is approximately 8 kilohms which should suit most electric guitars.

Preset VR1 is used to control the overall gain by setting the quiescent current flowing into pin 5 via R6.

## OSCILLATOR

The oscillator is composed of IC2 and associated components connected as a phase-shift oscillator. The output is sinusoidal at a frequency set by VR3 (the speed control) which varies the frequency from about 3 to 10 Hz . The peak-to-peak amplitude of the oscillator output is 5 volts about mid-supply, which is fed via Ro and C4 to R6 and then into pin 5. Thus the gain ICl is controlled by the oscillator output. The result is to cause the output signal from IC1 to be amplitude modulated as depicted in Fig. 1.
Resistors R13 and R14 effectively produce a split supply for IC2 with decoupling by C8.

The depth of modulation is controlled by R8 and S1, the DEPTH switch. For a deep effect S1 is set to short out R8. This resistor is in circuit for a shallow effect.

The oscillator is caused to cease oscillating by closing S 2 , the effect control. The musical signal appears at SK2 unaffected with an amplitude controlled only by VR1.

If a continuously variable depth control is required, R8 should be



Fig. 2. The complete circuit diagram of the Tremolo Unit.
replaced with a 220 kilohm potentiometer wired as a variable resistor, S 1 would then of course not be needed.

## COMSTRUCTION staris hare

The construction of the unit differs a little from that shown on the front cover as the design has been modified slightly, to improve its performance. The le.d. has been removed in the interests of battery economy and a larger case has been employed to make construction more easy.

## CIRCUIT BOARD

The circuit is built on a piece of 0.1 inch matrix stripboard, size 31 strips x 29 holes. The layout of the componets on the topside of the board and the breaks to be made along the strips on the underside are shown in Fig. 3.

No fixing of the board was found necessary on the prototype. It was laid on a piece of polystyrene tile and the case mounted switch and sockets held the board steady. If board fixings are required holes should be made in the case and board before proceeding; spacers to hold the board clear of the case are essential as well as fixing isolating breaks on the copper strips.


Begin by mounting the resistors, capacitors, link wires and preset followed by the i.c.s.; ICl is a T0-5 type and its legs have been formed to suit a di.i. arrangement. Pin 8 is adjacent to the tag on the body. These devices can also be obtained with pre-formed d.i.l. arrangement and IC2 is available also in a plastic d.i.l package as is more commonly found.
Attach sufficient lengths of flying leads to reach the case mounted components. Note that some

## 

Resistors

| R1 | $10 \mathrm{k} \Omega$ |
| :--- | :--- |
| R2 | $10 \mathrm{kk} \Omega$ |
| R3 | $1 \mathrm{k} \Omega$ |
| R4 | $10 \mathrm{k} \Omega$ |
| R5 | $10 \mathrm{k} \Omega$ |
| R6 | $4.7 \mathrm{k} \Omega$ |
| R7 | $10 \mathrm{k} \Omega$ |


| R8 | $150 \mathrm{k} \Omega$ |
| :--- | :--- |
| R9 | $1 \mathrm{M} \Omega$ |
| R10 | $100 \mathrm{k} \Omega$ |
| R11 | $10 \mathrm{k} \Omega$ |
| R12 | $3.3 \mathrm{k} \Omega$ |
| R13 | $10 \mathrm{k} \Omega$ |
| R14 | $10 \mathrm{k} \Omega$ |

page 333
All $\frac{1}{4}$ watt (or less) carbon $\pm 10 \%$

## Capacitors

| C 1 | $0.22 \mu \mathrm{~F}$ polyester | C |
| :--- | :--- | :--- |
| C 2 | $0.47 \mu \mathrm{~F}$ polyester |  |
| C3 | $0.22 \mu \mathrm{~F}$ elect. | C 6 |
| $0.47 \mu \mathrm{~F}$ polyester | C 7 | $0.47 \mu \mathrm{~F}$ polyester |
| C4 | $0.22 \mu \mathrm{~F}$ polyester | C 8 |
|  | $33 \mu \mathrm{~F} 6 \mathrm{~V}$ elect. |  |

## Integrated Circuits

IC1 CA3080 transconductance op-amp. 8 pin d.i.I. (preferable)
IC2 741 differential op-amp 8 pin d.i.I.

## Miscellaneous

VR1 $47 \mathrm{k} \Omega$ horizontal carbon preset
VR2 $100 \mathrm{k} \Omega$ carbon Lin. potentiometer
S1, 2 single-pole successional action footswitch (2 off)
SK1 stereo jack socket
SK2 mono jack socket
B1 PP3 9 volt battery
Stripboard: 0.1 inch matrix 31 strips $\times 29$ holes; PP3 battery clip; knob for VR2; aluminium diecast box approximate size $115 \times 90 \times 50 \mathrm{~mm}$; screened cable; connecting wire.




The completed component board removed from its case.
of these are screened cable to reduce extraneous hum and noise.

Prepare the case to accept the sockets, switches and potentiometer. The switches can be fitted to the lid. Secure the board in place and then the sockets and potentiometer and wire up according to Fig. 3 .
The prototype used stereo sockets but a stereo type is only required for SK1. The rear earth tag is used as the battery on/off switch. This is made along the shank of the input mono jack plug when inserted. The unit is automatically switched off when the input jack is removed from the unit.
Before connecting the battery set VR1 fully anti-clockwise.

## SETTING UP AND USE

Plug the instrument in at SK1 and connect via a suitable lead from SK2 to the amplifier input. A high impedance amplifier input is required for best results.

With S2 set for effect, S1 for maximum DEPTH, and VR2 fully clockwise (maximum SPEED), strum the guitar and finely adjust VR1 with a small screwdriver until the


The completed prototype unit with lid removed showing case mounted components, board positioning and wiring to the lid mounted switches.
tremelo sound is obtained. Turn VR2 anticlockwise and the modulation rate should decrease.
Operate S1 to obtain a less deep effect. The value of R8 may need altering to suit but will probably be in the range 100 to 220 kilohms.
Operate S1 to inhibit the effect and allow straight-through-playing.

If all the above tests are successful, the unit is ready for use.
The prototype was sprayed matt black (after application of a primer) and labelled using Letraset to produce a professional looking unit. Four rubber feet glued to the case underside prevents the unit sliding under operation.


By Dave Barrington

## Mail Order Only

IT IS sad that Home Radio Components Ltd, one of the oldest names in the electronic components business (at least on the retail side) had to move due to rent increase. To soften the blow, I am informed by the managing director that despite their enforced move they were in good heart and delighted with results of
their moving sale. He reported that at the moment they cannot forecast the restoration of a counter service and at the moment only trade by mail order.
He does advise that if you require something in a hurry to telephone first, the number is 01-648 8422.

## Constructional Projects

Some readers may experience difffculty in obtaining the tuning capacitor C2 called for in the Pocket Radio. The final value of this component is not critical and provided the value is near to 126 pF any type may be used.
The one used in our prototype was a miniature 4 -gang a.m.ff.m. type for printed circuit board mounting. These are quite common and ours was ordered as a "Min Tuner", Cat. No. FF52G, from Maplin.
The length of the ferrite aerial is non-standard and will have to be cut to size. This material is very brittle and extreme care must be taken during cutting. We suggest you "score" around the rod before final cutting as sometimes it tends to snap unexpectedly.
Our Low Cost Metal Locator this month is a very simple design, but the results have been very good, and should not present any problems re-
garding purchase of components. The p.v.c. insulated wire for the search coil is available generally and the final housing is left to individual choice.
In respect of the housing, the original search coil cover called for in our last article and available from Arrow Electronics, Dept. EE, Leader House, Coptfold Road, Brentwood, Essex, or a moulded plastic search coil head case available from Ambit International, Dept. EE, 2 Gresham Road, Brentwood, Essex, could be used.
The only component likely to cause any concern in the Electronic Canary is the transformer. Practically any type can be used in this project, but component values may have to be changed to obtain the desired sound effects if other than the specified type is used.
The LT700 transformer is available from Electrovalue at a cost of 44p, including VAT, plus 27p handling charge if ordered as a separate item.
The only point to be made regarding the Quad Simulator is that wirewound components rated as specified or greater must be used.
The Tremolo Unit and this month's Mini Module project use standard easy to obtain components and should not give any buying problems.


THE TWITTERING or chirping sound from this Electronic Canary unit is very distinctive, and will find use in replacing the traditional door bell or buzzer. Children will also find the sound novel, and will doubtless play with it purely for fun. Operation is economical, the current drain from a PP6 battery being very low and will give many hours of use.

## CIRCUIT DESCRIPTION

The circuit diagram for the Electronic Canary is shown in Fig. 1.

The oscillator is the type known as "self-quenching" or "squegging", and is based on the effects of phase reversal in a transformer.

The primary of the centretapped speaker transformer TI is connected between collector and base of TR1 and drives the base in the correct phase to produce audio oscillation. In a short interval the charging of Cl cuts off the oscillator, which however recommences when the charge leaks away. This effect continues so long as the push is held closed. When the push is released oscillation is maintained for a short while by the charge in C5, but falls away and ceases.

The capacitors C2, C3 and C4 are selected to produce a clear chirping or whistling sound, and the preset potentiometer VR1 allows some adjustment to this.

## COIPPOIFENTS apuroximate HIST £5

Fig. 1. Complete circuit diagram of the Electronic Canary.

Experiment shows that none of the values need be exactly as shown, but that all influence the result obtained. Changes to values are not recommended except as a basis for tests and experimenting for other sounds.

The author's prototype in fact used a different type to that specified in the components list, being a type T/T7. As far as we know, this is not generally available and the circuit has been slightly modified to use an Eagle LT700 type.

If any constructors are able to obtain or already have this type three component values need to be changed back to their original values to use the T/T7. These are R1 to 470 ohms, R2 to 10 kilohms and Cl to 100 microfarads.

If other types of audio output transformers are to hand, these could be used but will almost certainly need resistor and capacitor value changes to obtain the same result-or other sounds and effects


## CIRCUIT BOARD

The circuit is constructed on a piece of 0.15 inch plain matrix board having $19 \times 12$ holes and is shown in Fig. 2. The transformer has two tabs which project through slots in the board-these can be made by drilling a few holes closely side by side. The tabs are bent or twisted to hold the transformer.

The variable resistor, VR1, is fitted in a similar manner. Its wiper and one outer tag are connected together. If a fixed resistor is substituted for VR1 and R2, connect this from collector to base of


Fig. 2. Matrix board layout. The style of the preset may differ from that shown, but the connections will remain the same.

## COMPONENTS

## Resistors

R1 $120 \Omega$
R2 $22 \mathrm{k} \Omega$
$\frac{1}{4} \mathrm{~W}$ carbon $\pm 10 \%$

## Potentiometer

VR1 $5 \mathrm{k} \Omega$ miniature preset

## Capacitors

C1 $220 \mu \mathrm{~F} 10 \mathrm{~V}$ elect.
C2 $0.047 \mu \mathrm{~F}$ polyester
C3 $0.1 \mu \mathrm{~F}$ polyester
C4 $0 \cdot 22 \mu \mathrm{~F}$ polyester
C5 $1000 \mu$ F 10 V elect.

## Semiconductor

TR1 AC142 germanium pnp

## Miscellaneous

T1 Eagle LT700 audio output transformer
LS1 $8 \Omega 60 \mathrm{~mm}$ diameter or similar speaker
S1 single-pole push switch (door bell type)
B1 PP6 9 V battery
Matrix board, 0.15 inch $19 \times$ 12 holes; plastic or metal case -or-material for a home made case; aluminium for mounting brackets; connecting wire; battery clip to suit B1.


Fig. 3. Final interwiring details, showing the relative positions of the components. If a ready made case is used then a slight change in positioning may be necessary.

## Electronic Cดกดry



The completed Electronic Canary showing positioning of components. The preset potentiometer UR1 can be any common "skeleton" type.


Layout of components on the completed circuit board.

TR1. If the value is rather high, say over 27 kilohms, chirping is less strong. But if the resistor is too low in value, say under about 15 kilohms, only a continuous whistle is obtained. It is for this reason that VR1 is included.

Note the polarity for Cl and C 5 . Solder on the battery connector.
Two Veropins are inserted for the bell push connections.

Twin flexible bell wire runs from these two pins, through a small hole in the bottom of the case, and to the bell push.

## CASE

A ready-made plastic or metal box can be used, or a case may be easily constructed from 6 mm plywood as used on the prototype. The front panel is $145 \times 115 \mathrm{~mm}$, the top and bottom edges 100 x 65 mm and $100 \times 38 \mathrm{~mm}$ respectively. The sides are sloped from 65 mm at the top to 38 mm at the bottom, and are 145 mm long. An aperture to match the speaker cone is first cut in the front.

The sections are secured together with woodworking adhesive. Four 10 mm square or similar strips in the corners will give additional strength.

## WIRING

The remainder of the wiring is shown in Fig. 3, and shows the connections to the board and other components. The mounting bracket can easily be made from 18 s.w.g. aluminium and is screwed to the top so the whole unit can be hung on a wall

A small bracket for holding the battery in place is required and can be made from the same aluminium. Alternatively a self-adhesive strip or Blu-Tak can be employed.

Place the circuit board as in Fig. 3 (speaker omitted) and drill through both board and case front for countersunk 6BA bolts. These are about 25 mm long so that the board will clear the speaker when fitted with spacers or "spacing" nuts.

A piece of suitable material about $180 \times 140 \mathrm{~mm}$ is placed over the front and held taut with adhesive spread along the sides of the case. A strip of woodgrain or other self-adhesive material about 560 mm long is fitted to cover sides of the case to improve appearance.

The bird motif was cut from a greetings card and various decorations of this kind would be in keeping with the unit. A bird outline could be cut from coloured paper for use here instead.

Once completed the unit can be installed in any convenient position. The sound produced is quite loud and should be heard over quite a distance.
Once fitted it will certainly add that touch of individuality to your front door!


## EE CROSSWORD No 16 ey o. p. newtom

## ACROSS

1 Wiry and identical at the heart of it $(4,4)$.
4 Audible indicator due to electromagnetic action.
7 Wiry but like air.
8 Fine feathers for a set of clues.
9 Spacer.
10 Have a strong dislike.
12 Set of eight.
14 Puzzle variation.
17 Knee-deep in want.
19 Unlicensed radio station.
21 Base material.
24 Register a light garment of some calculators.
26 To charge up a battery in the generalised sense.
27 Necessary computing characteristic.
29 Horse-drawn carriages.
30 Patron Saint of sparks?
31 If you can't make these meet there's an open circuit.

32 Underground device for keeping the feline whiskers in trim?
33 A contract it LED for heading one of the nobility.
34 Tuneful set of calculator digits.

## DOWN

1 An educational series ( 5,2 ).
2 One is about a discord (Anag.).
3 Revered ancient object.
4 Amplifier, of a kind.
5 High-powered optical amplifier in a hurry.
6 Electrons getting out of hand, usually on a thermal basis.
8 It figures in a computer.
11 Entrusted with aerial defence from afar ( 3,3 ).
13 Repetitive noise on coils.
15 A reverse span.
16 Spacecraft from a stellar twin.
18 Transforms a core with a burdensome device.
20 Throw an electronic task.
22 Found before storms.


23 Adulterated with another metal.
25 No sleep on this for the prototype $(4,3)$.
28 Not the object of the screen.
29 Lizard.
30 Cut a tape, but not to open a fete.

Solution on page 382


By ADRIAN HOPE

## Sound Off

As far as I know there has never yet been a divorce action based solely on the grounds of over-loud hi fi. But disputes over home listening levels have certainly undermined many a state of marital bliss. The senario will be familiar to many a male; listening to a favourite piece of music he turns up the volume of the hi fi system and is berated with a "why must you play it so loud". Recently I chanced on some fascinating research work in the USA which explains why this should be.

Around 10 years ago the CBS laboratories in Stamford, Connecticut were commissioned to investigate modern methods of measuring the true loudness of sound. It's important to realise that there is a world of difference between volume level (as for instance) measured by a VU (volume unit) meter and subjective loudness.

In fact for a given VU meter reading, tones of different frequency may differ in loudness level by as much as 20 to 30 dB . Likewise although a sound level meter can provide an accurate measure of energy in a sound field, there is still no agreement on how the meter circuits should be "weighted" or filtered to bring their electronic characteristics into conformity with the sensory characteristics of the human ear and brain.

There are also more subtle factors. For instance if two recordings are made at identical energy levels, one of someone shouting and the other of the same person speaking, the shout recording will always sound subjectively louder. CBS argue that this is probably because of the feeling of urgency in the shouting voice.

Recording engineers and broadcasters have long since recognized that it is possible to make an announcement, of solo voice or instrument sound artificially loud by
doctoring its frequency characteristic. The Tamla Motown sound of a few years ago was achieved by packing as much energy as possible into every band of the frequency range; likewise radio commercials can be made artificially obtrusive by a similar technique.

Unfortunately existing sound level and VU meters may not show up such artificial loudness tricks and CBS was looking at new approaches to metering. For this research they asked teams of guinea pig listeners to compare the relative loudness of wide band signals. Some of the guinea pig teams were male and others female.

What CBS discovered to their surprise was that women hear high frequencies better than men. The maximum sensitivity in women being 4.5 kHz and in men 3 kHz .

This could well explain why women do object to loud hi fi sounds. Modern music has a very high energy content at around these levels, so it just sounds louder to women than men. And any distortion will be much more noticeable.

Quite why this should be is not entirely clear. Certainly the pitch of female speech is higher than male speech; but surely female ears should be equally well tuned to hearing both male and female sounds.

## Miniature Chatter Box

"We ain't seen nothing yet", as Jolson once said. In the USA recently a life sized robot called Argon walked into a bar and freaked out staff and drinkers alike by proclaiming "It sure is cold outside, I need a scotch".

In the event it turned out that the exercise was part hoax, because although the robot walked well enough on its own, it wasn't talking on its own. The words were coming from a man outside the bar transmitting by radio to a receiver and amplifier inside the robot's head. But talking gadgetry
is now a reality. There are two major developments in this area.

For years now audio engineers have been able to replicate human speech with reasonable intelligibility, by gating, filtering and modulating tones and hisses produced by generator circuits. But the circuits have been daunting in their complexity and Texas Instruments deserves the most hearty congratulations for now miniaturizing the essential elements into a package the size of a calculator. In fact the heart of the Texas speech synthesis package is a single MOS chip made in the same way as a calculator chip.

The details routinely released to the press by Texas in the UK so far have been very sparse, and written as if either the writer doesn't really understand how the system works or is a computer freak unable to speak plain English. Texas talk of LPC or linear predictive coding which appears to be a rather grand way of describing a filter which varies its parameters under fairly rough digital control to modify input signals so that acceptable analogue speech sounds result.
There are two forms of input signal, periodic or pitched and random noise.
The periodic input signals are used as the basis for vowels or voiced fricatives such as the sounds $\mathrm{Z}, \mathrm{B}$ or D. The random input signal is used for unvoiced sounds such as S, F, T and SH. Two roms, each with the capacity to store enough digital instructions for 100 seconds of speech, control the filter action on the periodic and random inputs.

## Translating Calculator

The inevitable next step will be to mate the voice synthesis chips with a fascinating new type of calculator which is being sold in the USA for around 200 dollars. These look like normal calculators, and perform all function calculations and metric conversions, but the calculator display is alpha-numeric; that is to say it reads off both numbers and letters. So a word punched into the keyboard comes up on the display.
Now here's the clever part. The calculator is designed to take plug-in memory modules, which contain well over a thousand words and phrases of a foreign language. With a French, Italian, German, Spanish or Japanese module plugged into the calculator a word or phrase punched into the keyboard in English will appear in the appropriate foreign language on the display.

A Russian translation module is planned in time for the 1980 Moscow Olympics and the logical, in fact inevitable, next step must be to couple speech synthesis with translation display. After all, most foreign phrases are quite worthless, unless you know how to pronounce them correctly.

## OOINE II IIGITALIII <br> 

By O. N. Bishop

problems much more difficult than this one and to produce an error-free answer in a few minutes (or less), but we shall learn more by "doing it digitally" with a few simple i.c.s on the Test Bed.

First we need to set out the problem in a way that a logic circuit can understand. We need a set of symbols for the projects that Jack might make, or might not make:

If $R$ stands for making the Roulette Game, P for the Power Supply, F for the Air Freshener and H for Hot Line, and if $\bar{R}, P, \bar{F}$, and $\bar{H}$ stand for not making the project indicated, we can put the statements of Jack and his family in simpler form:

Jack (1st): not $R$ and $P$ (cannot afford to make both)

Daughter: R or $H$ or both
Mother: If $\bar{H}$, then $F$
Son: If $F$ or $H$, then $P$
Jack (2nd): If $P$, then $\bar{F}$
The next step is to sort out what logical relationships occur in these statements. To remind you of the main ones, see Table 9.1.

Table 9.1. Composite truth table.

| Inputs |  | Outputs |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $A$ | $B$ | AND | OR | NAND | NOR |
| 0 | 0 | 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 1 | 1 | 0 |
| 1 | 1 | 1 | 1 | 0 | 0 |

In the input columns of Table 9.2, 0 means not make, that is to say $\bar{R}$ and $\overline{\bar{P}}$ respectively. In the output column, 0 means false, i.e. not consistent with Jack's statement. Conversely, 1 means true, i.e. consistent with his statement. From this we can see that the only action that is incon-

Table 9.2. The first condition in truth table form.

| Inputs | Output |  |
| :---: | :--- | :--- |
| $R$ | $P$ |  |
| 0 | 0 | 1 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |



Fig. 9.1. Using basic logic gates to illustrate De Morgan's Laws.
sistent with his statement is to make both $P$ and $R$-which is precisely what he said he would not do.

The sequence of outputs shows that this is a truth table for the NaND of $P$ and R. A NAND gate will give electronic form to this statement. In symbolic form we write the statement: $\overline{R . P}$.

The full-stop signifies the operation AND; the line above signifies NOT; combined they indicate and followed by Not, which we usually refer to as NAND, or negative AND.

Daughter's statement is obviously an or, and can be written: $R+H$. The " + " sign signifies the or operation.

Mother's statement is less obvious, but if you set out the possibilities in a truth table you will recognise that the relationship is or (remember that Mother does not say what she wants Jack to do about making $F$ if he does make $H$ ). We write mother's statement as: $F+H$.
By writing a truth table for Jack's second statement we find that this reduces to a NAND, giving us $\overline{P \cdot F}$.
The son's statement is a little more difficult to analyse, as it has three terms in it. We can try re-wording it.

If Jack makes $P$, his son will certainly be satisfied. But if Jack does not make $P$ then he must not make $F$ and not make $H$. In short, make $P$ or not $F$ and not $H$. In symbols this is written: $P+\overline{F . H}$.

To suit all the family we have to find a solution $Z$ consistent with every one of the separate statements. We AND all the statements together:
$Z=\bar{R} \cdot P \cdot(R+H) \cdot(H+F) \cdot(P+\bar{F} \cdot \vec{H}) \cdot \overline{P \cdot F}$ Three more lines of Boolean algebra would reduce this equation to its final solution but let us make the i.c.s. take over the work for us, as explained below.

## DE MORGAN'S LAW

The equation for $Z$ contains a number of ors and ands yet most of our i.c.s contain NORS and NANDS. To convert from ORS and ANDS to NORS and NANDS we make use of two simple equations:

$$
\begin{aligned}
& X+Y=\overline{\bar{X} \cdot \bar{Y}} \\
& X . Y=\overline{\bar{X}}+\overline{\bar{Y}}
\end{aligned}
$$

In words-to or two (or more) quantities, invert them, then nand them; to AND two (or more) quantities, invert them, then nOR them. There is no space here to prove these two equations, but you can check that they really do work by wiring up the circuits shown in Fig. 9.1.
Using the upper equation we can see that the daughter's and mother's statements can be re-written as $\overline{\bar{R}} \cdot \overline{\bar{H}}$ (the NAND of the inverted $R$ and $H$ ) and $\overline{\bar{H}} \cdot \bar{F}$ respectively. Then we can notice that the son's statement can be rewritten as the NAND of the inverted $P$ and the inverted mother's statement. Now we are ready to turn this into a TTL circuit.


## TTL SOLUTION

The realisation of these conditions in terms of logic gates appears in Fig. 9.2 a and built up on the Test-Bed in Fig. 9.2b. The circuit has four inputs, $R, P, F$ and $H$ and only one output, $Z$.

Fig. 9.2a. Circuit diagram for solving Jack Plug's project building problem. Below shows pinning details for two of the i.c.s used.


Fig. 9.2b. The circuit of Fig. 9.2a built up on the Test-Bed.


There are five NAND gates each corresponding to one of the statements, and each fed by two inputs either direct or inverted.
The son's NAND receives one of its inputs from the output of the mother's nand. The five nand outputs are all anded (to please everybody) by using a five-input NAND followed by an invert gate. Actually we have to use an eight-input NAND (the 7430) and let the three unused inputs float "high" (logic 1).
To find out which solutions (if any) are possible we need to apply all possible combinations of high and low inputs to the circuit. To do this systematically we can apply a fourdigit binary count, from 0000 (build nothing) to 1111 (build all). Any set of inputs which causes the l.e.d. to light indicates a combination of projects that satisfies all the requests of Jack and his family. If the l.e.d. never lights, there is no way of satisfying everybody.

The keyboard coder (Figs. 5.4 and 5.5 ) is useful for providing the binary inputs 0000 to 1001 ( 0 to 9 decimal); simply operate keys 0 to 9 in turn. To obtain inputs from 1010 to 1111, hold key 8 down and press keys 2 to 7 in turn. Try it and find out what Jack decided to build.

## PLAYING GAMES

If you want to play Chess against a logical circuit you will have to spend more than $£ 100$ on a microprocessorbased system. To learn a little about how logic circuits can be used for playing certain kinds of game, we must be content to analyse one of the more simple games.

A circuit that plays noughts-and-
crosses was described in the first DID series (July 1977). Now we present another simple game-The Pig and the Farmers. This game can be played on a board with counters but-we shall need a special board with electrical connections to the TTL circuit. A simple version is shown in Fig. 9.3.

When the Pig or a Farmer is to occupy a certain position on the board, the corresponding crocodile clip is attached to the appropriate contact. Note that there are two contacts at each position-one for the Pig and one for a Farmer. This makes the logic simpler. Pig and a Farmer may not occupy a contact at the same position at any time.

## RULES OF THE GAME

The three farmers begin at the positions marked 1 to 3, the Pig begins at any one of the positions marked 5 to 8. The Pig moves first and may move in any direction along one of the lines to the next contact along that line.

The Farmers move (one each turn) along the lines from contact to contact, but may not move backward. No jumping is allowed.

The Pig (which is played by the logic circuit) is aiming to move from his end of the board to the opposite end. If he reaches one of the contacts marked 1 to 3 he wins, scoring one point: if he reaches that at the extreme end, (No. 1), he scores two points.

The aim of the Farmers (played by you) is to corner the Pig at his end of the board, in which event you win a point. If the Pig is able to move backward and forward between the two same contacts three times, he wins.

## ANALYSING THE GAME

The logic circuit is designed so that it recognises the positions of the

Farmers and instructs the Pig what to do. There are 280 ways in which Pig and Farmers may be set out on the board, so obviously we shall not be able to arrange for three i.c.s to recognise all these positions.
We can begin to simplify the logical requirements by eliminating from our discussion those positions in which (a) Pig is trapped and has no move available and (b) Pig has only one contact available to move to. We can also eliminate those positions in which Pig has choice but among these choices is free to move in a forward direction. In general Pig should always take a forward move if this is open, so we can make this a "standing instruction to Pig". Also in the rare cases where Pig has no forward or sideways move available he can take any backward move. Thus the standing instructions to the person moving the Pig are:
(1) If you have no move availablestay put-you have probably lost.
(2) If you have only one move available-take it.
(3) If you have a forward move available-take it.
(4) If you have no forward or sideways moves available-move back.
(5) If none of the above instructions apply, refer to the logic circuit.

This leaves to the logic circuit the task of deciding what to do when the

Table 9.3. Required logic positions for Pig/Farmers game.

| Pig on | Farmers on** Pig should move |  |
| :---: | :---: | :---: |
| 8 | 2 and 6 | left (to 7) |
| 8 | 3 and 6 | right (to 5) |
| 7 | 3 and 4 | right (to 6 ) |
| 6 | 2 and 4 | left (to 7) |
| 6 | 3 and 4 | right (to 5) |
| 5 | 2 and 4 | lett (to 6) |
| 3 | 1 | right (to 2) |
| 2 | 1 | left (to 3) |

- only two farmers matter; the position of the third farmer, if it is important, is taken care of by applying the standing instructions.

Pig has a choice of sideways moves. There are only a few such positions, which is why it is possible to cover the logic required using only the three i.c.s that can be accommodated on the Test Bed. The positions in which logic is required are listed in Table 9.3.
From this table we can see that the circuit needs six inputs to describe the position of the Pig, and five inputs to describe the position of the Farmers. It needs two outputs, one to indicate "move left", the other to indicate "move right".
Certain combinations of positions occur twice in the table, see Table 9.4.

Table 9.4.

| Pig on | Farmers on |
| :---: | :---: |
| 8 | 6 |
| - | 3 and 4 |
| - | 2 and 4 |

We can make use of this fact to simplify the gate connections. These pairs of inputs can be anded and each used as a single input at later stages in the logic.

## AND-OR-NOT GATE

There is a combination of gates in one i.c. that will solve the logical problems of playing Pig and Farmers, the 7454. The pin-out details are shown in Fig. 9.5. It has eight inputs and one output. The output is nor-


Fig. 9.4a. The circuit diagram for the Pig and Farmers game.
mally high, but goes low when the following inputs go high:
$A$ and $B$ or $C$ and $D$ or $E$ and $F$ or $G$ and $H$ or any two or more pairs of inputs listed above.

In short, the inputs are anded in pairs, the and outputs are all ored, and the or output is inverted (nOT), so giving this combination of gates its rather lengthy name; Fig.9•4(a) shows


Fig. 9.4b. The circuit of Fig. 9.4a wired up on the Test-Bed.


Fig. 9.5. Pinning details for the 7408 and 7454 logic i.c.s.
how we use it in the game.
The and-or-not gate needs high inputs to activate it, and most of these come directly from the contacts on the playing board. The contacts are normally held low by the grounded resistors, R1-RII, but any given input becomes high when contacted by a Pig or Farmer occupying that position on the board. To obtain the combinations of positions listed in the previous section, we use three aND gates in a 7408 i.c. (pin out details in Fig. 9.5).

The complete layout of the logic circuit on the Test Bed is given in Fig. 9.4(b).

When both members of any pair of inputs of the 7454 is made high, the output goes low and the l.e.d. is extinguished. Thus the normal state is for both l.e.d.s to be lit, and this is to be interpreted as a red warning "do not go left or right, but keep straight on if possible". If one l.e.d. goes out, this is interpreted as advice to move to that side, and a warning not to go to the opposite side.

## COMPUTER GAMES

The Pig and Farmer circuit is designed for that game and for no other. The circuit has fixed wiring so connected as to maximise the Pig's chance of victory. The wiring is in effect part of the programming.

Note also that immediately contacts are made, the circuit indicates the recommended move (ignoring propagation delays in the gates). This scheme is different from that found in most computer games. The computer and microprocessor are extremely complex general-purpose logic circuits, not dedicated to any particular game.
The logic of the game exists not in the way gates are connected together,
but in the program stored in memory. A memory programmed to play Pig and Farmers would instruct the microprocessor to scan the contacts regularly and yery frequently to find out and record which positions were occupied. The program would then cause the inputs from occupied positions to be anded, and the resulting outputs ored, then inverted. But the difference is that in our circuit we have an AND gate an or gate and an invert gate for each logical operation.
In the microprocessor or computer the anding and other logical operations are all done by one special part of the device, the arithmetic and logic unit (ALU) which is used over and over again for each operation. Results of one operation are stored temporarily in memory, and recalled for the next operation. Thus analysis of the state of play involves a sequence of operations and at the end of perhaps a hundred steps the final output is used to control the indicator l.e.d.s.

Although microprocessors are very much in fashion these days, there are still many instances when their great flexibility and versatility are of no particular advantage, and the logical operations can be more cheaply and directly performed by a few simple logic i.c.s.

To be continued


Talking recently to the managing director of one well-established purveyor of electronic components, he reminded me of the Credit, Deposit Account Scheme which he started some twelve years ago. The scheme is so good, that I think it deserves to be more widely known. Let me just say a brief word about it.

It had always been the directors dream, to provide facilities where it was possible for any customer to pick the phone up and say to their supplier "Please send ten of your XYZ capacitors" and get them, perhaps the next day. The problem he had to solve was this: he could not open up accounts for all and sundry, without the time consuming and irritating process of taking up references, for obvious reasons.
He finally solved it in a quite ingenious way. He laid down that the cusomer made a deposit for half the amount of credit he required, that is to
say, if he wanted $£ 10$ credit, he deposited $£ 5$, and this obviated the need for any references, he just signed a simple form and he, (the customer) was in business. To make the scheme more attractive he supplied all account customers with prepaid envelopes.
He even went to the lengths of promoting an advertising campaign, called "Shopper A" and "Shopper B". Shopper " $A$ " trudging from shop to shop. and likely to die of exposure and shopper "B", by the fire, with his feet up, buttered toast in one hand and the telephone in the other, completing the same exercise in comfort. He tells me that he has over 1200 customers who use this method of purchase and he is surprised that the idea has not been adopted by other component retailers. With a scheme like that, who needs a counter service.

When I was a humble service engineer, my wages were such that I never expected to own anything more lux-
urious than a second hand push bike. Today, if the charges are anything to go by, I would expect a service engineer to call on me in a chaffeur driven Rolls. This was forcibly brought to my attention the other day when one of our electronic calculators went wrong.
As it will do almost any calculation and give you a print out, it is invaluable and I was anxious to get it back into service as soon as possible. I telephoned to the makers, who asked "Have you a maintenance contract?" When I answered in the negative, they informed me that their terms are $£ 21$ for the first half hour and $£ 12$ for each subsequent half hour.

It seemed to me that the minimum bill would be £33. If these charges cannot be justified, and I am sure they cannot, one wonders if it is done to encourage customers to buy a new machine. Perhaps the manufacturers argue if it lasts too long, then they will not, so let's kill it off by making repair charges impossibly high.
lam reminded of course of one of the first electric light bulbs, made 75 years ago and still working in a Chicago Fire Station. The company, sad to report went bankrupt. To me, the moral of the story, is not that they should have made a less durable lamp, but, that they needed better communication. Just imagine a company today advertising a product that would last a lifetime, customers would fight to buy it and the company's factories would be working 24 hours a day for years to meet the demand.


## conference timer

## YOUTH CLUES. COLLEGES.SOCIETIES -INSTTUTES.SCHOOLS.FETES

A lamp automatically lights when the speaker's preset time has expired. Has interrupt facility,
presettable "wind-up" indicator and remote secondary indicator panel.

zyeryden


## PART 4

WITH the construction of Labcentre completed we can now move on to testing the various facilities which the system offers, and to the important task of setting up the function generator.

Most of the testing is straightforward and can be carried out using the other Labcentre circuits and a simple multimeter. The function generator is rather different because there are five potentiometers which can be used to coax the maximum performance from the circuit, and these must be adjusted with care to achieve the very best results.

Ideally, the setting up of the function generator should be carried out with the aid of an oscilloscope, but many constructors will not have

## By R. W. Coles and B. Cullen

access to an oscilloscope, and so a simplified setting up procedure has been produced which will provide adequate results for most purposes.

A supply of jumper leads with 2 mm plugs at each end should be made up to facilitate testing. These should be made up on several colours to avoid confusion, and will also be found useful during the use of Labcentre for experiments at a later date.

## TESTING LABCENTRE

If all wiring has been completed and thoroughly checked, a three-pin plug fitted with a 2 amp fuse can be fitted to the end of the mains cable and the power turned on. First, the 13 amp distribution sockets can be
checked by plugging in an appliance such as a table lamp and turning it on with the switches mounted adjacent to the sockets.

Please note that this is not a safety check, merely a final operational check. It is vital that the wiring to these sockets should have been very carefully checked before final assembly. A proper safety check can be carried out at this stage with the aid of a "Tester-Plug", of which there are several examples now available from electrical shops.

If the sockets are operational, the mains on/off switch can be set to on, at which point the neon-lamp within the rocker switch should illuminate to show that the mains supplies are now available to the Labcentre power supply circuits.

Using a multimeter set to an appropriate d.c. voltage range, the +5 volts +15 volt -15 volt supplies can now be checked at the front panel screw terminals, SK1 to SK4. Remember to check not only that the correct voltages are available, but also that they are wired up to the correct terminal, $(+5 \mathrm{~V}$ to yellow, +15 V to red, -15 V to black and 0 V to green).

A further check of correct power supply operation can be carried out by shorting each supply in turn to the 0 V terminal using the multimeter set to a 5 amp or greater d.c. current range. This test ensures that the current limiting circuitry within the regulator i.c.s is operating correctly.

Do not maintain the meter connections for more than a few seconds, as this could cause the regulators to get quite hot. If you wish to omit this test due to pure cowardice, you have my fullest sympathy! If, however, you do carry it out, the actual currents measured will vary depending on the type of regulators you have used.
Typically the 5 V supply will limit at about 1 amp , and the 15 V supplies will limit at about 750 mA . The precise current is unimportant, but it is wise to recheck the voltages to make sure they have returned to their operational levels when the short circuit test has been completed.

## PUSH BUTTON PULSER

With complete confidence in the power supply circuitry now established, it is time to move on to the logic circuitry.

Set the cmos/TTL switch to TTL and connect a multimeter set to the 20 volt range between the $Q$ output of the push-button pulser and 0 volts. A reading of 0 volts should be obtained which changes to +5 volts when the button is depressed and back to 0 volts when released. Transfer the positive meter lead to the $\bar{Q}$ output and check that it behaves in the opposite way to the $Q$ output.

Now switch the cmos/TTL switch to cmos and recheck the pulser outputs.

This time they should swing between 0 volts and 15 volts as the button is operated. Switch back to TTL and connect the $Q$ output to each of the l.e.d. inputs in turn using a jumper lead. Operation of the button should cause each le.d. to blink on and off. Try this again with the switch in the cmos position, there should be no difference in l.e.d. operation.

Also, to test the four switches, you can wire one input of each uncommitted switch to the +5 volt supply and connect jumper leads from the other inputs to the l.e.d.s. Operation of each switch should control the appropriate l.e.d.

## FUNCTION GENERATOR

The function generator has to be set-up as well as tested, and this means that access to the five preset potentiometers and the circuit board will be required. The function of these presets is shown in Table 4.1.

With the circuit board accessible all five preset potentiometers should be set to mid travel (note that VR1 and VR3 are muiti-turn components), before power is applied.

If no oscilloscope is available, access to the front panel will also be necessary so that the l.e.d.s and other facilities can be used to advantage to aid with the setting up process.

## LOW FREQUENCY

Connect the $f$ and $f / 8$ outputs of the clock divider to convenient l.e.d.s using jumper leads, and set the function generator range switch to x10 and the frequency control to the counter-clockwise position. With these control settings the function generator frequency is required to be about 10 Hz , and therefore the output of the $f / 8$ divider stage will be about $1 \cdot 25 \mathrm{~Hz}$.

This is a low enough frequency to be directly observed on the l.e.d. and so the setting up procedure can begin with the adjustment of VR3 until an l.e.d. on period of about half a second is observed. With the low frequency roughly set, VR4 can be adjusted to obtain equal on and off times on the l.e.d. connected to the $f$ divider output. This will be quite difficult to judge unfortunately, although it is worth spending some time to try to get it right because VR4 controls the half cycle duration ratio or "symmetry" of the sine and triangular waveforms as well as that of the square output.

If you do find this is too difficult to judge at the normal 10 Hz low frequency limit, try readjusting VR3 to get the very lowest frequency at which the 8038 still oscillates before adjusting VR4 for equal on and off times. But remember you must watch the $f$ output, not the $f / 8$, which will always have a symmetrical output due to the action of the binary counter stages.

When you are satisfied with the on-to-off ratios of the $f$ waveform, readjust VR3 while watching the $f / 8$ l.e.d. to set the lowest frequency to about 10 Hz ; l.e.d. on periods of from a half to 1 second will be quite satisfactory in view of the lack of precise dial calibration. The Labcentre function generator is not, after all, meant to provide an absolute calibration standard, but can claim to be a flexible tool for experimentation and design.

With the low frequency setting of the function generator completed the other divider outputs of $f / 2$ and $f / 4$ can be checked out by connecting them in turn (or at the same time) to the l.e.d.s on the front panel.

When this has been carried out, the jumper leads can be removed from the dividers, the function generator "triangle/sine" S2 switch should be set to "sine" and the output voltage control should be set about one quarter clockwise, in readiness for the setting of the high frequency preset.

## CAUTION!

Removal of the front panel exposes dangerous mains voltages. it is essential that power is turned off at the wall switch before removal, and that every precaution is taken to avoid contact with the mains wiring while adjustments are being made with the power applied. Fallure to observe these precautions could result in electrocution.

## HIGH FREQUENCY

Without the aid of an oscilloscope or frequency counter, the high frequency setting will have to be done by ear, and for this the audio amplifier is needed.

Set the function generator frequency dial (VR2) fully clockwise and connect a jumper from the red output socket (SK5) to the blue audio amplifier input socket SK13. (No earth connection is required.)

Turn the loudspeaker switch to the "on" position whereupon an audio tone should be heard from the speaker. Reduce the volume to a comfortable level using either the amplifier volume control or the function generator output level control and then adjust the VR1 preset to achieve a tone frequency of about 1 kHz .

Most people can judge a 1 kHz tone frequency fairly accurately, it being the brrr ... and below the seee... in about the wheee . . . area! Those with musical inclinations will no doubt be able to use their skills (and their instruments) to good effect to facilitate the more accurate setting of this preset control.

Table 4.1. Preset Functions

| Potentiometer $\quad$ Function |  |
| :--- | :---: |
| VR1 | Set high frequency |
| VR3 | Set low frequency |
| VR4 | Set mark/space symmetry |
| VR5 | Set sine shape (+ve) |
| VR6 | Set sine shape (-ve) |

The setting-up process described above also takes care of the x 1000 range, although of course the absolute frequencies achieved will depend upon the value of the capacitor C15.

The last two presets to be adjusted are VR5 and VR6 which control the action of the sine control and the action of the sine converter in the 8038.

The sine converter, you may remember, "whittles away" sharp corners of the trangular waveform which is produced across the capacitors C14 and C15 to produce a very good approximation to a sine waveform. The positive and negative half cycles of the sine waveform are tailored by individual transistor "trees", and the presets VR5 and VR6 are provided so that each half cycle can be separately adjusted.

Adjusting these presets for the best sine shape is difficult without an oscilloscope, but fortunately the range of adjustment is not large, and if you are unable to detect any difference by ear, the skeleton pots can be set to mid travel and left there.

With the function generator now set up, the Labcentre console is now ready for use and so the mains power can be turned off at the wall switch and the system reassembled.

## FINISHING THE JOB

If you plan to invest in a briefcase for your Labcentre system, then you will no doubt wish to achieve a really professional looking finish. Careful attention to detail at this stage will allow you to produce a portable laboratory which is not only extremely



Fig. 4.1. Construction details for the tool board. The elastic retaining strips are partioned and anchored by means of twisted insulated wire. Use of Velcro enables easy removal of the board. The component boxes are glued to the baize.
practical, but also attractive enough to be the envy of all your friends!

After fitting the console, it is necessary to take stock of what other items need to be carried in the Labcentre case. The most useful accessory is an analogue multimeter, and no doubt many constructors will already possess one of these. Those who haven't will find the small investment well worth it.

Other items which you may wish to use include: a logic probe, a signal injector or tracer, a digital multimeter or counter and perhaps even a calculator for those tricky design calculations!

When you have decided just what other test instruments are to go into the Labcentre case you should obtain a quantity of polyurethane foam and fit it around the central console using a hobby knife or razor blade to trim it to shape.

Cut-outs for the various accessories can also be made by careful application of the blade, and delicate items can also be given a foam "lid" to ensure their protection during the possible rough and tumble of the Labcentre existence.

In the prototype, the only extra equipment added was a multimeter, but a quantity of small plastic boxes were obtained and used to house components such as resistors, capacitors and semiconductors.

## TOOL BOARD

The final touch, and a very important one, is the tool board which fits in the lid of the case. The owners of
the prototype wanted to be able to remove the tool board and so it was mounted using Velcro strips. This allows the briefcase to be used for other purposes when necessary, and it can be useful to be able to move the tool board to the workbench when a constructional session is in full swing.

The construction of the toolboard is detailed in Fig. 4.1. The exact size may need to be changed to fit your own particular case design, and of course, everyone will have their own ideas about what tools are needed and so the drilling details for the anchor points may need modification.

The hardboard can be cut to size with a saw, or better still, one of the nibbling tools available for the purpose. The covering material used on the prototype was green baize, which gave a very businesslike finish.

Various other materials could be used, and another possibility would be to paint the hardboard, say, green, and then paint on the tool outlines in white so that you can see at a glance if any tool is missing.

The tools can be supported with elastic as shown, and the bias binding is added to make the removal of the board a simple process.


The prototype tool board removed from the lid of the briefcase for use on the construction bench.

## SOME SUGGESTED TOOLS

Tools for use with Labcentre can be chosen from the following list to suit personal preferences:
Solder sucker
Soldering iron (miniature)
Screwdrivers (various)
Pliers (pointed nose)
Wire cutters
Wire Strippers
Spanners (BA or metric)
Modelling Knife
Trimming Tools
Spot face cutter (for stripboard)
Terminal pin insertion tool
Wiring Pen
Wire Wrap Tool
Heat Shunt
Wire Bending Gauge
I.C. Insertion Tool
I.C. Extractor Tool

Small files
Contact burnisher
Tweezers


## USING THE LABCENTRE

There is no set procedure for using Labcentre, everyone will have their own ideas on how to use it, and what to use it for. It can be used in many different fields, such as design, experimentation, tutoring and trouble shooting, and for all kinds of circuits including the digital and analogue variety. Most of the detailed uses for Labcentre haven't even been invented yet!

Despite this wide scope, a few cperational guidelines may help.

## PRECAUTIONS

1. To ensure circuit safety, always turn off the mains switch while making interconnections.
2. Be sure to check the position of the cmos/ttl switch before using the pulser or divider outputs. Remember, the cmos position will damage TTL and cmos circuits powered from 5 V rails.
3. Be careful not to short the pulser, divider, or function generator outputs to any of the supply rails as this can result in ruined i.c.s.
4. Do try and avoid power supply short circuits. The regulators are short circuit protected to 0 volts but extended shorts could cause overheating. Short circuits between power rails could destroy the regulators.
5. Never try to wire up any circuits on the breadboard which use mains or other high voltages.

Fig. 4.2. Using a Zener diode and series current limit resistor to provide intermediate voltage levels: (a) positive (b) negative with respect to OV.

5. Voltage rails between 5 and 15 volts can be generated with a Zener diode and a resistor as shown in Fig. 4.2. The resistor should be chosen so that the voltage drop across it is equal to $\left(15-V_{z}\right) V$ at the maximum load current of your circuit, where $V_{z}$ is the Zener voltage. To get a useful output current range, 1 watt Zener diodes are best for general use.



## LOW COST METAL LOCATOR By R. A. Penfold

THE MAIN design requirement for this metal locator was for the most simple and inexpensive unit that would give a good level of performance.

A beat frequency oscillator (b.f.o.) type circuit is used, and this is based on a single inexpensive cmos i.c. and very few passive components. The performance of the unit is very much in line with that of other simple locators with small coins; ${ }^{1}{ }_{2} \mathrm{p}, 1 \mathrm{p}$, or similar are detectable at depths of up to approximately 25 mm , and larger coins such as $2 \mathrm{p}, 10 \mathrm{p}$ and 50 p pieces detectable at depths of up to about 50 or 80 mm . Large pieces of metal can be detected at depths in excess of 300 mm .
The unit is self-contained apart from the crystal earpiece which is used to provide the output tone. The presence of metal is indicated by a small change in the pitch of this tone.

## B.F.O. PRINCIPLE

This type of locator uses an inductive/capacitive ( $L C$ ) type oscillator as the basis of the unit. The frequency at which the oscillator operates is dependent upon both the value of the inductor and
the capacitor used in the tuned circuit, and a change in either will result in a change in operating frequency. The inductor is a large air-core type and it is usually termed the search coil.

## FREQUENCY SHIFT

If a piece of metal is brought close to the search coil it will cause a change in its inductance, and hence also a change in the operating frequency of the oscillator. The problem with this system is that the frequency shift produced will usually be only about $0 \cdot 1$ per cent or less, and is too small for the human hearing mechanism to readily perceive.

This is overcome by using the b.f.o. principle which is outlined in Fig. 1. Here the output from the oscillator is fed to one input of a circuit known as a product detector. The other input of the product detector is fed from a second oscillator which is termed a beat frequency oscillator. The output from the product detector will contain a number of frequencies, most of which are filtered out by an r.f. filter leaving just the difference frequency which is fed to the earpiece.

In practice the search oscillator operates at quite a high frequency, and this design employs a nominal frequency of 125 kHz . The b.f.o. is adjusted to a frequency just above or below the search oscillator frequency so that the difference frequency at the output of the product detector is a low audio one, which will be reproduced by the earpiece.

If a piece of metal is brought close to the search coil, causing a change in frequency by about $0 \cdot 1$ per cent, with a nominal operating frequency of 125 kHz this represents a frequency shift of 125 Hz ( $0 \cdot 1$ per cent or one thousandth of $125,000 \mathrm{~Hz}=125 \mathrm{~Hz}$ ). The difference in the frequency of the two oscillators will obviously change by this amount, causing the output tone from the earpiece to change by about 125 Hz . Since the tone from the earpiece could have previously been less than 125 Hz if desired, this shift in output frequency can produce a significant change in the output tone that will be readily apparent to the user.

## CIRCUIT DESCRIPTION

The complete circuit diagram of the detector is given in Fig. 2. The only active component in the unit is a cmos quad 2-input nor gate. Although this is a logic i.c. it readily lends itself to use in this type of circuit.

Gates G1a and G1b of the i.c. are used as the b.f.o., and these are connected in a conventional cmos astable or free running multivibrator circuit. This can be tuned several kilohertz either side of the search oscillator frequency by means of VR1.


Fig. 1. Principle of the beat frequency oscillator circuit. Here two oscillators run at very nearly the same frequency thus producing an audible beat note in the earpiece. Any slight shift in the LC oscillator frequency will cause a change in the beat note heard. This change in frequency is caused by bringing a piece of metal in the vicinity of the search coil. This alters the inductance of the coil, thus changing the frequency of the tuned circuit.

Conventionally an LC oscillator is used as the b.f.o., but this is comparatively expensive due to the need for an inductor and a variable capacitor. This $C R$ alternative seems to work perfectly well and give good stability provided it is operated from a stabilised supply. The unit is therefore powered from a simple Zener shunt stabiliser circuit formed by R5 and D1.

Gate G1d is used as the basis for the search oscillator and this uses a rather unusual configuration. The two inputs of the gate are connected together so that it functions as an inverter. Search coil L1 then biases the inverter to act as an amplifier. The series capacitance of C4 and C5 acts as the
unusual configuration. The output from the b.f.o. is coupled to one input of G1c via R2, and the output from the search oscillator is coupled to the other input by way of R3. The output from a nor gate will be high, virtually equal to the positive supply voltage if both inputs are low, almost at the negative supply voltage. Any other combination of input states will produce a low output state.
Therefore, an input from the search oscillator as in Fig. 3a and a slightly higher input frequency from the b.f.o. as in Fig. 3b would produce the output waveform of Fig 3 c . If this output is then passed through a suitable r.f. filter so that the rapid changes in the signal are smoothed out and the pulses are


Fig. 2. Complete circuit diagram of the Metal Locator.
capacitive part of the tuned circuit and the junction of these two acts as an earthed tapping.

This produces a phase inversion through the tuned circuit at its resonant frequency, which with the inversion through the amplifier results in positive feedback and oscillation at the resonant frequency of the tuned circuit. This has a nominal frequency of 125 kHz , and the prototype had a measured operating frequency of $123 \cdot 86 \mathrm{kHz}$.

There will be minor variations between various units built to this design, but provided the correct components are used and the search coil is constructed exactly as described, the unit will operate well within the permitted frequency band used for metal locators.
Gate G1c is used as the product detector, and again this is a rather
integrated to produce a sort of averaging of the output level, a waveform something like that of Fig. 3c will be obtained. This final signal is the required difference signal, as is demonstrated by the fact that the total number of pulses in $a$ and $b$ equals the number of pulses in b.

## R.F. FILTER

Resistor R4 and C3 are the r.f. filter network. The inputs to Glc are couped via high value resistors rather than directly as it is advisable to use a very loose coupling here. With a tighter coupling there is a tendency for the two oscillators to lock onto the same frequency even when they should be a few hundred cycles apart. This would make the unit difficult to use and insensitive. The use of

loose couplings and supply decoupling capacitors C1 and C6 virtually eliminates this effect, and the oscillators have to be within a very few cycles of each other's frequency before locking will occur.
Switch S1 is an ordinary ON/OFF switch and is ganged with VR1. The unit has a current consumption of about 5.5 mA which gives many hours of operation from a PP3 battery.



COMPONENT BOARD
All the small components are assembled on a piece of 0.1 inch matrix stripboard using the component layout given in Fig. 4. The board has 16 strips by 24 holes and is cut down from a larger piece using a hacksaw.
The two $3 \cdot 2 \mathrm{~mm}$ diameter mounting holes and the breaks in the copper strips are then made, after which all the components and link wires are soldered into place

The i.c. is a cmos device and will be supplied in some form of protective packaging to prevent damage due to high static voltages. This should be the last component to be soldered into circuit and it should not be removed from the packaging until then. It is advisable to use a soldering iron having an earthed bit when connecting this device. A socket could be used for the i.c., but due to the low cost of the device this is probably not justified.


The case is a small plastic box which measures about $111 \times 60 \times$ 27 mm , and VR1 and the output socket are mounted at one end of the lid. This leaves sufficient room for the completed component panel to be mounted on the lid using two short 6BA screws.

## 

## Resistors

R1 $6.8 \mathrm{k} \Omega$
R2 $1.2 \mathrm{M} \Omega$
R3 $680 \mathrm{k} \Omega$
R4 $15 \mathrm{k} \Omega$
All are $\frac{1}{4} W$ carbon $\pm 5 \%$

## Potentiometer

VR1/S1 $5 \mathrm{k} \Omega \mathrm{Lin}$. carbon with double-pole switch

## Semiconductors

IC1 CD4001 Quad 2-input NOR gate
D1 BZY88C6V86.8V 400 mW Zener diode

## Capacitors

C1 100 nF polyester
C2 390 pF ceramic plate $\pm 5 \%$ or better
C3 1 nF ceramic plate
C4 10 nF polystyrene $\pm 5 \%$ or better
C5 $5 \cdot 6 \mathrm{nF}$ polystyrene $\pm 5 \%$ or better
C6 $100 \mu \mathrm{~F} 10 \mathrm{~V}$ elect.

## See <br> page 333

## Miscelianeous

B1 9V PP3 battery
L1 $\quad 35$ turns $10 / 0.1 \mathrm{~mm}$ wire (see text)
SK1 3.5 mm jack socket
Stripboard, 0.1 inch matrix 16 strips by 24 holes; small plastic box, $111 \times$ $60 \times 27 \mathrm{~mm}$ or similar; battery connector to suit B1; small round control knob; crystal earpiece with 3.5 mm plug; connecting wire.
Materials for framework; (see text), $800 \times 25 \mathrm{~mm}$ dowel (broomstick or similar, $250 \times 15 \mathrm{~mm}$ dowel, hardboard or thin plywood $200 \times 180 \mathrm{~mm}$ ( 2 off) insulating tape, bicycle handgrip, aluminium foil if required.

It is advisable to earth the body of VR1 to the negative supply rail of the unit, as otherwise a slight hand capacity effect will make precise tuning of the unit a little difficult. This earthing can be accomplished by looping the bare end of a short insulated wire around the mounting bush of VR1 when it is being mounted. This lead will then be trapped against the body of VRI, and the free end of the lead can be earthed to the appropriate tag of SK1.

## SEARCH COIL

To make the search coil a temporary coil former having a perimeter or circumference of 508 mm is required. The author used a 152 $x 102 \mathrm{~mm}$ chassis that was at hand, but a suitable former can easily be constructed from wooden battens if no ready made alternative is available.

The winding consists of 35 turns of $10 / 0 \cdot 1 \mathrm{~mm}$ p.v.c. insulated wire wound fairly tightly around the former (but not so tightly that it will be difficult to remove the coil) with leadout wires one metre long. Four bands of insulation tape are used to hold the winding together so that it can be removed from the former.

These initial steps of the search coil construction are illustrated in Fig. 5.

# LOW COST <br> METAL LOCATOB 

Fig. 6. Constructional details for the framework as used on the prototype. Dimensions are not critical and if required can be varied to suit individual requirements.
 should be fitted at this stage.


Layout of components on the completed circuit board for the Low Cost Metal Locator.

## SHIELD

The search coil is formed into an oval shape about $180 \times 150 \mathrm{~mm}$, and it is then a good idea to fit it with a shield. This is not essential, but without a shield there will be a small change in the output frequency when the search coil is placed close to the ground, even in the absence of any metal there. This would make the unit more difficult to use with a consequent indirect loss of performance.

A shield can simply consist of strips of aluminium cooking foil used to completely cover the coil, except over the band of insulation tape at the point where the two leadout wires emerge from the coil.

A lead about one metre long must be attached to the shield so that it can be earthed to the negative supply. This can be achieved by baring about 40 mm at one end of a p.v.c. multistrand lead one metre or so long. The bare wires are twisted together, heavily tinned with solder, and then tightly wound around the shield close to where the two leadout wires emerge from the coil.

To complete the search coil it is tightly bound with a layer of insulation tape to hold the whole assembly together.

## FRAMEWORK

Details of the wooden framework used on the prototype locator are provided in Fig. 6. This can of course be altered slightly to suit the particular materials to hand or locally available, but make sure that the search coil is held very firmly in place and that
its leadout wires are not allowed to simply flap around.

Strong, rigid construction is essential, or as the detector is moved around there will be small changes in the output frequency without any metal in the vicinity of the search coil.

The main part of the case is screwed to the handle part of the framework using a couple of small woodscrews. A hole is drilled in the side of the case near to the upright pole to provide an access point for the leadout wires from the search coil and shield (if fitted). The final wiring of the unit can then be completed. The lead from the shield connects to the earthed tag of SK1.

It is a good idea to fit a handlebar grip (as sold by cycle shops) over the end of the handle. The handle will almost certainly be too small in diameter for the grip to be a good tight fit over the end, but this can be rectified by using some bands of insulation tape around the end of the handle.

## IN USE

With a crystal earpiece plugged into SK1 and the unit switched on,

## OPERATING LICENCE

Under the terms of the Wireless Telegraphy Act 1949, a licence is required to operate this Metal Locator in the UK. The unit has been designed to operate on a frequency of 125 kHz , as such the design should not be altered. Applications for a licence should be made to the Home Office, Waterloo Bridge House, Waterloo Road, London SE1.
by adjusting VRI (PITCH) it should be possible to produce a strong audio tone from the earpiece. It is likely that it will be possible to produce an output tone at virtually any setting of VR1, but most of these signals are weak spurious ones, and it will be obvious when the main response has been found due to the far higher output volume.
The variable resistor VR1 will control the pitch of the output tone, and if this is adjusted for the lowest possible tone, the two oscillators should lock to the same frequency, producing a null in the output. In use VRI is just slightly offset from this setting so that a very low pitched tone is produced.


Positioning of the circuit case and the handle grip.
A very low output note will provide the best sensitivity since, for example, a frequency shift of 10 Hz is quite significant on a 20 Hz signal, but would probably not be perceptible on a signal at a frequency of, say, 2 kHz .

With VR1 set one side of the null point it will be found that the presence of most types of metal produces a reduction in the output tone, whereas an increase in the tone will usually occur if it is offset in the other direction.

Sensitivity is best if it is adjusted so that a reduction in the output pitch is provided, as a small reduction in the pitch of a tone is a little more easily detected than a similar increase in pitch.

# Mnl|-MOOULES By George Hylton in combinateion 

So far in this series, a number of audio modules have appeared. We thought it might be a good idea to demonstrate how these can be connected together and used in combination.

The Audio Effects Oscillator (November 1978, p. 818) is a good starting point because it forms a handy source of signals. It can be used in conjunction with the Audio Modulators (February 1979, p. 86) to manipulate the signal from some other audio source such as a radio tuner or the output of the Microphone Amplifier (December 1978, p. 862).
Several signals can be mixed and faded in or out with the help of the Passive Mixer (October 1978, p. 746). And, of course, for more ambitious setups you can build more modules and add them to your system.
If you don't want to be stuck with battery power, the Versatile Power Supply (March 1979, p. 157) enables them all to be worked off the mains.
This really exhausts the ostensibly audio modules, but if you happen to have built the Meter Amplifier (this issue) it can be pressed into service as a lowpower monitor amplifier for driving a high-impedance speaker (though its quality of reproduction is poor) and even the Continuity Tester (January 1979, p 24) could be operated as a second variable frequency audio source!

One important general point is that none of the audio modules produces any appreciable output power. They are designed for use in conjunction with a normal audio chain consisting of a power amplifier with its own power supply unit and preferably its own tone controls because these are always a useful adjunct to a sound effects "kitchen" where weird noises have to be cooked up.

## DRAMATIC EFFECTS

The Audio Effects Oscillator in its original form has a frequency control, an output level control and an input terminal for applying a modulating voltage if required. The modulation facility is optional and the oscillator can be used as a straightforward tone source.
It is very useful to build two oscillators, one for frequencies in the speech and music range and the other for very low frequencies. In my original article । suggested a range of $1-30 \mathrm{~Hz}$ for the v.I.f. version. This, with suitable application of tone controls, can produce passable imitations of clocks ticking, hearts beating and similar repetitive noises.

The other, higher-frequency oscillator can be used on its own to make whistles
and lower tones. Switched on and off it can give "time pips" or "morse code". When modulated by the v.I.f. oscillator it can produce quite a good imitation of a two-note police siren. The quality of the modulated output varies with the amount of modulation voltage. A few tenths of a volt produces continuously varying pitch. A volt or so produced on-off modulation; interrupted bursts of tone.
easily get into a muddle, with leads going in all directions, so for any sort of permanent or semi-permanent installation it may be preferable to settle for a fixed arrangement whose facilities though limited are quickly got used to. Building an extra Passive Mixer will enable a fair degree of flexibility to be preserved.
A possible arrangement is shown in Fig. 1 (but it must be stressed that the user is at liberty to make any interconnections he likes and that this one is merely for the purposes of illustration). Here the v.l.f. oscillator output is tied permanently to the audio oscillator's modulation input. The audio oscillator can still be used unmodulated, however, by turning down the v.l.f. output.


Using a pair of effects oscillators (one v.l.f., the other audio) you can equip yourself to make tape recordings of stories or plays with at least some of the sounds needed for dramatic effects. You can, with the aid of a gram deck and the mixer, bring in background music as well. In no time at all you can be your own radio programme producer!

## INTERCONNECTIONS

The most versatile means of interconnecting the various modules is oldfashioned flexible leads with crocodile clips at each end. With these you can connect anything to anything and see what happens. The snag is that you can

The output of the microphone amplifier is taken to one of the modulators. Modulating speech with a very low frequency can produce "Dalek" type sounds. Modulating with a high frequency produces odd sounding distortions. Modulating with both at once . . . well, why don't you try it and see!
Connecting the audio oscillator via a slide switch is useful for producing morse-code-like sounds. Choose a switch which can be operated quickly and silently.
Modulating music (from a radio, tape deck or gramophone playback unit) at v.l.f. produces a tremolo effect.
No doubt readers will develop their own tricks for making use of these audio modules. If you hit on something useful, please tell usabout it!

# (4) THE GREAT BRITISH Electronics ${ }^{2}$ azaqar 

ALEXANDRA PALACE, LONDON N22. 28-30 JUNE 1979
YOU WILL FIND US RIGHT IN THE CENTRE OF THE GREAT HALL ON STAND NUMBER 163
O. N. Bishop, author of "Doing It Digitally" will deliver the EDUCATIONAL
SEMINARS $\begin{aligned} & \text { O. N. Bishop, author of Deiter } \\ & \text { EE lecture/demonstration Getting Started In Electronics } \\ & \text { Construction daily. }\end{aligned}$


MANy otherwise very good multimeters fall down rather badly when it comes to measuring small a.c. voltages. I have an excellent general purpose multimeter, for example, which in theory will measure down to 1 volt a.c. full-scale. Unfortunately, on this $1 V$ a.c. range its internal resistance is a mere 300 ohms or so; this is so low that in many situations accurate measurement is impossible because the meter's impedance loads the circuit to which it is connected so much that the voltage I'm trying to measure falls.
Another problem is that a good many of the audio voltages one would like to measure are 100 mV or less and a 1 -volt a.c. range cannot measure them accurately.

An answer to this problem is to provide the meter with a pre-amplifier which does three jobs.

Firstly, it amplifies by some convenient small amount such as 10 times, which is enough to convert the meter IV range to a $0.1 \mathrm{~V}(100 \mathrm{mV})$ range.

Secondly, it has a high enough input impedance (in the present circuit about 90 kilohms) to avoid serious loading of the majority of sources of audio signals.

Thirdly, it has the ability to supply large enough output currents at a low enough impedance to drive the meter.

The amplifier in this Mini Module does all this but can do more besides.

In addition to driving an a.c. voltmeter it can also act as a self-rectifying circuit to drive a d.c. meter with a current proportional to the a.c. input voltage. In this way almost any standard d.c. milliammeter or microammeter can be turned into a sensitive a.f. voltmeter with a full-scale deflection of around 100 mV r.m.s.

The same d.c. meter can also be converted into a linear frequency meter or a capacitance meter with the aid of an external audio oscillator, which need not be accurately calibrated.

## THE CIRCUIT

The circuit consists of an operational amplifier in i.c. form (which provides the required gain) and a complementary class B output stage (which provides the output current). Negative feedback (via R5 and R4 in Fig. 1) sets the gain and removes most of the distortion of the unbiased output pair TR1, TR2.

With the values of R5 and R4 shown the gain is 11. This is a very useful gain when a d.c. meter is used for a.c. voltage measurement because it corrects for a cause of error which would make the meter "read low"

For use as a pre-amplifier for an a.c. meter however a gain of exactly 10 is needed and can be arranged by increasing R4 to 11 kilohms (for example, two 22 kilohms in parallel) or by connecting 1 megohm across R5.

Fig. 1. Circuit diagram of the meter amplifier


## CONSTRUCTION

The prototype is housed in an electrician's metal switch box of the type designed to hold two standard sized square light switches. These are the boxes ("patresses") which are often buried in house walls to take flush-mounting switches.

The older kinds, designed for deeper switches than the modern "plaster depth" switches are more useful for electronics construction and are still obtainable. The box I used measures $130 \times 70 \times 32 \mathrm{~mm}$ deep but anything similar should be usable.

The front panel is a piece of Formica and to the back of this is glued the circuit board, which in this case is a piece of plywood. This was converted into a component board by tracing the layout diagram and sticking it to the plywood, then inserting pins at every junction to act as solder tags.

Miniature household pins were used, of the kind known to stationers as "lills". These are about 15 mm long and can be inserted with pliers. They are pushed right in until they hit the Formica.

For lead-through connections I used coppered hardboard pins. To avoid damage to the panel and base, holes were drilled for the pins, just large enough to enable them to be pushed through. They were then locked in place by applying a "collar" of solder where they emerged on either side.

It seemed hardly worth providing an internal battery for a piece of equipment which may only occasionally be needed so the 9 V supply was also terminated on pins to enable an external battery to be connected.

The component layout (Fig. 2) is not critical and may be altered to suit some other shape and size of container.

## USING THE AMPLIFIER

## (1) As a Meter Pre-amp

For use as a "x10" meter pre-amp the a.c. voltmeter is connected to the output terminals. The gain is set to 10 as described earlier, by adjusting R4 or R5. For gains other than 10 which may be needed for some meters, keep R5 at 100 kilohms but
use a new value for R4, calculated from: R4 (new $=100$ kilohms $(A-1)$ where $A$ is the required gain.
This formula should work for gains up to about 100. Example: your a.c. meter has a range $0-300 \mathrm{mV}$ and you want to use the meter to convert this to $0-50 \mathrm{mV}$. The gain required is $300 / 50=6$ and the formula shows that the new R4 must be 100 kilohms/ $6-1)=20$ kilohms.
Note that the meter amplifier will not give outputs of much more than 1V r.m.s., so whatever meter you use in this mode must have a full-scale deflection of 1V r.m.s. or less.

## (2) As Self-Rectifying Meter Amp

To use as a self-rectifying meter amplifier the meter must be a d.c. moving coil type, with a linear scale and full-scale deflection of not more than about 10 mA . The d.c. voltage drop across the meter should not exceed $500 \mathrm{mV}(0.5 \mathrm{~V})$, at full-scale current. (Most detached milliammeters and microammeters meet this requirement but some multimeters on their d.c. ranges do not.)

It is not always realised that a class B amplifier functions as a rectifier. It does so in that the current drain from the battery is in proportion to the amplitude of the driving signal. Measure the d.c. which flows through either output transistor and you have measured the strength of the signal. If the amplifier is linear (distortionless) then so is the meter indication.

Ordinary class B audio amplifiers are not suitable because the output transistors are biased so as to pass a small current even without an input signal and a meter would read this all the time. The class B output stage in the Meter Amplifier is unbiased and passes no current until an input is applied. Most of the resulting distortion is removed by negative feedback and the circuit gives good linearity in meter terms though not in hi-fi terms.

## Components

```
Resistors
    R1 10k\Omega
    R2 150k\Omega
    R3 220k\Omega
    R4 10k\Omega see text
    R5 100k\Omega see text
    R6 1 k\Omega
All carbon film }\pm5%\mathrm{ tolerance or
better, \frac{1}{4}\mathrm{ or }\frac{1}{3}\textrm{W}\mathrm{ .}
```


## Capacitors



```
Working voltage 12 V d.c. or over. Type and tolerance not important.
```


## Semiconductors

```
IC1 741 operational amplifier in 8-pin DIL package
D1 1 N914 or any small silicon diode
TR1 2N3704 silicon npn
TR2 2N3702 silicon pnp
```


## Miscellaneous

Electrician's box for two switches, deep type (or any box about $130 \times$ $70 \times 30 \mathrm{~mm}$ ). Formica for front panel (Light coloured plain design so that lettering can be put on). Bolts for fixing panel (usually 4BA); 4 off. Plywood for base. Not too thin. Pins for use as tags. 15 mm long "lills". $\frac{3}{4}$ inch hardboard pins, coppoered, for lead-throughs. Earth tag.

Since no output current can flow unless there is an a.c. load it is necessary to connect a load resistance to the output terminals when the circuit is used in this way. The required value depends on the meter but is easily calculated:

Load resistance $=1000 /$ (twice meter full-scale current).

If the meter current is in mA the answer comes out in ohms; if in $\mu \mathbf{A}$, in kilohms. Example: Meter current,


Fig. 2. Plywood circuit board glued to Formica front panel. All components and wiring soldered to miniature household pins.

2 mA full-scale d.c.; load required $=$ $1000 / 4=250$ ohms. Remember to set the gain at 11 for this application. The d.c. meter is connected to the meter terminals.

## CAPACITANCE

If the load is a capacitance rather than a resistance the current taken depends on the frequency of the input signal since the reactance (impedance) of a capacitance goes down as the frequency goes up. If an audio oscillator with adjustable frequency and output level is available this effect enables the Meter Amplifier in its self-rectifying connection to be used as a capacitance meter.

Some sort of standard capacitance is needed to set up the circuit. This can take the form of any closetolerance capacitance of very roughly the size you want to measure.

Connect the standard capacitance to output. Apply about 100 mV of audio to input. Adjust the frequency until the scale reading fits the standard: for example: if your standard is 47 nF adjust for a reading of 47 . Leave the audio oscillator as set and remove the standard. Connecting the capacitor to be measured will now give a reading proportional to its capacitance; for example: 33 nF will read 33 , and so on. The waveform of the oscillator is not important. (Readers who built Mini Module 2, the Audio Effects Oscillator, may like to try using it for this purpose.

## FREQUENCY

Frequency can be measured by much the same means as capacitance. If a certain frequency gives half-scale reading when a certain capacitance is connected as the load then twice the frequency will give full scale reading and all lower frequencies will give readings in proportion.

This is strictly true only if the strength and waveform of the input signal are kept constant. This is not always possible, but variations can be largely ironed out by applying an oversized input signal, big enough to overload the op. amp. The op. amp output will then be a square wave of constant amplitude.

An input voltage of around IV is suitable: don't apply huge inputs or you will damage something.

Example: You want to use the 50 Hz mains (at low voltage from an insulated step-down transformer) as the standard of frequency. Apply about IV. Connect various capacitors to the output until you find one (or several in parallel) that makes the meter read " 50 ". It will read other frequencies in proportion, e.g. 33 Hz will now read " 33 " and so on.

## Next month: Voltage Splitter

# MIEROPROGESSOR Bistas 

$\mathrm{A}^{\mathrm{s}}$explained in Part 3, static RAM will be the likely choice for most hobby purposes.

The 2112 (Fig. 4.1) is a typical example of a static ram. This device is organised as 256 words of four bits and is therefore a 1 K ram. Two of these chips used together will provide 256 bytes of memory which can be used for programs or data in small microprocessor systems. By adding further pairs of 2112 devices and an external chip-select decoder, a memory array of almost any size can be constructed.

If you study the block diagram of the 2112 you will see that there are a total of eight address inputs ( $2^{3}=256$ ) and four bidirectional lines which provide the data input/
output connection to the microprocessor data bus. The four remaining pins are used for 5 volt power ( $V_{c o}$ and GND), Read/Write selection (R/W), and finally Chip Enable (CE) which gates the data at the current address on to the I/O bus, and can therefore be used as a chip select input in a large memory array.

Notice that the memory latches themselves are organised into a $32 \times 32$ array (not $256 \times$ four) and that the low address bits select a complete row of 32 locations from which the three high address bits select the required four data bits.

The select circuits are actually logic gates, arranged as a decoder in the row case, and as a data selector for the column case. The
data out of the column selection circuits is connected back to the data I/o pins via three-state buffers controlled by CE and R/W to ensure that memory data will only be gated on to the bus when this chip is selected and is in the READ state. This ensures that during a write operation the data out of an addressed location does not conflict with the input data, the outputs of the three-state buffers being forced to the third, or high impedance, state.

Well, so much for rom and ram. Now let's have a look at input/output circuits.

## INPUT/OUTPUT PORTS

Microprocessors operate very rapidly. So when they carry out an


Fig. 4.1. The Intel 2112 1,024-bit Static Random Access Memory. No "refreshing" is required with this type of RAM.

Fig. 4.2. The Intel 82128 -bit Input/Output Port. All principal functions of a microprocessor system can be implemented.

## MICROPROFILE: "BUILD YOUR OWN ROM"'

Fig. 4.4. Circuit for the homebrew ROM, made up from standard TTL components.


NOTE DIODE POSITIONS SHOWN ARE ARBITRARY 8 HAVE NO SIGNIFICANCE

At first reading, information about ROMS, PROMS, EPROMS and rams is all pretty confusing isn't it. But don't despair, it's really quite simple, and the best way to prove it to yourself is to build a ROM of your own so that you can find out at first hand just what makes these mysterious circuits "tick."

The rom design presented here is not really intended for use in a real microprocessor system because of its limited capacity, but it's not just an ornament either, and I have no doubt that you will be able to dream up all sorts of uses for it. (One possibility would be to expand the design to provide, say ten seven-bit words so that you can make a "custom programmed" seven-segment display decoder.)

The circuit of our home-brew ROM is shown in Fig. 4.4. As you can see, it uses readily available TLL parts, four 7420 dual four-input NAND gates, one 7404 hex. inverter and one 74125 quad three-state buffer. The only other things you need are four 1 kohm resistors and some IN4148 diodes.
The circuit shown provides a 32 -bit ROM organised as eight four-bit words, but this is easily expanded by adding an extra 74125 device to give a 64 -bit ROM organised as eight words of eight-bits. Expanding the number of words is a little more difficult.

## CIRCUIT OPERATION

The 7420 gates are used to perform the address decode and the "chip select" function. Individual four-input gates will give a "low" output only if all their inputs are "high", in normal NAND gate fashion. Each gate is connected to a particular combination of the address inputs which are buffered and inverted by the 7404.

The three address inputs have a total of eight possible states from A0, A1, A2=logic zero or "low" to AO, A1, A2 $=$ logic one or "high". Each word in the ROM must respond to one of these eight combinations and this is what the 7420 gates are for.
Consider the address pattern 010. This is the address code for word number 2 in the rom and so gate number 2 is connected to $\overline{A 0}$ A1 $\overline{\mathrm{A} 2}$. Only when the pattern 010 occurs on the address inputs will all of these three signals be at a logic one, and in this event the output of gate 2 will go low, providing the common C/S input is also a one.
Notice that only one gate output will be low at any one time, and also that if we had several of these ROMS

wired up as a larger memory array, we could select between them by using external gates driving their individual $\mathrm{C} / \mathrm{S}$ inputs. The use of the C/S input then, is to allow memory expansion-easy isn't itl
Well, we have selected our word in the memory, but what happens now? The horizontal lines in the ROM matrix can be thought of as "word" lines, and the vertical lines can be thought of as "data" lines. To program a logic zero into any data bit location, all you have to do is connect a diode between the data line and the word line. To program a logic one, you don't make any connection at all. Let's see what happens.

With our 010 input code, the output of gate 2 is low and if there are any diodes connected to this word line they will pull their respective data lines low also. The data lines which are not connected to word line 2 with a diode will remain high, pulled there by the 1 kohm resistors. The result is a pattern of four logic levels applied to the input of the 74125 .

A low level on the common "three state control" input of the 74125 will transmit that pattern to any external circuit, and in a microprocessor application the OE signal as it is often called, would be in the form of a fast low-going pulse. With a handful of IN4148 diodes you can go through programming each word to have any four-bit data pattern you desire, using a soldering iron as a programming aidl

What you have made is not really a rom is it? It's actually a PROM or even a EPROM if you are good at unsoldering diodes! But if you think of yourself as the manufacturer, and if you encapsulate your completed circuit in a block of Araldite, it really is a ROM I suppose.

Those who know something about logic have probably realised that the diodes and the 74125 form OR gates with the distinction that the number of inputs on these gates varies with the number of words which happen to have zeroes programmed in those bit locations. If you don't quite see how it works, build it. You can use toggle switches on the address and C/S lines, and for test purposes you can wire OE to ground. You can use L.E.D.s at the outputs of the 74125 (don't forget a series resistor) or you can use a meter.
Once you understand this circuit, other memory circuits are easy. To visualise a rAM for example, just imagine flip-flops being used in place of diodes!



Fig. 4.3. The Intel 8251 Programmable Communication Interface. This is a Universal Synchronous / Asynchronous Receiver/Transmitter (USART) chip designed for data communications in microcomputer systems.
output operation, to send an eightbit word to a peripheral device such as a seven-segment l.e.d. display for example, the data word remains on the microprocessor data bus for only a few microseconds. This isn't much use to the display as it stands of course, and so it is necessary to add an eight-bit storage latch to the system which is strobed at just the right time to catch the data sent by the microprocessor and hold it for the peripheral as long as may be necessary.

## LATCH REQUIREMENTS

Now an eight-bit latch is no great problem, a flick through your old "steam-TTL" data book will reveal lots of likely candidates, but before declaring "problem solved" let's think about what other features we need in this application.
First of all, the latch inputs must not load the data bus too much. (Rule out the rusty old 7475 latches you had in the junk box!)

Secondly, the microprocessor may need many of these output ports, and so it sends out a port address to select the one it is about to talk to. We therefore need a latch which can be "selected" as well as "clocked."

Finally, we need to consider the requirement for input ports because latches are used for output and input, and a device which can do both jobs will be an asset.
For input ports, the latch is simply connected the "other way round" so that its inputs are con-
trolled by the peripheral (say a keyboard) and its outputs are connected to the data bus. In this case the microprocessor does not want the latch outputs controlling the bus when the port is deselected and so we add the further requirement of three-state outputs, so that a high impedance is presented to the bus for most of the time.

## INPUT/OUTPUT CHIP

One device which meets all of these requirements is the 8212 eight-bit input/output chip made by Intel and others. (Fig. 4.2.) When used as an output port the inputs DI 1 to DI 8 are connected to the data bus, and the outputs DO 1 to DO 8 are connected to whatever peripheral component (lamps, displays, relays, printer, etc) the system needs to drive.
To make an input port, just connect up the other way round, although in this case the peripheral components may be switches, sensors, or a keyboard, for example.

Address selection is carried out by means of the DS1 and DS2 inputs, which may be enough by themselves in some smaller systems, or may be driven by the outputs of a memory decoder chip in larger ones.

Data is clocked into the latches with the STB line which is connected to the control bus in the case of an output port, or is controlled by the peripheral device for an input port.

A reset line, CLR, is provided to clear the latches. This can be connected to the microprocessor reset line so that the system is initialised when power is applied.

The three-state output buffers only need their third state when the 8212 is used as an input port. In this case the DS1 and DS2 inputs control the three state outputs, only allowing the latch data through to the bus when the port is selected.
For output port use, this threestate facility isn't required, and so a mode input, MD, is provided so that microprocessor output data can remain available whether the 8212 is selected or not.

The other flip-flop and gates are for use in a very powerful type of microprocessor input/output scheme called Interrupt Driven I/O, and we may return to this in later articles.

## L.S.I. CHIPS

The 8212 is a very handy chip to have around, but really it's no more than a super-TTL part and is quite modest in its capabilities despite the appearances of its bulky 24 -pin package!
By using mos LSI technology, chip manufacturers are able to produce much more useful devices than the simple 8212. Chips which can be changed from input to output under software control, and chips with three eight-bit I/O ports in a single 40 -pin package are all available, at a price!

These input/output "systems" make microprocessors even more powerful and at the same time they reduce the amount of software needed for some jobs. Some of the many complex peripheral chips now available do special dedicated jobs, and one such device is known as a UART (Universal, Asynchronous, Receiver/Transmitter).
A UART lets a microprocessor system talk not in parallel eight-bit bytes, but in a serial bit-by-bit fashion to peripherals such as teletypewriters or vdus. The uart contains a serial transmitter, a serial receiver, and all the control circuitry to make the communication timing and formatting "transparent" to the microprocessor. One example of this type of device is the Intel 8251 Programmable Communication Interface. (Fig. 4.3.)

To be continued


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# By Pat Hawker, gзva 

RECENTLY, BBC audience research has suggested that more than 40 per cent of all radio listeners tend to keep their receivers tuned to just one single channel. It is by no means certain whether this is due primarily to preference (channel loyalty), lack of information on programme schedules, ingrained sloth, or as I suspect, lack of confidence in their ability to tune to other channels.

Very few listeners really bother to memorise wavelengths or frequencies and at best tend to remember the "spot on the dial". In some parts of the world "channel numbers" are used, particularly for v.h.f./f.m. stations. Almost certainly a large number of radio listeners cannot readily find their way among the radio waves without the distortion that comes from inaccurate tuning or (on v.h.f.) the unpleasant sounds that stem from signals below the "f.m. threshold" unless some adjustment or repositioning is made at the same time to the telescopic rod aerial.
A couple of years ago, Duncan MacEwan, BBC's Chief Engineer, Radio, described his "ideal" radio set of the future. First on a long list of requirements was the elimination of the conventional dial and the elimination of all manual tuning in favour of touch-button or "memorised" selection.
In support of this view, I consider that one of the more important benefits to the viewer that came from moving television on to u.h.f. instead of v.h.f. was the general adoption of push-button channel selection instead of the old "turret plus fine tuner". Anyone of any age can readily change channels even if they do have to get up from their chairs to do so. The recent attempts to popularise quite complex remote control systems for television (some including "multifunction" buttons which do different things at different times) seems to me a retrograde step, at least for older viewers.

For some years I used a three-position switch-tuned v.h.f. tuner which, in combination with automatic frequency control, proved extremely easy and reliable to use; however when LBC and Capital Radio opened in the London area I was forced to modify the unit so that it now includes a manual tuning knob.
Simple to operate is not necessarily the same thing as electronically
simple, of course. Indeed some of the latest tuning systems on the new, better-class of shortwave receivers are often electronically complex but, almost for the first time, make it reasonably simple for the owner to tune to a required station without having "safe-cracker's fingers"though the constantly changing conditions of h.f. will never make shortwave broadcast reception (unless we eventually have 26 MHz satellite transmissions) really granny-proof.

## Saturday's private army

The dreadful news of the brutal death of Airey Neave, M.P., in the House of Commons car-bomb explosion brought back to me memories of the month I spent as an "unofficial" radio operator for his Escape-line Unit (IS9) in Holland in 1944.

At the end of October, that grim winter, I had come with an MI6 officer to Nijmegan where there existed, unknown to the enemy, a private powersupply telephone network over which it was possible to talk directly to Dutch Resistance groups in the occupied north of the country. But hardly had I set up my radio station in a house on the outskirts of the town when Major Fraser (Hugh Fraser, M.P., one-time SAS and Phantom officer) then second-in-command to Airey Neave ("Saturday" of M19) arrived next door with his own signals operator. Foreseeing mutual interference, I attempted to argue prior rights but my claims were cheerfully swept aside by the simple expedient of IS9 moving into both houses and taking me over in a manner, I soon realised, that had characterised the Escape Unit's roystering campaign across France, Belgium and Holland.

It had certainly been an imaginative War Office appointment to put Airey Neave, the first British officer to escape successfully from Colditz, in charge of IS9 whose official task was to recover "evaders" (aircrew and others hidden in occupied territory) but it was asking a lot of an essentially gentle character, no matter how resourceful, to keep fully under his control the cheerful tearaways and tough romantics who made up his small "private army", which had by then expanded to include sundry Canadians, Belgian SAS parachutists, Belgian and Dutch resistance workers.

They came to Nijmegan fresh from the first highly successful "Pegasus"
operation which brought back some 140 Arnhem survivors, but still embarrassed by the loss of Peter Baker ("Harrier", later to be first an M.P. and then a guest of HM Prisons) who had gone bravely but foolishly and against orders across the rivers into occupied territory. IS9 began planning a "Pegasus 2" operation but from the first such plans seemed doomed. Baker had been in close touch with "King Kong" (the Dutch resistance leader who had been turned by the Germans) and there had also been an ill-advised press conference at which one of the senior officers brought out in Pegasus had spoken openly of the many Arnhem evaders still hidden on the wrong side of the rivers, Maas and Waal.
However, Pegasus 2 was helped by unusually good communications: apart from the power-supply telephone and 450 MHz "S-phones" there were clandestine h.f. links via London and Eindhoven, including those of the Dutch internal radio service of the RVV and OD resistance groups. After several weeks of preparation, the operation proved, as Airey Neave described it in his book "Saturday at M19", a grim disappointment. Many of the evaders were killed or captured in a carefully prepared ambush and it was clear that the Germans knew in advance what was afoot.
But although Pegasus 2, the last major evader operation mounted by IS9, was a tragedy, it should not be forgotten that the exploits of the escape-lines of M19 as a whole were remarkably successful, with several thousands brought out of occupied Europe.

Shortly afterwards I left Nijmegan and the private army of IS9 to pursue a rather different aspect of the secret radio war that continued until May 1945, taking as a souvenir of those interesting weeks one of 'S9's silk escape maps that is on my wall now as I write this small tribute to the man who led it.

## New Contacts

Those short-wave enthusiasts and radio amateurs who assume that reflection from the F2 layer of the ionosphere always fades out above about 30 MHz have been caught out already in this high maximum sunspot cycle.

Apart from reports of the reception of British television signals in the United States, many long-distance signals have been heard in Europe on the 50 MHz amateur band (not normally available to amateurs in the UK). Signals from North America, South America and southern Africa have all been heard at good strength.

A new world record has been set up with a two -way 50 MHz contact between amateur stations in Argentina and Korea.

# Everyday News 

## MADLEY-I

With phone calls to and from Britain doubling every four to five years, and telex and computer data growing just as fast, the strain on the Post Office resources increases each year.

Currently running at 12 million a month, of which 4 million are accounted for by Intercontinental calls and six out of every ten go by satellite.

To help cater for this huge growth in overseas phone calls, he Post Office has just brought a new satellite communications station, Madley 1, into operation at Madley near


With a diameter of 32 m (105ft), this dish aerial is the biggest so far built for the Post Office, and will eventually be carrying up to 2,000 phone, telex or computer data calls, as well as TV transmissions, between Britain and East Africa, the Middle and Far East, India, Australia and New Zealand. Its powerful microwave radio signal, operating in the $4 / 6 \mathrm{GHz}$ band, is beamed to an Intelsat IVa satellite positioned 22,500 miles out in space over the Indian Ocean.
The site at Madley was selected, after a three-year search, for several reasons, namely: an "electrically quiet" area free from radio interference, particularly from the inland microwave network which uses the same frequencies. Plenty of room for expansion, altogether six such aerial complexes are planned. And, finally, easy access to good telecommunication links with London.

Work first began in autumn 1976 and it was built for the Post Office under a contract in which Marconi Communications Systems Ltd., was prime contractor, supplying the radio and communications equipment and supervised the main sub-contractors IDC Ltd. (buildings) and the Mitsubishi Electric Corporation (aerial dish). As
prime contractor, Marconi Communication Systems also co-ordinated the efforts of an international team of subtractors including Mitsubishi Electric Corporation, Japan, for the antenna sub-system and Comtech, in the USA for the low-noise amplifiers.
It is foreseen that Madley 1 will be used with the next generation of international telecommunication satellites, Intelsat $V$, where the system channel capacity will be doubled. In terms of quantity of equipment, Madley 1 , with its 32 -metre antenna, is one of the largest satellite earth stations operating in the Intelsat system. In all, 55 chains of receiving equipment, 14 chains of transmitting equipment and 10 high power amplifiers give Madley 1 the capability to communicate with about 40 countries simultaneously and Marconi is already manufacturing equipment to extend this capacity.
So far, the Post Office has spent about $£ 10$ million at Madley, under its $£ 1,000$ mil-lion-a-year programme of investment in new plant needed to cater for growth and provide for new services in the future. Of this Marconi's share has been $£ 6$ million and IDC of Stratford on Avon share $£ 1$ million.

## NEW LAUNCH

The latest in the UK/Ariel series of scientific satellites, UK6, is scheduled for launch on Ascension Day, May 24, from Wallops Island, Virginia.

Seventy radio and 18 TV studios will be used at the Moscow Olympic Games in 1980. They will need 135 mixing consoles and to achieve all the programme, switching and talkback lines a total of 500 full-sized 19 -inch racks of equipment will be needed.
Among the other equipment needed are 576 monitoring loudspeaker systems and 77 technical checking and control desks.

## Shetland growth

New primary and secondary radars are being supplied to Sumburgh Airport in the Shetland Isles by Marconi and Cossor as part of a $£ 30$ million airport development programme.

North Sea oil has made Sumburgh Britain's fastest growing airport. Last year it averaged nearly 1,000 aircraft movements a week and handled nearly 700,000 passengers.

It will carry cosmic ray and X-ray experiments designed by Leicester and Birmingham Universities, and University College London, to extend knowledge of high-energy astrophysics. The Royal Aircraft Establishment also has two experiments in the payload in the field of solar and CMOS arrays.

The project is funded by the Science Research Council and cost $£ 4 \cdot 0$ million with Marconi Space and Defence Systems as main contractor for the spacecraft.

## China Trade

Trading with China, now the "in" thing, is nothing new to SE Labs (EMII) Ltd. Recent orders worth $£ 500,000$ for transducers and recording equipment are only a followon from trading links first established some 15 years ago.

## ANALYSIS

## DOING TRICKS WITH PACKETS

During the next few years we can expect to hear quite a lot about packet switching．The idea is to load international communications networks far more effectively and thus help to contain the enormous increase in traffic already forecast without a proportionate increase in the number of com－ munications channels
One way of obtaining higher utilisation is to increase the speed of transmission so that for any given length of time a longer message can be transmitted or，conversely，for a given message length the transmission time is reduced．This is already happening but once a channel is＂seized＂it is occupied by the subscriber for the duration of his call whether or not a message is actually being transmitted．And if there are pauses，then the time taken for pauses is wasted． But if the channel is shared by a number of subscribers or users and they are sending and receiving at comparatively low data rates on their local lines，all their messages can be interleaved at a high data rate on the trunk channel and it can be used continually by all subscribers rather than inter－ mittently by single subscribers．
One way of doing this is to break up each message into small＂packets＂of information at local packet switching exchanges and squirt the packets down the high speed link．

The packets are then re－assembled into the complete message at the remote exchange and passed to the recipient． Although the user has lost exclusive use of a line because other peoples packets are also being transmitted，it all happens so quickly that the user appears to have a contin－ uous and exclusive path．

You can see the snag right away．Imagine yourself at the receiving exchange of the high speed trunk channel with all these bits of messages coming in a constant stream，none of them as complete messages and for different destinations． Happily electronics comes to the rescue in the form of the mini and micro computer．Each packet of message is tagged with destination and with error protection codes and the sorting and switching is all automatic．So is automatic routing so that if long－distance satellite links are full your message can go by submarine cable or vice versa．Conversion of local low－speed data to the high－speed long－distance links and reconversion at the remote end is also automatic．

Packet switching is already in use on some private data networks．With the huge growth in data traffic it must come universally on the public networks and electronics will make it possible．The British Post Office has just started a packet switching service to the USA and will start extending the system to internal use next year．

Brian G．Peck

## New Standards

The first part of a major revision of BS3549＂Methods of Measuring and Expressing the Performance of Tele－ vision Receivers＂has just been published by the British Standards Institute．
Of particular interest to television manufacturers and service engineers，it is being extended to include appro－ priate requirements for colour television．

Part one is entitled ＂General Considerations and Electrical Measurements Other Than Those at Audio Frequencies＂and copies can
be obtained from BSI Sales Department，price $£ 12 \cdot 20$ ．The complete works will consist of four parts and will be issued at a later date．

## 瓷 资 资

A solid－state matrix camera for industrial and commercial use has been developed by Reticon Corp．It has 10,000 photodiodes on a 6 mm square silicon chip．It can be used for generating a TV picture or for applications like optical character recognition and pattern and shape recog－ nition．UK agent is Herbert Sigma，Letchworth，Herts．


#### Abstract

\section*{Take－overs}

Taking over the former Tandberg colour TV factory in Scotland is Mitsubishi．The Japanese audio company Aiwa is reported as being in negotiation for a plant in Ulster． Japanese electronics com－ panies already manufactur－ ing in the UK as Sony， National Panasonic（Matsu－ shita），Hitachi（with GEC） and Toshiba（with Rank），the last two being joint ventures．


The latest entry into the great large scale memory race is ITT Semiconductors． Their Footscray，Kent，divi－ sion is to produce a 64k random access memory （RAM）with samples avail－ able by mid－year and produc－ tion by the end of the year．
At present they manufac－ ture 4 k and 16 k memories and the 64 k device is part of a $£ 10$ million development programme，of which $£ 2$ million is being funded from the Department of Industry．

## PRESTEL GOES PUBLIC

The Post Office Prestel viewdata system（available only in the London area），got away to a＂low－key＂ start with only 400 or so private subscribers due，it was said，to non－availability of suitably adapted TV sets．Rapid expansion of the service is，however， expected with a hoped－for million customers in the UK after five years．

Later this year the public service will become avail－ able to business users as well，and will be extended to Birmingham，Manchester and Edinburgh．
Prestel is based on the world－leading viewdata system invented at the Post Office＇s Research Centre． It puts，it is claimed，Britain years ahead of any other country in the mass marketing of electronic information．

## Overseas Sales

The Post Office has ex－ ploited this world lead by selling the technology to West Germany，The Nether－ lands，Hong Kong and the USA，and is currently en－ gaged in sales discussion with many other countries．

The latest country to sign an agreement to purchase Prestel software is Switzer－ land．This takes the form of programmes，or instructions， which control the operation
of the computer on which the information is stored．

Like those with Germany and Holland，the Swiss deal is for the＂pilot trial＂soft－ ware，designed primarily for demonstration and informa－ tion development．Because Switzerland is multilingual， all the index pages will be translated into French and German．
The Post Office budget for Prestel this year is $£ 30$ mil－ lion and the technology is selling world－wide．



0NCE one "goes into" electronics, and friends and relatives discover one's new interest, the well-meaning darlings start to off-load all manner of old electrical fittings and superheterodyne white elephants. We smile and thank them politely, dumping ninety per cent of it straight in the dustbin, but we do not discourage them, for the odd ten per cent of their junk may prove of value.
Such was the case when I took delivery of a car boot-load of old 15A ceramic wall sockets and plugs, from a relative whose house had recently been rewired. Having decided that 15 A plugs do not marry easily with $0 \cdot 1$ inch Veroboard transistor and i.c. layouts, the screws were removed, and the rest dumped.
Amongst this stuff was also a few yards of heavy duty wiring cable, which was kept, for some unthinkable contingency. This cable, some 15 mm wide, and 8 mm thick, and covered in heavy grey plastic sleeving, has made a fine headband for the home-made stereo headphones which I describe this month.

The Stereo Headphones described here can be constructed with items that will be found in many a junk box, yet the prototype was found, after many comparisons, to be capable of better sound reproduction than some commercial phones.
First, acquire two plastic measuring cups, as found in tins of Smash instant potatoes. Purchase two 76 mm plastic flowerpot saucers. After drilling a $\frac{3}{8}$ inch hole through the centre of each flowerpot saucer, you will find that a 57 mm diameter loudspeaker will fit snugly between the flanges on the underside of the saucer, and that the Smash measuring cup will fit over the two, with an amount of swivel adjustment available from the "ball-and cup" movement.
 nominal impedance of 75 ohms, which allows them to be used of course with any output up to that impedance, and the higher rating means that they are slightly reduced in sensitivity, which is an asset particularly when the phones are to be connected to the loudspeaker outputs of equipment that does not boast an earphones socket.

Obtain a length of heavy duty electrical cable (as referred to above). About 460 mm should be sufficient.

The beauty of a headband made from such cable is that it is suitably bendable, already covered, and has insulated conductors through the . middle. With a little manoeuvring, the copper conductors can be cut and worked back into the outer sleeving, after soldering a length of single- or multi-stranded wire of more manageable diameter, to the ends of the two outer conductors. Drill a $\frac{3}{16}$ inch hole through the centre of the cable, about 20 mm from each end, make a similar diameter hole through the centres of the Smash measuring cups, and fix with suitable bolts, washers and nuts.

Now fix the speakers to the flowerpot saucers, a suitable adhesive being UHU, or aircraft glue. After making two tiny holes in suitable places in each cup, thread the small wires through, and allowing the minimum working length, cut, bare and solder to the speaker at one end, soldering the wires to the speaker at the other end.

A suitable hole is made in the second cup also for the supply lead, which must have three cores -low-current mains cable would suffice, although the writer used four-core GPO cable, one core being cut back, and not used.

Having bent the headband into shape, and checked that all is working OK, the saucer baffles and cup enclosures are now adjusted to suit the wearer, and then glued together.

After fixing a stereo jack plug to the other end of the supply flex, the headphones are ready to be used. Some sort of lambswool, chamois leather or plastic edging could be affixed to the edges of the plastic saucers, but the bare rims have proved to be very comfortable when worn for long periods of recording, and certainly less "sweaty" than conventional padded plastic muffs previously endured.

The omission of such material probably also improved the bass response, through direct contact with the bone area around the ears. In tests, there seemed to be response even when a 100 Hz Baxandall bass circuit was adjusted.

Such a claim would appear to be unlikely, and the only explanation seems to be that there is a direct transmission of bass notes through the bone. The 76 mm flowerpot saucer does however allow a greater area for resonance than conventional headphones, and this may also contribute to the increased sound spectrum evident.

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I:T is virtually impossible today to build a pocket radio which can compare in price and performance with the mass produced Japanese radios which are currently on the market. To build a radio of similar performance would involve a great deal of time and would probably cost much more than if you had purchased one in the first place.

However, the cost is not the most important factor in this hobby, it is the experience and knowledge gained and the excitement of putting together electronic components to produce, in this case, a source of entertainment and information.
This receiver is not intended to compete with the far superior superheterodyne type of radio, but was designed with simplicity in mind and the use of a p.c.b. makes it a suitable project for the novice constructor.

## CIRCUIT DESCRIPTION

Two integrated circuits are used, one is the ZN414 t.r.f. radio i.c., the other an LM380 small power amplifier i.c. The circuit diagram of the radio is shown in Fig. 1.

The ZN414 is in itself a complete a.m. receiver (less aerial and tuning capacitor) and as such could be used on its own without an amplifier if only headphone reception is required.

This i.c. requires a supply of between 1.1 and 1.5 volts, and unless this is reasonably stable, whistles and distortion will result. Perhaps the easiest method of obtaining this voltage is the system shown in the circuit. Here two diodes are connected in series across the supply voltage.

In a silicon diode, the junction does not conduct until the potential across it reaches a certain minimum, about 0.6 V . When the diode does conduct the voltage across it remains steady at this voltage. Thus with two diodes connected in series we obtain 1.2 V , which is within the limits of the i.c. Resistor R3 is a current limiting resistor for the diodes.

## TUNED CIRCUIT

The tuned circuit, consisting of L1 and C2 selects the required frequency. After amplification and detection within the i.c. an audio output is available at pin 3, and thus across the volume control VR1.

Resistor R1 provides feedback between the output of IC1 and its input. Resistor R2 together with the supply voltage sets the overall gain, and in most cases will be sufficient. However there may be times where more sensitivity is required, if this is so then the value may be adjusted to suit.

However, the value is rather critical, and if a value very different from that specified is used, it may result in instability.
With the given values for L1 and C1, a high $Q$ tuned circuit is produced resulting in good selectivity. The values of both coil and capacitor may be altered if a tuning range other than medium waves is required. It is possible by careful changes in values to tune from the long wave band up to the 160 m Amateur band.

## AUDIO AMPLIFIER

As the output from the ZN414 is unsuitable for direct connection to a loudspeaker an amplification stage is incorporated in the form of IC2, a general purpose audio i.c. type LM380. This i.c. has a fixed gain of 50 and can deliver up to 2 watts.
The only component which needs to be mentioned is C5. This decouples the power supply and eliminates instability. It is important that this is mounted close to the i.c.


All the wiring details are shown in Fig. 2. A printed circuit board layout is shown and in the interests of stability this should be used. It is possible to use some other form of wiring, such as stripboard but this is not really to be

Fig. 1. The complete diagram of the Pocket Radio.



## COMPONENTS

Resistors
R1 $100 \mathrm{k} \Omega$
R2 $100 \Omega$
R3 $15 \mathrm{k} \Omega$
All $\frac{1}{4}$ W carbon $\pm 5 \%$

## Potentiometer

VR1/S1 5 kilohm log. with ganged switch.

Capacitors
C1 $0.01 \mu \mathrm{~F}$ polyester
C2 126pF nominal, plastic dielectric tuning capacitor
C3 $0.22 \mu \mathrm{~F}$ polyester
C4 $10 \mu \mathrm{~F} 10 \mathrm{~V}$ elect.
C5 $0.1 \mu \mathrm{~F}$ polyester
C6 $4 \cdot 7 \mu \mathrm{~F} 10 \mathrm{~V}$ elect.
${ }^{,} \mathrm{C} 7100 \mu \mathrm{~F} 10 \mathrm{~V}$ elect

## Semiconductors

IC1 ZN414 t.r.f. radio i.c.
IC2 LM380 audio amplifier
D1, D2 1 N4002 silicon diode (2 off)

Miscellaneous
L1 80 turns 28 s.w.g. enam. copper wire (see text)
LS1 $8 \Omega 60 \mathrm{~mm}$ diameter loudspeaker
B1 9V PP3 battery
Plastic case type PB1 or similar, minimum size $115 \times 75 \times$ 40 mm ; printed circuit board; ferrite rod, $10 \times 75 \mathrm{~mm} ; 28 \mathrm{~s}$. w.g. enamelled copper wire; batery clip to suit B 1 ; one 14 -pin i.c. socket; small round control knob; tuning knob 40 mm diameter; connecting wire.

Fig. 2. Complete wiring details for the radio. At top left is shown the topside of the p.c.b. and its position in the case. Note also the positions of the battery and coil. It is important that they are placed in these positions otherwise the lid will not fit in place. Next comes the rear of the front panel showing the connections required to the various controls. Finally at bottom is the underside of the printed circuit board, which is shown full size.

recommended. Incidentally for those who have never made a p.c.b., this is an ideal project to start with as it is very simple to build.
All interconnecting wires should be as short as possible, particularly the leads from the coil to the tuning capacitor, and from the capacitor to the board.

If the beginner is unsure about his ability to solder, particularly IC1, it is recommended that a three-pin transistor socket is used.

## CASE

The prototype was built in a commercial plastic case, type PB1. There is of course no reason why a different case already to hand cannot be used, providing it is not metal and has minimum dimensions of $115 \times 75 \times 38 \mathrm{~mm}$. No drilling details are given as they and the size of the case will depend to a large extent on the size of components used.

The p.c.b. was mounted on the bottom of the case using two 6BA screws and nuts. The battery can be held in place by double-sided adhesive tape or Blu-Tak.


Components mounted on the printed circuit board.

## COIL

The coil, L1, is a home-made component and is constructed by winding 80 turns of 28 s.w.g. enamelled copper wire, close wound on a paper former, which is then slid over a 10 mm ( $3_{8}$ inch) ferrite rod. The length is unimportant providing it is at least 75 mm long and can also fit in the case.

The tuning capacitor C2 was salvaged from an old miniature radio, but a readily available type can be obtained from a number of suppliers. It is not essential to keep to the value specified as the tuning range is quite wide. In-
 deed, it may be an advantage to use an old capacitor already to hand, and change the number of turns on the coil, particularly if a different tuning range is envisaged.

## SPEAKER

The speaker was attached to the front panel using two small clamps and countersunk 6BA screws. Alternatively the speaker can be stuck down using an impact adhesive if it is desired not to mar the front panel.

Letraset or similar can then be used to label the front panel as required.


## IN USE

After construction has been completed, and the board and other wiring checked for errors the battery may be connected.

Connect the battery and switch on, rotate the volume control until a "hissing" sound is heard-this is white noise being picked up by the radio and suggests that it is working correctly.

Rotate the knob of C2, whereupon a number of stations should become audible. If none are heard then this is probably due to the coil being in the wrong position on the ferrite rod. Sliding the coil along the rod while at the same time adjusting C2, you should be able to find a position where a number of stations come in at roughly the same level. If all is well the coil can be stuck in position. Blu-Tak or similar is suitable.

The receiver gives good reception of local stations, and after dark many overseas stations can be picked up. The number and their level depends to a large extent on your location and of course the time of day.

Although the ZN414 is not being utilised to its full capacity, it nevertheless gives quite good results despite its simplicity. I

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TRUE QUADRAPHONIC systems often use complex electronic circuitry to achieve the end result, which to many people is basically a surround-sound type of effect.

The original concept has probably been largely forgotten due to the often fanciful "gimmicky quad" type of records. It would seem therefore perfectly acceptable to relate the "surround-sound" type of effect to true Quad, although one must make the statement that the two are completely different but in most instances the effect is hardly noticeable.

This has in fact been proved by the prototype which has been in use by the author now for several months, and on many occasions has fooled a few "hi fi knowalls" in believing that they are listening to true quadraphonic.

The purpose of this article is to describe an add-on unit, which when connected to your existing stereo equipment will give quite a good quadraphonic effect. The difference in signals between the two rear speakers is sufficient to be detected, the resulting overall
sound akin to that of "surroundsound". In fact the more the stereo separation in the original recording, the greater and more dramatic the quadraphonic effect will be.

## PRINCIPLES

The basis of the Quadraphonic Simulator is a modification of the Hafler configuration.


Fig. 1. Hafler's basic idea of adding a single speaker.

Hafler demonstrated, that by adding a single speaker behind the listener and using the difference signals between left and right channels, interesting quadraphonic effects could be achieved, Fig. 1. However by using two speakers out of phase with each other a more realistic effect can be achieved, Fig. 2.

There are at least two disadvantages to the arrangement of Fig. 2.


Fig. 2. By using two speakers, a more dramatic effect is obtained.


Fig. 3. Circuit diagram of the Quad Simulator. This is suitable for systems up to 10 W r.m.s, output.


Fig. 4. Complete wiring details for the unit. Note the case has been "opened out" for clarity. There is nothing critical about the layout and could be varied as required.

## QuaO FlyUGator



## COMPONENTS

Resistors
R1 $10 \Omega 5 \mathrm{~W}$
R2 $22 \Omega 3 \mathrm{~W}$
R3 $22 \Omega 3 \mathrm{~W}$
All are wirewound $\pm 5 \%$

## Potentiometers

VR1 100 3 W wirewound

## Capacitors

C1 $220 \mu \mathrm{~F} 16 \mathrm{~V}$ elect.
C2 $220 \mu \mathrm{~F} 16 \mathrm{~V}$ elect.

## Miscellaneous

PL1, 2 2 pin DIN plug (2 off)
SK1.4 2 pin chassis mounting DIN socket (4 off)
SK5 standard insulated stereo jack socket
Aluminium or plastic case to choice, approximate dimensions: $152 \times 102 \times 57 \mathrm{~mm}$; small round plastic knob; length of twin screened cable; nine way tag strip with two tags earthed; connecting wire.


One being that the connection of the extra speakers tends to reduce the separation between the main channels. Secondly, the volume of the rear speakers can only be adjusted by adding low value resistors in the main speaker circuit: this may be inconvenient and could cause a mismatch.
Despite these disadvantages the circuit here seems to function quite well.

## CIRCUIT DESCRIPTION

The complete circuit diagram for the Quadraphonic Simulator is shown in Fig. 3.

The first point to notice about the circuit is the similarity between it and the Hafler configuration. The one major difference being the inclusion of R1. This is basically included to reduce the level of signal applied to the rear speakers.
Remember of course that not a great deal of volume is needed in the rear speakers, thus the reduction in volume is hardly noticeable.

## HEADPHONE OUTPUT

An addition to the basic circuit is the provision of a headphone output, this part of the circuit consists of R2, R3, VR1 and the output socket, SK5. Here the two
fixed resistors set the basic limit on the volume through the headphones.

Since you are mixing both front and back signals, the effect in the headphones is not the same as with speakers, nevertheless, the effect is still quite original. From this point of view, the variable resistor VR1 is included to be able to mix the two signals in varying degrees, to produce the most pleasing effect.


The unit is quite easy to build, and could if desired be incorporated inside your hi fi system. This would mean a fair amount of change in the hi fi set-up so is not recommended for the beginner.

An option is available whereby the headphone output components could be omitted. If this is done, wiring it into the hi fi system would be that much easier. The option also exists if the unit is to be as originally intended, and that is as an outboard item.

All the wiring details are shown in Fig. 4. There is nothing critical about the layout and could be varied if required. A tag strip is used as a central connection point for most of the wiring. Note that although twin screened cable is shown for the input connection, this is by no means essential and ordinary four core cable can be used.

The choice of case is also up to the constructor, either a metal or plastic type can be used, and it is not essential to keep to the dimensions specified. Thus any case which the constructor has to hand can be used.

When using a metal case it is important to use insulated sockets, particularly for SK5. If this is not observed then a situation will occur where the jack plug could possibly come into contact with earth, the resulting large increase in volume doing something terrible to the listeners nerves!

## IN USE

Once constructed and checked for any mistakes the unit can then be put to use.

Simply disconnect the original loudspeakers from the hi fi and plug them into the unit sockets marked left front and Right front. Connect an additional pair of speakers to the remaining two sockets, and connect the two leads into the speaker output sockets of the hi fi. When completed the unit is ready for use.

As stated earlier, VR1 is used to give a reasonable balance when using headphones.
The Quad Simulator has now been used by the author for some time, and has been found of sufficiently high quality to solicit several requests for the cricuit and other details.

It is for this reason that the design has been presented here with the hope that many who build the simulator will have many hours of pleasure rediscovering old stereo recordings.


## ШORHSHOP

# matters 

By Harry T. Kitchen

## Organisation

The pleasure that is derived from the construction of electronic equip. ment is very largely governed by the quality of that construction, and that in turn can be governed by the facilities available. My earliest forays in this direction were in a tiny bedroom that constituted my "digs", and this in turn required a very surreptitious manner of working. I now have the luxury of a room devoted to photography and electronics. In between I have made do in all sorts of ways. The advice that follows is therefore exceedingly practical.

Firstly, always endeavour to avoid falling foul of the authorities, be they wife, mother, landlady, or who ever. This is plain commonsense, since falling foul of them may mean that your activities may be curtailed; terminated even. There is a more basic reason than this, and this is good old fashioned courtesy. Let courtesy be your key word, and if anything fails, at least you will have a clear conscience.

## Component Storage

Assume you have been granted certain facilities. Off you go. Then comes the time to clear all up after you. Where do you store everything? The less storage space, the greater the need for a disciplined mind. You can stuff everything into cardboard boxes, and of course they will suffice. Until they disintegrate. Jam jars are more durable if handled with care, and the contents can be reasonably easily identified. Better, but not ideal.
I have yet to find anything better, for the storage of most electronics components, than the plastic drawers that slot into one another, both vertically and horizontally. They are available in several sizes, all compatible. Labels can be inserted into the fronts of the drawers aiding identiflcation of the contents; the transparent fronts are an aid here. I have a large number of these drawers, bought over a period of several years. These house all my resistors, capacitors, transistors, diodes, as well as the veritable multitude of "small things" that are such an adjunct to our hobby.
The resistors and capacitors are housed in decade increments. Resistors of differing wattages or tolerances are segregated. Similarly, capacitors are segregated into types; electrolytic,
polycarbonate, polystyrene. Transistors are segregated into types: germanium or silicon; r.f. or a.f.; high power or low power.
Rigid segregation is the key to efflciency and speed, and whilst it may sound profligate, remember that you can buy drawers as you require them; I bought mine in lots of two dozen. I have never regretted doing so.

## Tool Care

Tools cost money. In the interests of reliability and longevity, they should be treated with care and not chucked carelessly into a pile. Cardboard boxes can be pressed into use, within the foregoing proviso. Tool rolls can be made or bought, and are to be preferred.

If your tool collection is at all extensive, a proper tool box is well worth while. I have several "engineers" tool boxes, bought when they were around $£ 4$ each. They very probably cost double that now, but they are to be recommended. A useful compromise is the partitioned drawer sold for the storage of cutlery.

When you have completed a constructional session, particularly on "borrowed" premises, i.e. the kitchen table (beware, I have a patent on 'eml), always clear up properly after you. it takes very little time, but it can go a very long way indeed towards the maintenance of cordial relations all round.

## Buying Components

We have so far considered the organisation of the home workshop as though it were fait accompli. This it will not always be. There will be the occasional constructor who will buy his components as he requires them. No storage problem for him. Others, like myself, will like to keep a large variety of components in stock so that almost any project can be commenced without the infuriating delay of obtaining components to suit.
My decade storage may be a clue to my buying philosophy. Yes, I buy my resistors and capacitors in decades, in every sense of the word. Several years ago I saw an advertisement for 5 per cent $\frac{1}{2} W$ carbon film resistors at ridiculous prices. I bought a few to check. They were eminently satisfactory, so in I plunged.
I bought 10 off every decade in the E12 series from $10 \Omega$ to $10 \mathrm{M} \Omega$. A lot of
resistors indeed, and they have been very well worth while. At a later date I added 5 per cent $\frac{1}{6}$ W carbon film resistors, and a variety of wire wound types. Then came a consignment of "computer stock" 2 per cent $\frac{1}{2} W$ metal film.
Similarly with capacitors. I bought 10 off every decade in the Mullard C280E series, plus a wide variety of electrolytics and polystyrenes. The electrolytics are axial-ended, and I set an upper limit of 16 volts working as I did not envisage frequent use of higher voltages. Besides, my money had run outl

## Transistors

The transistor problems were solved by buying a selection of the most commonly used types, a.f. and r.f., low and high power. Some complementary types were also bought. For diodes, I decided to standardise on the 1 N4004 and 1 N914 types, and these fulfil the vast majority of existing equipments. These semiconductors augmented an existing stock, and not for worlds would I part with my "red spot", "white spot" or "top hat" germanium transistors, or the equally venerable GEX34 series of diodes, plus the odd GET1 transistor. They're almost museum pieces I
My philosophy is designed primarily for convenience, and if you agree with it, fine. If not, modify it to suit yourself. Many firms now advertise resistors and capacitors at very low prices, and give a small discount if 10 or more of each value are bought.

The same sometimes applies to semiconductors. I have never, ever, bought dubious ("unmarked", "remarked", or whatever) semiconductors. Use such components, and what do you do if your equipment fails to function? You could end up mystified and miserable, all for the sake of a few coppers. Is it worth it?

"Make me happy, darling-give me the run of your electronic components shop."

## BOOKS AND COMPONENTS

## BOOKS BY BABANI

## V.A.T. Zero Rated

Beginners Guide to IC's
Engineers \& Machinists Ref. Tables
BP6 BP14-
$\mathrm{BP22}$
BP24
BP27
BP32
Handbook of IC Audio Preamplifler a Power Amplifler Construction 50 Cicts use Germ/SiI/Zener Diodes 50 Projects Using Relays/SCR/Triacs 50 Field Effect Trans Prolects Digital IC Equivs \& Pin Connection Linear IC Equivs a Pin Connection 50 Simple LED CIrcuits How to make Walkie-Talkies IC 555 Timer Projects
Projects on Opto-electronics Radio Circuits Using IC's Mobile Discotheque Handbook Electronics Prolects for Beginners Popular Electronic Projects IC LM3900 Projects
Radio Stations Guide Coil Design \& Construction Manual Handbook of Integrated Circuit 1st Book Hi-Fi Speaker Enclos 1st Book Hi-Fi Speaker Enclosures Circults for Model Rallways Shortwave Circuits \& Gear for
Experimenters \& Radlo Hams Experimenters \& Radlo Hams Electronic Gadgets \& Games BP217 Solid State Power Supply Handbook BP221 28 Tested Transistor Projects BP222 Short-wave Receivers for Beginners BP223 50 Prolects using iC CA3130 BP224 50 CMOS IC Prolects
BP225 A Practical Intro to Digital IC's Build Advanced Short-wave Receivers Beginners Gulde to Building Electronic Projects

## NEWNES BOOKS

## V.A.T. Zero Rated

Order No. Description QUESTIONS \& ANSWERS:
216 Transistors 3rd Ed.
217 Integrated Circuits
218 Radio \& Television
219 Electronics
220 Colour TV 2nd Ed.
$221 \mathrm{HI}-\mathrm{Fi}$
22220 Solid State Proj, for Car
22320 Solld State Prol. for Home
224110 Int. Circ. Proj, for Home
225110 Thrysistor Projects
226 Operational Amp. Proj. for Home
227110 Practical IC Proj, for Home
228 Electricity
229 Beginners Gulde to Electronics
230 Beginners Guide to Television
231 Beginners Guide to Transistors
232 Beginners Guide to Electric Wiring
233 Beginners Guide to Radio
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235 Electronic Diagrams
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240 Somiconductor Handbook Part 1 241 Semiconductor Handbook Part 2 242 Electronics Pocket Book
243 Radio Value \& Somiconductor Data 244 Beginners Guide to Integrated Circuits

209 BI-PAK TTL Data Book
BI-PAK CMOS Data Book

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## SWITCHES

| Description | No. |  | Price |
| :---: | :---: | :---: | :---: |
| DPDT miniature slide | 1973 |  | $14 p^{*}$ |
| DPDT standard silde | 1974 |  | 15p* |
| Toggle switch SPST | 1975 |  |  |
| 11 amp 250 V a,c. |  |  | $33 \mathrm{p}^{*}$ |
| Toggle switch DPDT | 1976 |  |  |
| 1 amp 250 V a.c. |  |  | 420. |
| Rotary on-off mains switch | 1977 |  | 50p. |
| Push switch-Push to make | 1978 |  | 14p* |
| Push switch-Push to break | 1979 |  | 18p ${ }^{*}$ |
| ROCKER SWITCH COI | Colour | No. | Price |
| A range of rocker | RED | 1980 | 30p* |
| switches SPST-moulded BLACK 1981 |  |  | 30p* |
| in high insulation. | WHITE | 1882 | 30 p , |
|  | BLUE 1983 |  | $30{ }^{\text {p }}$ |
| choice of colours ideal | YELLOW 1984 |  | 30 p * |
| for small apparatus. | LUMINOUS | 1985 | 30 p * |
| Description <br> Miniature SPST toggle, 2 amp |  |  | Price |
|  |  |  |  |
| 250 V a.c. | 1958 |  | 70p* |
| Miniature SPST toggle, 2 amp |  |  |  |
| 250 V a.c. | 1959 |  | 75p ${ }^{\text {* }}$ |
| Miniature DPDT toggle, 2 amp | 1960 |  | 80D* |
| Minlature DPDT toggle, centre | 1961 |  |  |
| off, 2 amp 250 V a.c. |  |  | 95p* |
| Push button SPST, 2 amp 250 V a.c. | 1962 |  | 90p* |
| Push button SPST, 2 amp | 1963 |  |  |
| Push button DPDT, 2 amp |  |  | p* |
| 250 V a.c. | 1964 |  | £1-20 |
| MIDGET WAFER SWITCHES |  |  |  |
| Single-bank wafer type-sulfable for switching at 250 V a.c. 100 mA or 150 V d.c. in non-reactive loads make-before-break contacts. These switches have a spindle $0.25 \mathrm{i}^{\mathrm{n}} \mathrm{dia}$. and $30^{\circ}$ indexing. |  |  |  |
| Description | Order No. |  | Price |
| 1 pole 12 way |  |  | $48 p^{\circ}$ |
| 2 pole 6 way |  |  | $4{ }^{\text {ct }}$ p |
| 3 pole 4 way | 1967 |  | 48p |
| 4 pole 3 way |  |  | 48p* |
| MICRO SWITCHES | Order No. |  | Price |
| Piastic button gives simple 1 pole change over action |  |  |  |
| Rating 10 amp 250 V a.c. |  |  | 25 |

FUSE HOLDERS AND FUSES
Description
$20 \mathrm{~mm} \times 5 \mathrm{~mm}$ chassis mounting $1 \operatorname{in} x$
IIn
IIn
In chassis mounting 1in car inline type Panel mounting 20 mm Panel mounting $\frac{11}{2}$ in


## NUTS AND BOLTS

BA BOLTS-packs of BA threaded cadmium plated screws siotted cheese head.
Supplied in multiples of

| Type | No. | Price | Type | No. | Price |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 1 in OBA | 839 | 21-20 | tin 4BA | 846 | 32p |
| $\frac{1}{2}$ in OBA | 840 | 75 p | \%in 4BA | 847 | 25p |
| $\operatorname{lin} 28 \mathrm{~A}$ | 842 | 65p | IIn 8BA | 848 | 40p |
| In 2BA | 843 | 45p | y 3 In 6BA | 849 | $21 p$ |
| ilin 2BA | 844 | 52p | tin 6BA | 850 | 25p |
| 1 in 4BA | 845 | 46 p |  |  |  |



\footnotetext{
BA WASHERS-fiat cadmium plated plain stamped washers supplied in multiples of 50 .

| $\begin{aligned} & \text { Type } \\ & 0 B A \\ & 2 B A \end{aligned}$ | $\begin{aligned} & \text { No. } \\ & 859 \\ & 860 \end{aligned}$ | Price $14 p$ 12p | Type 4BA 6BA | $\begin{aligned} & \text { No. } \\ & 851 \\ & 882 \end{aligned}$ | $\begin{aligned} & \text { Price } \\ & \text { 12p } \\ & \text { 12p } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| SOLDER TAGS-not tinned supplied in multiples of 50. |  |  |  |  |  |
| Type | No. | Price | Type | No. | Price |
| 08A | 851 | 40p | 48A | 853 | 22p |
| 28A | 852 | 28p | 6BA | 854 | 22p |

TRANSFORMERS


## CASES AND BOXES

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| No. | Length | Width | Height | Prico |
| :---: | :---: | :---: | :---: | :---: |
| 155 | 8 in | 51 in | 2 in | E1.25 |
| 156 | 11 in | 6 in | 3 in | ¢2.12 |
| 157 | 6 in | 43in | 佬行 | 51-30 |
| 158 | sin | 5317n | 21 in | E1-78 |
| ALUMINIUM BOXES. Made from bright ali., folded construction each box complete with half inch deep Iid |  |  |  |  |
|  |  |  |  |  |
| and scre No. | tength | Width | Height |  |
| 159 | $5{ }^{\text {2 }}$ in | 2 in | 1/in | Price |
| 160 | 4 in | 4 in |  | 82p <br> $62 p$ |
| 161 | 4 in | $2 \pm \frac{10}{10}$ |  |  |
| 162 | 5yin | 4 in | 119 | ${ }^{70 \mathrm{p}}$ |
| 163 | 4 in | $2{ }^{21} \mathrm{I}$ | 2 in |  |
| 164 | 3 in | 2 in | 1 in | 4 P |
| 165 | 7in | 51 n | 21 in | E1-09 |
| 166 | 8 in | 6 \% | 3 in | E1-32 |
| 167 | 6 in | 4 in | 2 in |  |
| V.A.T. Add $12 \frac{1}{j} \%$ to prices marked ${ }^{*}, 8 \%$ to those unmariked Items marked are zero rated. |  |  |  |  |
|  |  |  |  |  |  |  |
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## REG 1 - POWER SUPPLY

The regulator module, REG 1 provides $15-0-15 v$ to power the CPR 1 and MC. 1. It can be used with any of our power amp supplies or our small transformer TR 6 . The power amp kit will accorn-

## POWER AMPLIFIERS

It would be pointless to list in so small a space the number of racording studios, educational and government establishments, etc. who have been using CaiMSON amps satisiactority for quit some time, We have a reputation for tha highest quality at the lowest prices. The power amp 8 availabie in five types, they all have the same spaciticationignal to noise ratio 110 dB ; frequency 8 ohms; 71.0 . insignificant; slew rate $775 \mathrm{mV}(250 \mathrm{mV}$ or 100 mV on request) ; size $120 \times 80 \times 25 \mathrm{~mm}$.

POWER SUPPLIES
ies which use our superb TOROIDAL transformers only 50 mm We produce suitable power supplies which use our superb ToROIDAL/ransformirs).

## POWER AMPLIFIER KIT

The kit includes all metalwork, heatsinks and hardware to house any two of our power amp modules plus a power supply, It is contemporarily styled and its quaiity is consistent with that of it with confidence in a few hours.


ACTIVE CROSSOVERS
$\times 02$
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D
OUBTLESS some builders of the EE2020 Tuner Amplifier will be looking around for a pickup-turntable unit. For them, and others starting out into the hi fi field, the choice has to be made from a large range of record playing equipment, and many factors have to be taken into account.

There is one particularly interesting possibility for the handyman to consider. That is the purchase of a kit of parts for assembly at home. This is precisely what is offered by Connoisseur with their new "Package Deal" which has been announced at an opportune time. It comprises the BD1 Turntable Kit, SAU2 Pickup Arm, and Plinth with acrylic cover.

The BD1 Belt Drive Turntable Kit has been available for home assembly for many years, and is justifiably popular. It operates at $331_{3}$ and 45 r.p.m. using a flexible belt drive system. The Connoisseur Pickup Arm Type SAU2 is equally popular and is an ideal complementary precision instrument. The wooden Plinth provides stylish housing and completes the ensemble.

The work involved in assembling this kit is by no means onerous, and requires hardly anything more than a screwdriver and a pair of pliers, save a special spanner-and that's provided.

## BD1 BELT DRIVE TURNTABLE KIT

The eight-page instruction booklet gives minutely detailed assembly instructions for the BD1 Belt Drive Turntable Kit and is well illustrated with practical diagrams.
This booklet, however, requires revising, for our copy states on the
front page "Speed change is accomplished by push button. . .". In fact, a lever is used for this purpose on the actual equipment submitted to us.

We followed the stage-by-stage directions, during which it became apparent that further revisions are required to this booklet. "Operation $1^{\prime \prime}$ revealed that a different type of connector block was supplied to that illustrated.

## KIT SPECIFICATION

MAINS SUPPLY
200/240 or $100 / 125$ volts a.c. Adjustable by user. Power, 5W. MOTOR
Synchronous constant speed motor runs at 375 r.p.m, at 50 Hz ; 450 r.p.m. at 60 Hz . Drive pulleys for 50 Hz or 60 Hz available.
TURNTABLE
$10 \frac{1}{4}$ in ( 260 mm ) lathe turned aluminium casting. Precision ground spindle. Manganese bronze bearing.
DRIVE
Precision ground round section rubber belt. Unique drive arrangement via groove in outer surface of turntable rim.
PERFORMANCE

Wow and flutter: Less than $0.1 \%$ FINISH
Black hammer with satin silver and matt black trim.
Control knob: bright finish turned alumium.
Plinth: Afromosia wood, waxed finish. Dust cover: transparent acrylic.

The mains switch panel assembly came as a completed unit. The 10W ballast resistor in the mains input line was mounted in such a way that one end would have made direct contact with the brass spacing, pillar when this was fitted. (Mains to chassis!) Disaster was averted by easing the resistor upwards and away from the pillar hole by a distance of $1_{4}$ inch. (In addition to this action, it would be a wise precaution to fit a piece of plastic sleeving to this pillar.) NOTE: A plastic-sleeved pillar is now being provided by the manufacturer.

The motor came with the two-speed pulley already fitted. After the motor was mounted, the turntable and belt in position, it was found necessary to move the pulley slightly upwards on spindle to align correctly with turntable belt groove.

Hand rotation of the turntable proved the correct operation of the speed change device: the belt moving from one pulley groove to the other as the lever was thrown.

A hinged plastic cover protects the pulley. There is not too much room for comfort. When initially closed the cover was found to foul the pulley. The only remedy was to slacken the motor screws and ease the motor away (towards the rear corner of the platform). This cleared the trouble.

Then mains was applied and the action of the speed change checked. It functioned perfectly, at $331_{3}$ and 45 r.p.m.

## SAU2 PICKUP ARM

A template is provided to ensure correct positioning of pickup base in relation to the turntable centre spindle. But this was not required in the present case since the mounting board (part of the BDI Plinth and Cover Kit) came already drilled for the pickup and the arm rest. It was therefore a simple task to mount and secure the pickup with its locking nut.
Dual signal output leads are supplied. These terminate at one end in a 5 -pin plug and this fits into the socket on pickup base. The other end of the cables are fitted with phono plugs. A separate earth lead is provided and is intended to be connected to the preamplifier chassis.

The head shell is suitable for all standard cartridges with $1_{2}$ in fixing centres. Lateral adjustment is provided to give minimum tracking error.
The SAU2 instuction leaflet explains how to adjust for required height of arm; for required tracking weight, which is performed with the aid of the precision balance supplied; and for correct bias weight. An additional small weight is supplied to substitute for the larger weight when lower tracking weights are used.
A hydaulic arm lifting device is incorporated. This is set correctly at the factory.

The Connoisseur
BD1/SAU2 "Package Deal" is currently available from retailers at a price of less than $£ 50.00$ (VAT extra).


THIS month we explain some of the alpha, and alpha-numeric codes to be found on resistors, capacitors and transistors to aid value and application identification.

## BS1852 RESISTANCE CODE

It is common nowadays to find the value of a resistor printed on the body in the form of an alpha-numeric code.

Letters R, K, and M, are used as multipliers $\mathrm{R}=\times 1, \mathrm{~K}=\times 1,000$ and M $=\times 1,000,000$. The position of these letters among the two most significant digits also acts as the position of the decimal point. For example, $4 \mathrm{~K} 7=4 \cdot 7$ kilohms, $\quad \mathrm{M} 22=0.22$ megohms $=220$ kilohms.

The tolerance of the resistors is indicated by the addition of a second letter appearing in the last position: $\mathrm{F}=1 \%, \quad \mathrm{G}=2 \%, \quad \mathrm{H}=2.5 \%, \quad \mathrm{~J}=5 \%$, $\mathrm{K}=10 \%, \mathrm{M}=20 \%$.

Example of the full code: 4K7J $=4 \cdot 7$ kilohms $\pm 5 \% ; 2 \mathrm{R} 2 \mathrm{~K}=2 \cdot 2 \mathrm{ohms}$ $\pm 10 \%$.

## ALPHA-NUMERIC CAPACITOR CODING

Ceramic plate capacitors of Russian origin are commonplace nowadays and these use a alpha-numeric coding similar to the above-resistor system. For example $33 \mathrm{~K}, 4 \mathrm{~K} 7$, the K acting as a $\times 1000$ multiplier and decimal point position giving the value in picofarads. Therefore $33 \mathrm{~K}=33,000 \mathrm{pF}$ or $0.033 \mu \mathrm{~F}$, and $4 \mathrm{~K} 7=4,700 \mathrm{pF}$ or $0.0047 / \mathrm{F}$. The K can also be used merely as a decimal point with the units in nanofarads, (nF), i.e., 4K7 = $4 \cdot 7 \mathrm{nF}$.

The letter H is sometimes used in place of K for some values (reason unknown) e.g., $1 \mathrm{H} 5=1 \cdot 5 \mathrm{nF}$ or $1,500 \mathrm{pF}$ or $0.0015 \mu \mathrm{~F}$.

We have also in our possession some ceramic capacitors (unknown origin) marked 4 n 7 which is $4 \cdot 7 \mathrm{nF}$. Also another tiny bead type ceramic marked with only 10 , the value being 10 pF .

## FOR BEGNNERS

## TRANSISTOR CODING

Since the invention of the transistor in 1948 there have been tens of thousands of transistors manufactured and marketed. There are many different shapes, sizes, pinning configurations and employing various numbering systems for identifying the devices.

At present there are three main numbering systems and a number of "house" systems from different manufacturers.
In the USA, the "Jedec" system is used and prefixes three and four digit numbers by " 2 N ". Tables have to be consulted for all information concerning the device.

In Europe the "Proelectron" system is employed. Two letters followed by three numbers (eg BC109) are classified as consumer devices while three letters/two numbers (eg BFY52) identifies an industrial device.
The first letter is always A (germanium) or B (silicon), see Table 1 for decoding of second letter.

## JAPANESE SYSTEM

The Japanese Institute of Standards has the " $2 S$ " system. The $2 S$ is followed by a letter and then

TABLE 1: Explanation of the Pro-electron coding system for transistors.

numerals, the letter having application significance: A pnp/high frequency; B pnp/low frequency; C npn/ high frequency; D npn/low frequency; $\mathbf{H}$ unijunction; $\mathbf{J} p$-channel f.e.t.; $\mathbf{K} n$-channel f.e.t.; OS photodevice.

Some Texas Instruments devices are coded with a $2 S$ followed by up to four numerals.

## HOUSE CODES

Some of the popular "house" codes are: Motorola: MJ - Silicon, power, metal MJE -Silicon, power, plastic
MP -Germanium, power, metal
MPF -f.e.t., plastic MPS -small signal, plastic
Texas:
TIP -power, plastic
TIS -Small signal
Ferranti: ZTX - Small signal, plastic

## ABBREVIATIONS

The direct meaning of some common abbreviations to be found in EVERYDAY ELECTRONICS are shown below.


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| :---: | :---: | :---: | :---: | :---: |
| SN74HION | 0.55 | SN74S03N 0.77 | 74LS161N | 0.85 |
| SN74H11N | 0.55 | SN74S04N 0.94 | $74 \mathrm{LS162N}$ | 0-80 |
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| SN74H21N | 0.55 | SN74S20N 0.77 | 74LS164N | 1.10 |
| SN74H30N | 0.55 | SN74S40N 0.77 | 74 LS 165 N | 1.15 |
| SN75H40N | 0.55 | SN74S64N 0.77 | 74LS166N | 1.65 |
| SN74H51N | 0.55 | SN74S65N 0.77 | 74LS168N | 1.45 |
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| 74LS02N | 0.26 | SN74S387N 3.05 | 74 LS 196 N | 0.80 |
| 74LS03N | $0 \cdot 26$ | SN74S470N 5.06 | 74LS197N | 0.80 |
| 74 LS 04 N | 0.29 | SN74S471N 5.06 | 74LS221N | 0.80 |
| $74 \mathrm{LS05N}$ | 0.29 | SN74S472N13-48 | 74LS240N | 1-50 |
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## The Exira ordinar Experiments of

 Proiesso Eversure by Anthony John BassettBOB and the Prof. are inside a giant vacuum chamber Dressed in "space suits" which the Prof's robots have made to protect them from the vacuum, they are carrying out experiments on uses of electronics in vacuum conditions. As Bob and the Prof. are both keen followers of the Space Explorations which are in progress, they can see that there is certain to be a large expansion in the use of electronics in the vacuum of Space; a very interesting field to follow and one which will in future provide an increasing number of really interesting jobs for electronics experts.

Who knows, some of today's readers of this magazine might be fortunate enough to participate in this exciting work!

## ELECTRON BEAM

Bob has noticed that an electron beam when passed through various vapours which the Prof. has released into the vacuum chamber, does not appear to be scattered by collision with the molecules of vapour; rather to his surprise, the opposite seems to happen and the presence of the vapour molecules seems to improve the focusing of the electron beams.
They can observe the electron
beam directly in the presence of certain vapours which glow with the spectral colours of the elements or ions present along the path of the beam.
"It is the presence of a trail of positive ions along the path of the electronbeam which keeps the path of the beam narrow and prevents the electrons from being scattered," the Prof. informed Bob.
"When an electron beam passes through a vapour or gas at low pressure, it produces mainly positive ions from the atoms or molecules which it encounters. Thus the presence of a gas vapour is necessary in order to observe this effect."
"I find this very puzzling, Prof. Surely the beam should produce equal numbers of positive ions and negative ions? At school we are taught that when an ionising beam passes through a gas or vapour, it spits the molecules into equal numbers of opposite charges! How can the electron beam produce more positive ions than negative ions when the molecules of the substance are neutral to start with?"
The Prof. explained: "This is because the mechanism of ion production is different. With a beam of ionising radiation, the
radiation is so energetic that it splits the molecules up into ions of opposite charge. The oppositely charged ions fly apart under the influence of the radiation bombardment, and exist as charged particles for a time until they become neutralised.
"The electron beam which we are using behaves differently and the predominant mechanism of ion production is by 'knocking off' electrons from the outer electronic shells of atoms to produce simple ions, or vapour molecules to produce more-complex positive ions. Only a small number of molecules are split, so the production of negative ions along the path of the electron beam is comparatively little."

## CORE OF POSITIVE IONS

"The positive ions are much heavier than electrons and they remain in the path of the electron beam to give a 'pencil' of positive charge along the path of the beam, and this tends to keep any electrons which tend to stray or scatter, attracted back to the path of the beam, thus improving the focus of the beam and the 'point size' where it may hit a fluorescent viewscreen or radar screen.
"Most of the electrons which are
picked off to do not join the main electron beam, but travel for a short distance away from the beam, and then combine with a neutral molecule to form a negative ion. Thus, surrounding the beam and its core of positive ions is a tube of negative ions which help to keep the electrons of the beam on the correct path, as any which stray are repelled back by the negative charge.

## BETTER RADAR

"This combined action of attraction by a positive column of ions and repulsion by a surrounding tube of negative ions, is a powerful influence in keeping the electrons to a narrow path to give a finely focused radar display.
"Use of this technique has resulted in very considerable savings in cost of production of radar sets, because the electron beam can remain focused with a much lower e.h.t. voltage than would be needed for a radar tube which used a high vacuum. The radar tube itself can be cheaper because it does not need a separate e.h.t.
connection. Due to the lower voltage used, this may be connected in some cases on an ordinary tube base.
"The power supply is much simpler, and because the velocity of the electrons in the beam is lower, less deflection energy is needed, which means that the deflection circuits are less critical and less likely to go wrong!
"So, just by introducing a very small quantity of gas into each radar tube, the makers have contributed to the spread of the safety which has been brought about by the increased use of radar, both by making it affordable for more users and more reliable!
"Just as the developments in rocketry by the Germans during World War II were very important in forming the basis for the development of modern space rockets, so the British developments in radar and ion beam technology are now being taken further, and the behaviour of such beams in the conditions of space will be very important for future
explorations and uses. Such beams may be used to probe areas where men cannot go!"
"This sounds really interesting, Prof., can you tell me more about the use of electronically produced beams in space?"
"Yes, certainly, Bob."
To be continued

## Crossword No. 16-Solution



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the attack and decay and also providing filtering. Does not the attack and decay and also providing filtering. Does not be used with it and with other electronic instruments. Component set using dual rotary pot Printed circulit board
GUITAR SUSTAIN (P.E. Oct. 77)
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GPO cables te Functions yor well as a signal tracer for fault finding in radios etc. etc. By
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Limpet Stat must be mounted in close contact callibrated $90^{\circ}-190^{\circ} \mathrm{F} 15 \mathrm{amp}$ conAppliance Stat fix Hke a volu

$$
\begin{aligned}
& 16 \text { amp contact } 30-80^{\circ} \mathrm{E} 85 \mathrm{p} \text {. } \\
& \text { dito but for }
\end{aligned}
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frec,uency to 25 Hz , this freouency gives correct frecuency to 25 Hz , this frequency gives correct ringing note
for GPO bells. These units were made for the GPO so obviously are first class. Completely enclosed and safe to mount on the wall or stand on a shelf. Price $£ 3-20$.
Telephone Extension Ealls in bakelite wall box, these will save you missing calls when you are out in the garden or shed, etc. Price 23. 16.
Variable Mains Supply A bench mounting unit which con-
tains an isloation transformer for safetty tains an isloation transformer for safety and a 2 amp variac
for adeptability. With this you will beableto getcontinuouly foradeptability. With this you will beable to get continuously
variable mains supply from zero to full voitage at 2 amps. A real time saving device, price only $£ 20 \cdot 75$.
Answering Machines still available, see this month's newsletter but supplies are going down rapidly and this may well be your last chance to acquire one of these. Spares + accessories also available. A very large purchase this month enables us to offer a range
of radio Htems. You will find the prices well below average Cassefte Recorder/Player Japanese or Hong Kong mad these have all the normal facilities record, playback, fasi rewind, etc, also sockets for stop/start, microphone, earphone
and lead for mains as these and lead for ma
batteries. 89.50 .
AM/FM Radios There's no doubt that FM does give better reproduction in good areas so a more adult member of the
family wlll be pleased with one of these. The ones we have are in leatherette cases and are battery/mains radios having the malis unit buitit in and are completete with mains plug. These cover medium wave and VHF with optional AFC. Price 26 -75. Soft Toy Radios Not necessarily only for the younger memappeal. Dolls, poodles, elephanis and rabbits each with zip compartment at the bottom where the radio fits. Medium wave only, working from PP3 batteries. When ordering please state preference and if possible give an alternative. $£ 4 \cdot 50$. Car Battery Power Unit made for Rank Radio. This unit has been designed to operate $6 V$ battery powered equipment
from a 12 V car battery, it ilsed voltage and gives protection to your equipment in case of accidental reversal of connections also against excessive car battery voltage should this occur. The unit is very robust and virtually everlasting if used sensibly, It uses a negative earth circuit butit will operate in a positive earth car providing the instrument being played is not connected to the car
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